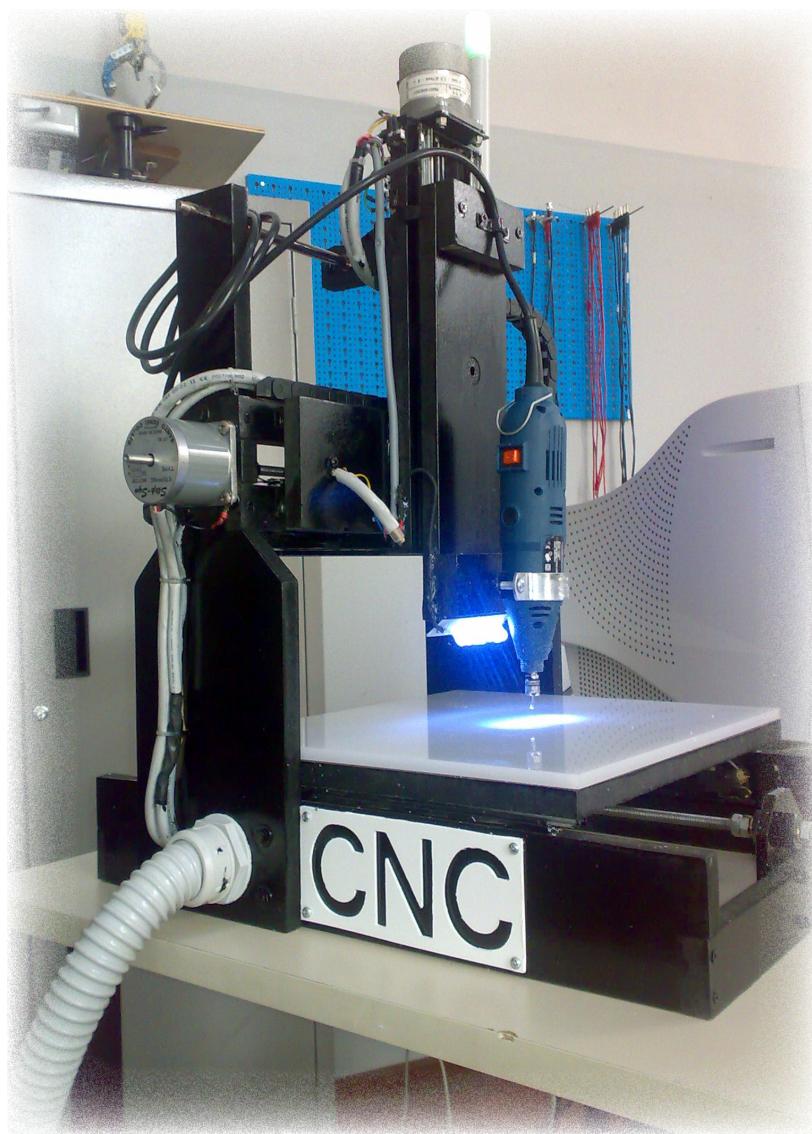


*Specializzazione: Elettronica e Telecomunicazioni*

Ceccato Roberto - Nassuato Denis

5BET



**CNC a 3 ASSI**

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## ⇒ SWITCHING O “A COMMUTAZIONE” - ATX

Possiedono circuiti più complessi rispetto ad un alimentatore tradizionale, ma hanno diversi vantaggi, tra cui un minore ingombro e peso a parità di potenza, un rendimento maggiore e quindi minore calore prodotto, sono meno adatti per l'uso in laboratorio, essendo caratterizzati: da un elevato ripple; e dalla generazione di componenti spurie ad alta frequenza che possono interferire nel funzionamento di alcune apparecchiature.

Il principio di funzionamento si basa sul fatto che un trasformatore è più efficiente, richiede un nucleo ferromagnetico più piccolo ed è molto più compatto, a parità di potenza, all'aumentare della frequenza operativa. Negli alimentatori elettronici vengono utilizzati particolari trasformatori fatti funzionare a frequenze di decine o centinaia di migliaia di Hertz invece dei 50 Hz della rete elettrica di distribuzione. Il nucleo di questo trasformatore è in Ferrite, materiale realizzato con polveri metalliche incollate, invece dei tradizionali lamierini di ferro, che alle alte frequenze comporterebbero una notevole perdita di energia.

In un alimentatore elettronico la tensione di rete viene per prima cosa raddrizzata e livellata con un condensatore. Successivamente un circuito oscillatore genera a partire da questa corrente continua, una corrente alternata di elevata frequenza, che viene applicata ai capi dell'avvolgimento primario del trasformatore; la tensione in uscita, presente ai capi dell'avvolgimento secondario del trasformatore, viene raddrizzata e livellata.

La funzione di stabilizzazione è solitamente ottenuta retroazionando l'errore del segnale in uscita sul regime di funzionamento dell'oscillatore. In pratica, un circuito misura la tensione di uscita, e se questa risulta troppo alta viene ridotta l'energia inviata dall'oscillatore al trasformatore, se invece la tensione scende, viene aumentato il flusso di energia. Grazie a questo sistema molti alimentatori switching sono in grado di accettare in ingresso un'ampia gamma di tensioni e frequenze. Per esempio gli alimentatori per Notebook spesso possono essere collegati sia alla rete europea a 230 V/50 Hz, sia a quella statunitense a 115 V/60 Hz.

L'apparato è reso più complesso dalla presenza di sistemi di protezione contro sovraccarichi e cortocircuiti, e da filtri necessari per evitare che il segnale ad alta frequenza si propaghi verso il carico oppure ritorni verso la rete elettrica.

### ● IL RIPPLE

È uno dei parametri che caratterizzano la qualità di un alimentatore, la sua misura si effettua con l'oscilloscopio accoppiando l'ingresso in AC. Si tratta principalmente del rimanente residuo della componente alternata della rete elettrica, rettificata dai diodi e livellata dai condensatori di filtro; a questa piccola componente oscillatoria spuria, possono aggiungersi disturbi indotti dal carico o dovuti al funzionamento interno dell'alimentatore (presente soprattutto negli alimentatori a commutazione). Questo residuo è costituito da una lieve oscillazione della tensione di uscita avente, nel primo caso, andamento identico alla sinusoide di ingresso, in pratica questo piccolo disturbo si sovrappone alla tensione continua fornita in uscita. Per usi generici, questa componente residua risulta ininfluente sull'utilizzo, in altri casi, esempio fornire alimentazione ad un circuito in progetto, operante a bassi livelli di tensione o elevato guadagno, un ripple relativamente alto può essere di disturbo. Per questa ragione i migliori alimentatori da laboratorio sono in tecnologia lineare, con un ripple garantito su tutto il range di tensione/corrente, contenuto entro pochi millivolt.



ALIMENTATORE ELETTRONICO ATX DI UN COMUNE PERSONAL COMPUTER, PRIVATO DEL COPERTO

## ⇒ MOTORI STEPPER

I motori passo-passo, spesso chiamati anche step o stepper, sono considerati la scelta ideale per tutte quelle applicazioni che richiedono precisione nello spostamento angolare e nella velocità di rotazione, quali la robotica, le montature dei telescopi ed i servomeccanismi in generale. Tuttavia ultimamente, per le applicazioni high-end, vengono spesso sostituiti da motori brushless o da attuatori voice-coil.

Indice

- 1 Vantaggi dei motori passo passo
- 2 Difetti dei motori passo passo
- 3 Il principio di funzionamento
- 4 La costruzione elettromeccanica



### VANTAGGI DEI MOTORI PASSO PASSO

- Quelli costruiti con tecnologia comune hanno un costo non elevato, relativamente ad altri tipi di motore con analoghe prestazioni.
- È possibile realizzare azionamenti di precisione controllati da computer in catena aperta, cioè senza utilizzare sensori di posizione o di velocità. Sono quindi utilizzabili con relativa semplicità e senza richiedere particolare potenza di calcolo.
- Hanno un'elevata robustezza meccanica ed elettrica: infatti non esistono contatti elettrici strisciante e, se necessario, possono essere realizzati anche in ambiente completamente stagno.
- È facile far compiere all'albero piccole rotazioni angolari arbitrarie in ambedue i versi e bloccarlo in una determinata posizione.
- La velocità di rotazione può essere molto bassa anche senza l'uso di riduttori meccanici.
- Hanno molto spesso momento d'inerzia piuttosto basso
- Sono molto stabili nella posizione a rotore bloccato e non presentano pendolamenti come nei sistemi brushless
- Se dimensionati bene non necessitano di alcuna taratura.

### DIFETTI DEI MOTORI PASSO PASSO

- Richiedono sempre circuiti elettronici per il pilotaggio, in genere di tipo digitale.
- Hanno un funzionamento a scatti e producono vibrazioni, soprattutto ai bassi regimi e se si adottano le tecniche di pilotaggio più semplici.
- Il loro rendimento energetico dipende dalla tecnologia costruttiva adottata, la potenza meccanica espressa come coppia e misurata in Nm (Newton per metro), a parità di assorbimento in corrente, dipende spesso dal tipo di pilotaggio elettrico/elettronico adottato.
- Permettono una velocità di rotazione massima intorno a 1000-1500 rpm. Esistono tuttavia motori che raggiungono i 4000-5000 rpm tramite sistemi di retroazione ad anello chiuso. La loro caratteristica di coppia tuttavia scende quasi esponenzialmente al crescere della velocità.

## IL PRINCIPIO DI FUNZIONAMENTO

I motori passo-passo sono motori che, a differenza di tutti gli altri, hanno come scopo quello di mantenere fermo l'albero in una posizione di equilibrio: se alimentati si limitano infatti a bloccarsi in una ben precisa posizione angolare.

Solo indirettamente è possibile ottenerne la rotazione: occorre inviare al motore una serie di impulsi di corrente, secondo un'opportuna sequenza, in modo tale da far spostare, per scatti successivi, la posizione di equilibrio.

È così possibile far ruotare l'albero nella posizione e alla velocità voluta semplicemente contando gli impulsi ed impostando la loro frequenza, visto che le posizioni di equilibrio dell'albero sono determinate meccanicamente con estrema precisione.

## LA COSTRUZIONE ELETTROMECCANICA

I motori passo-passo si dividono tradizionalmente in tre grandi gruppi: motori a magnete permanente, motori a riluttanza variabile e motori ibridi; questi ultimi sono i migliori. In realtà la quasi totalità di quelli oggi reperibili sono proprio del terzo tipo.  
Struttura di un motore passo-passo

Un motore ibrido è costituito da un rotore e da uno statore; nella fotografia a lato è riportato un esemplare non particolarmente recente ma in cui si vede chiaramente la struttura.

Il rotore appare come una coppia di ruote dentate affiancate e solidali all'albero (i "denti" sono chiamati coppette) costituite da un nucleo magnetico (le due ruote sono permanentemente magnetizzate, una come NORD, l'altra come SUD) e le coppette in materiale ferromagnetico. Il numero di denti è variabile ma 50 è in assoluto il più frequente. Tra le due ruote è presente uno sfasamento esattamente pari ad 1/2 del passo dei denti: il dente di una delle due sezioni corrisponde quindi alla valle dell'altra. Nel rotore non sono presenti fili elettrici e quindi manca completamente ogni connessione elettrica tra la parte in movimento e quella fissa. In genere il rotore è montato su cuscinetti a sfera, anche nei modelli economici.



Lo statore appare come il classico insieme di avvolgimenti ed il circuito magnetico è costituito da 4 o, più frequentemente, 8 "espansioni polari" (otto in quello mostrato fotografia). All'interno dello statore sono presenti piccoli denti che si affacciano esattamente a quelli del rotore o meglio, sono esattamente affacciati al rotore solo il gruppo di denti appartenenti ad una espansione polare e a quella opposta; le altre coppie sono sfalsate rispettivamente di 1/4, 1/2 e 3/4 del passo dei denti. Avvolti intorno ai poli magnetici dello statore ci sono i fili che, opportunamente percorsi da corrente, generano il campo magnetico.

All'esterno sono evidentemente presenti le alimentazioni dei vari avvolgimenti; in pratica le fasi possono essere avvolte secondo due schemi:

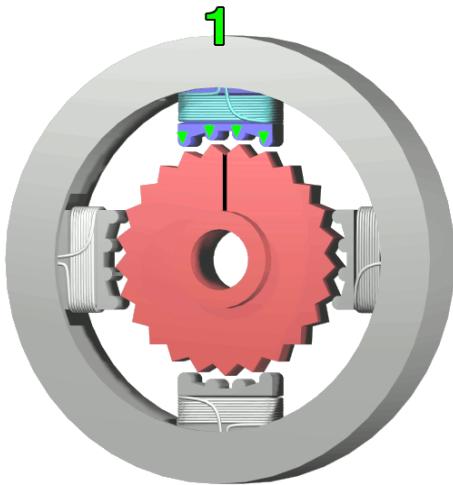
- Sono presenti due soli avvolgimenti (avvolti su più espansioni polari) e quindi all'esterno arrivano due sole coppie di fili: in questo caso si parla di motori bipolari in quanto la corrente dovrà percorrere le fasi nei due versi al fine di creare gli opportuni campi magnetici.
- Sono presenti quattro avvolgimenti avvolti a coppie, in antiparallelo, sulle espansioni polari; all'esterno arrivano almeno cinque fili (spesso sono infatti presenti delle connessioni interne al motore tra le varie fasi).

Si parla in questo caso di motori unipolari in quanto la corrente nella singola fase ha sempre lo stesso verso. È possibile creare due campi magnetici opposti semplicemente scegliendo in quale dei fili debba passare la corrente.

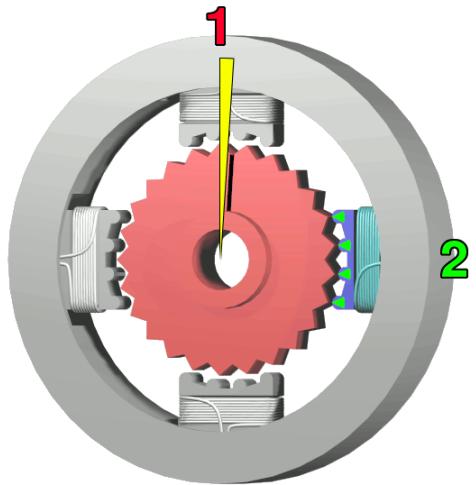
Una tipologia particolare di motore passo-passo è utilizzabile sia in configurazione unipolare che bipolare: si tratta di quelli a 6 o 8 fili.

Il numero di differenti posizioni di equilibrio presenti in una rotazione completa dell'albero è in genere indicato come passi per giro e dipende del numero dei denti del rotore e dai poli dello statore, non dal numero di fili uscenti o dal numero delle fasi, questo numero è spesso stampato sul contenitore ed espresso in gradi.

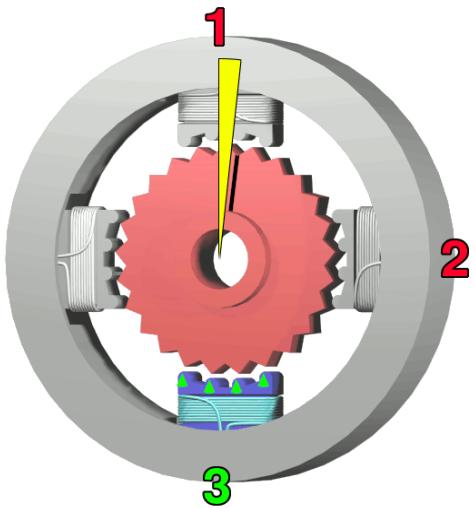
## ⇒ CONFIGURAZIONI A,B,C,D IN MODALITÀ PASSO INTERO



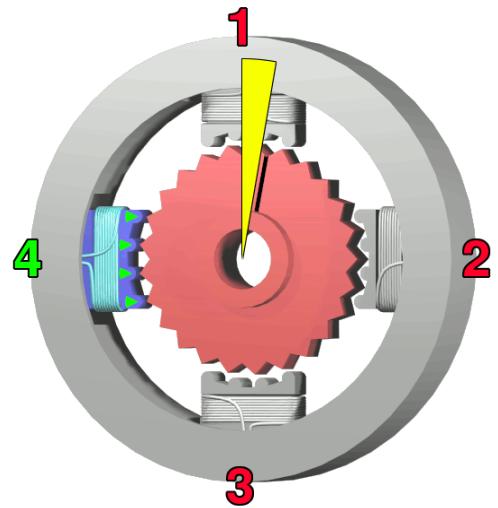
- PRIMA CONFIGURAZIONE



- SECONDA CONFIGURAZIONE



- TERZA CONFIGURAZIONE



- QUARTA CONFIGURAZIONE

## ⇒ BREVE STORIA E INTRODUZIONE AL CONTROLLO NUMERICO

Le macchine CNC, che fino agli anni '80 erano usate solo per lavorazioni ad alta precisione, sono oggi molto diffuse e impiegate in quasi ogni campo della meccanica. La tecnologia delle macchine CNC ha coperto un po' tutti i rami della meccanica; le macchine a CNC più comuni sono presse piegatrici, punzonatrici, torni, fressatrici e macchine di taglio lamiera (laser, ossitaggio, plasma, waterjet ecc.). Esse rappresentano l'evoluzione delle macchine NC, perché permettono il controllo numerico diretto da un computer esterno (DNC).

CNC è l'abbreviazione di "Computer Numeric Control", ossia una macchina per lavorazioni meccaniche, vedi tornio o fresa, la quale viene comandata da un computer per eseguire delle lavorazioni o pezzi.

Con un programma CAD (vedi Autocad), viene disegnato il particolare da realizzare, poi con un programma CAM (vedi Mastercam) viene generato il programma usato dal controllo del CNC.

Una macchina a CNC tipo hobbistico è generalmente costituita dalle seguenti parti:

- o Computer
- o Programma di controllo CNC
- o Il file in formato Gcode del pezzo o lavorazione da eseguire
- o Il controllo elettronico del CNC (Stepper drivers)
- o Il gruppo alimentazione del controllo elettronico.
- o I motori passo passo (step motor)
- o I programmi CAD, CAM ed interpreti vari.
- o La fresa che esegue le lavorazioni

## COMPUTER

Su questo computer è di norma installato il programma di controllo del CNC

- La porta parallela è il modo di comunicazione normalmente più usato tra computer ed il controllo elettronico.

## PROGRAMMA DI CONTROLLO CNC

• Nel nostro caso abbiamo utilizzato il software MACH3 della artsoft. Il suo compito è quello di trasformare il file in Gcode del pezzo da eseguire in una serie di segnali (passo e direzione), da inviare ai motori passo passo della fresa.

• Con questo programma è normalmente possibile eseguire spostamenti manuali degli assi della fresa, introducendo opportuni comandi sulla finestra video del programma di controllo (se presente).

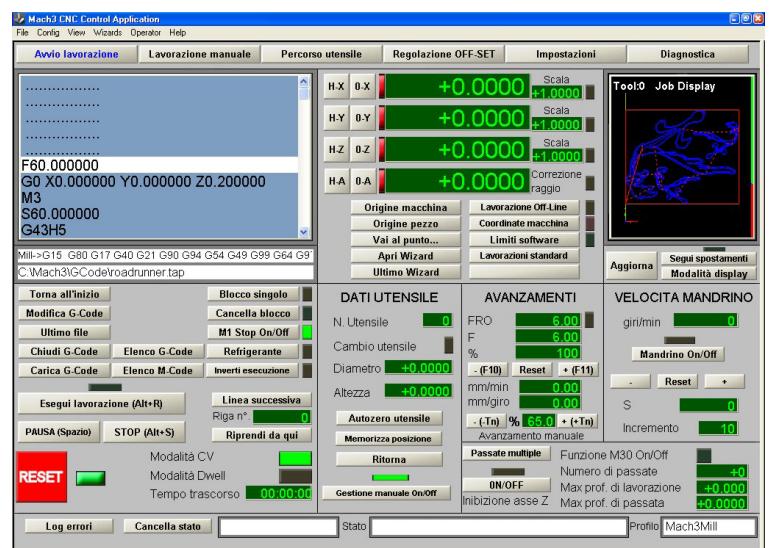
• Con riferimento al programma MACH3, utilizzando opportuni tasti della tastiera, si ottiene manualmente il movimento assi della fresa.

• È inoltre possibile inserire direttamente e mandare in esecuzione comandi in Gcode.

• Su di una finestra del programma è possibile vedere il disegno del percorso utensile inerente il pezzo in lavorazione.

Una ulteriore finestra visualizza le coordinate dei tre assi e in altre finestre vengono visualizzati altre informazioni d' impostazioni e lavorazioni in corso.

- Ogni programma di controllo ha una sezione dedicata ai parametri di settaggio che generalmente sono:
- Identificazione della porta parallela che comunica con il computer.
- Il settaggio dei pin di tale porta, con il quale si determina l'associazione del passo/direzione di ogni mo-

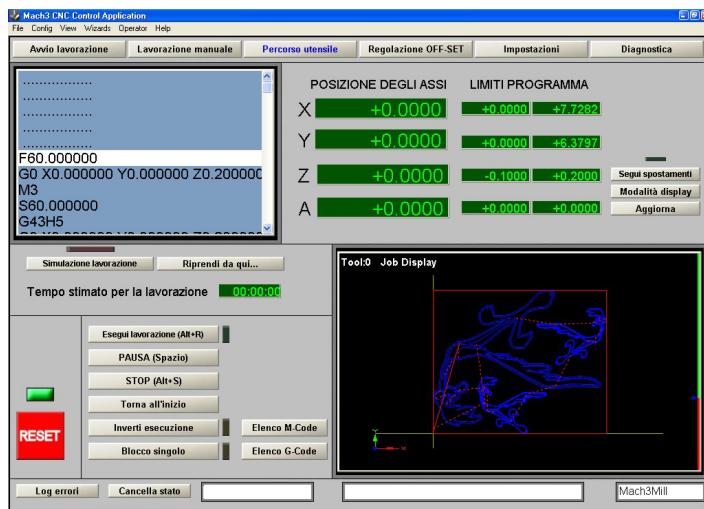


SOFTWARE MACH3 UTILIZZATO

tore passo passo con il relativo pin e dei vari segnali di input e output usati.

- L'impostazione del valore di spostamento per ogni passo di rotazione dei motori.
- Le massime velocità ed accelerazioni che i motori possono tenere.
- Eventuali compensazioni di gioco (backlash correction) che possono essere presenti tra le boccole e aste filettate degli assi di spostamento.
- Settaggi per eventuali fine corsa installati.

File della parte o lavorazione da eseguire



• È un file generalmente creato da un programma CAD/CAM, oppure può essere generato da un programma di conversione tipo ACE che lo ricava dal file di disegno in formato .dxf precedentemente realizzato con un programma CAD.

- Questo file è in formato ASCII e può essere editato con Notepad.

• Di norma questo file è scritto in linguaggio Gcode, alcune volte in HPGL (file usato per il plottaggio di disegni).

- Un file Gcode è composto da: Blocchi, Comandi e Coordinate.

• Blocchi: sono linee ASCII contenenti i comandi e le coordinate, sono praticamente uno o più comandi che devono essere eseguiti prima di passare al blocco successivo.

- Comandi: sono composti da una lettera e da un numero, i più comuni sono i seguenti:

- ⇒ G00 Avanzamento rapido con interpolazione lineare
- ⇒ G01 Interpolazione lineare (avanzamento con moto di lavoro)
- ⇒ G02 Interpolazione circolare in senso orario
- ⇒ G03 Interpolazione circolare in senso antiorario
- ⇒ G04 Arresto temporizzato
- ⇒ G17 Scelta del piano XY
- ⇒ G18 Scelta del piano XZ
- ⇒ G19 Scelta del piano YZ
- ⇒ G70 Misure in pollici
- ⇒ G71 Misure in mm
- ⇒ o G90 Coordinate assolute
- ⇒ G91 Coordinate relative.

- Vi sono molti altri comandi tipo M usati per controlli ausiliari in fase di lavorazione.

Il file in formato Gcode, pur essendo costruito con comandi standard ed unificati, può non essere letto da qualsiasi programma di controllo CNC.

A livello hobbistico, i programmi di controllo CNC che girano sotto DOS, riescono a riconoscere ed elaborare solo una piccola parte di questi comandi Gcode e generalmente sono sufficienti per la realizzazione di lavorazioni anche complesse.

Questo insieme di comandi che il nostro programma di controllo CNC riesce a riconoscere viene denominato “Post-processore”.

Il programma CAM quindi che dovrà trasformare il disegno della parte da realizzare nel file Gcode, dovrà essere settato ed impostato in modo tale da usare il Post-processore del programma di controllo CNC che andrà a comandare la fresa, ossia dovrà impiegare solo un certo tipo di comandi Gcode che possono essere interpretati dal programma di controllo CNC.

## CONTROLLO ELETTRONICO (STEPPER DRIVERS)

- È costituito da una o più schede elettroniche che alimentano e fanno muovere i motori passo passo nella giusta sequenza ed in maniera sincronizzata.

- Il controllo elettronico può essere generalmente di due tipi:

- o Per motori unipolari, si riconosce da vivoste resistenze presenti in vicinanza del connettore di collegamento dei motori.

Sono controlli non molto costosi.

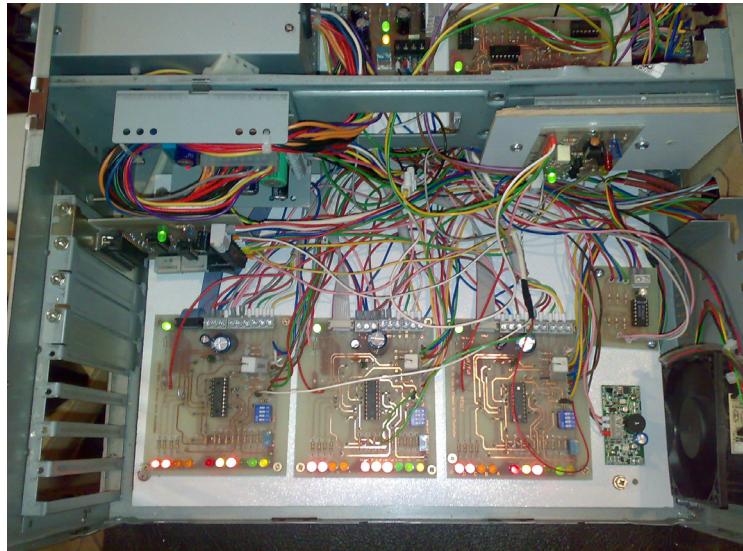
I motori presentano solitamente 5 o 6 fili e non hanno molta potenza.

- o Per motori bipolari che presentano normalmente 4,6 o 8 fili, mai 5.

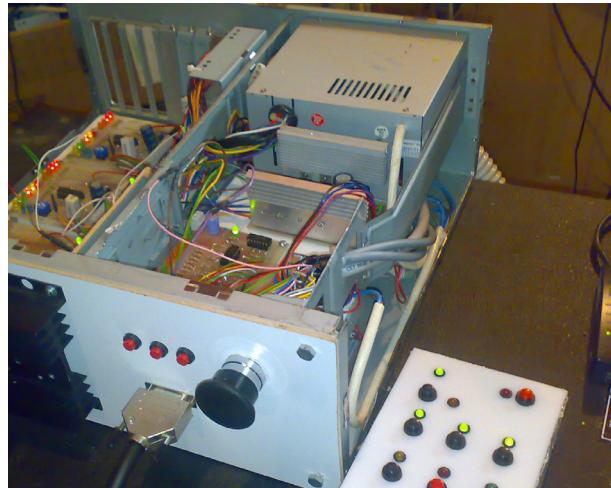
Sono più performanti ed i componenti elettronici che compongono il controllo sono sofisticati per il loro funzionamento.

Per i motori bipolari, la tensione di alimentazione normalmente va da 5 a 20 volte il valore di targa riportato sul motore.

- Solitamente all'interno del box di tale controllo, risiede pure la parte di alimentazione di potenza.
- Per il collegamento del controllo elettronico al computer, normalmente si usa ol connettore DB25 (quello della porta parallela), ma anche via USB o RS 232.



## GRUPPO DI ALIMENTAZIONE (POWER SUPPLY)



- Da questo gruppo viene prelevata l'alimentazione di potenza per i motori e per l'alimentazione delle schede.
- L'alimentazione di potenza, nel caso di drivers per motori unipolari, viene erogata da un alimentatore switching da computer che fornisce in uscita 5V con una corrente max di 20A. solitamente tale tensione viene raddrizzata da un ponte a diodi e livellata da un condensatore elettrolitico di circa 10.000mF.
- Per l'alimentazione delle schede normalmente si usa un secondo trasformatore toroidale che fornisce una tensione che viene poi stabilizzata a 5 Volt tramite un relativo alimentatore.

### Motori passo passo (step motor)

- Per il nostro uso solitamente si impiegano motori di provenienza "surplus", ossia smontati da macchine dimesse (vedi stampanti), o da resti di magazzino non più usati.

Si hanno quindi prezzi più accessibili rispetto all'acquisto di motori nuovi.

- Per uso bipolare si usano motori con 4, 6 o 8 fili.

- Per uso unipolare da 5 6 o 8 fili.

- I motori sono classificati anche per il numero di passi che compiono per fare un giro, su frese CNC solitamente si impiegano motori da 200 passi (1,8 gradi).

motori da 48 passi sono poco costosi ma non molto adatti all'uso su frese CNC .

## PROGRAMMI CAD, CAM ED INTERPRETI VARI.

- Con il CAD noi disegniamo la parte che poi viene lavorata. Il più conosciuto è Autocad, ma molti altri CAD, anche migliori, possono essere impiegati, l'estensione del file è solitamente .dxf
- Con il CAM noi trasformiamo il file del disegno da .dxf ad un file in G-code. Sarà questo file G-code che verrà letto dal Programma di controllo CNC che permetterà alla nostra fresa di eseguire le lavorazioni. Se non disponiamo di un programma CAM, possiamo utilizzare semplici programmi di conversione o interpreti che ci trasformano il file .dxf in G-code, il più conosciuto è ACE.
- Vi sono dei programmi di controllo fresa CNC che al loro interno hanno già un CAM, vedi DeskCNC, oppure posono convertire un file da .dxf in G-code, vedi Kcam4.

La fresa che esegue le lavorazioni

È la parte meccanica di tutto il sistema, su questa troviamo montati i motori passo passo e l'utensile fresa che eseguirà le lavorazioni.

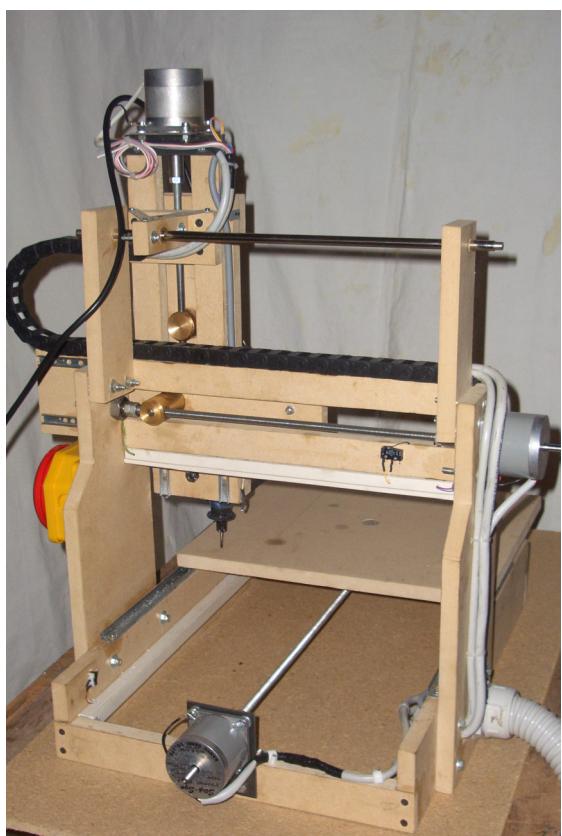
La sua struttura è in funzione delle dimensioni dei pezzi da lavorare e del tipo di materiale che andremo a lavorare.

Materiali duri richiederanno una struttura della fresa robusta e rigida.

Soltamente gli assi di lavorazione sono tre: X,Y e Z.

Il sistema di guida dei tre assi è molto importante che sia preciso e scorrevole

## CAPACITÀ DELLE MACCHINE CNC



La maggior parte dei centri di lavoro controllati da computer sono dotati di movimento verticale del mandrino in grado di realizzare incisioni, sculture e lavorazioni di grandissima precisione. Lavorazioni realizzate su queste macchine possono anche arrivare ad essere indistinguibili ad un occhio non espertissimo rispetto a oggetti d'arte realizzati a mano. Se utilizzate con frese coniche o sferiche possono arrivare a creare superfici assolutamente lisce, di altissima precisione, in modo rapido, automatizzato e ad un costo estremamente contenuto. Le migliori macchine CNC arrivano a una precisione di un decimillesimo di millimetro (100 nanometri).

Le macchine CNC più avanzate sono dotate di testate orientabili (assi tilting) in grado di ruotare giroscopicamente lungo due assi (normalmente chiamati A, B, C o Q). Questo consente di inclinare l'utensile rispetto a tutti i piani di lavoro rendendo possibile realizzare figure molto complesse anche con forme di tipo organico e difficilmente ottenibili persino da parte di una lavorazione manuale (come ad esempio una cavità con un singolo foro molto stretto ma sufficiente per far passare la fresa). Le teste tilting permettono anche di ottimizzare la lavorazione avendo la possibilità di lavorare con l'utensile inclinato rispetto al piano d'avanzamento evitando di utilizzare la parte dell'utensile a velocità zero (centro di rotazione della fresa). Con i controlli numerici più recenti

iniziano ad implementare funzioni di lavorazione avanzata delle superfici, permettendo di lavorare superfici 3D NURBS in modo nativo, senza bisogno di programmi CAM intermedi. Lo sfruttamento ottimale delle caratteristiche di queste macchine può avvenire o attraverso appositi accessori per copiatura e stampi, oppure via computer, attraverso una catena di programmi: prima si crea con il CAD un oggetto, che poi viene passato al CAM che si incarica di creare il programma per la realizzazione su una data macchina, e infine il programma

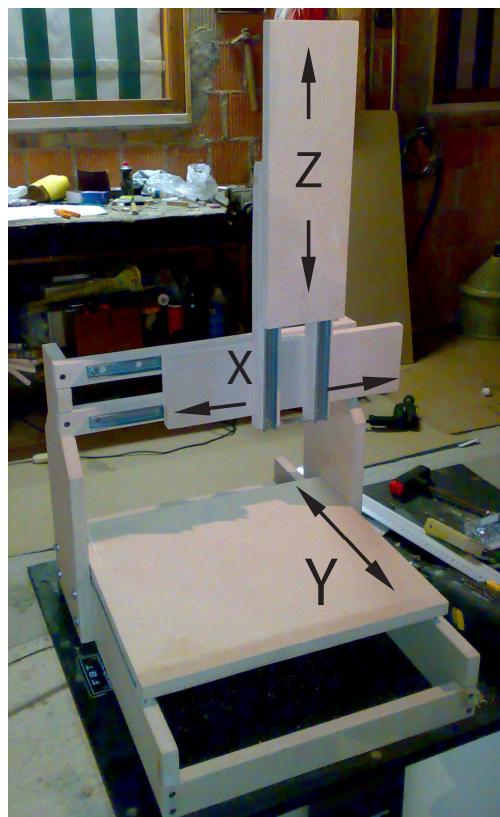


viene eseguito dalla macchina creando l'oggetto materiale. Praticamente tutte le macchine CNC moderne sono "a circuito chiuso" (a catena chiusa o retroazionato): ogni asse è dotato di encoder (encoder rotativi oppure righe ottiche) che permettono al controllo di conoscere in ogni istante la posizione della testa di lavorazione rispetto al pezzo. L'uso di righe ottiche permette il recupero completo dei giochi della trasmissione misurando direttamente e non attraverso i cinematismi la posizione dell'asse e quindi una miglior precisione di lavoro.

## ASSI

Caratteristica principale di queste macchine è il numero di gradi di libertà disponibili, detti assi della macchina. Per le frese sono generalmente 3, 4 o 5, per i torni si va da 2 a 4, le punzonatrici e le macchine per taglio hanno generalmente 2 o 3, mentre le pannellatrici (FINN-POWER, Salvagnini...) arrivano a 24 assi. Le piegatrici possono avere fino a 11 assi e sono le macchine "singole" più complesse; le macchine speciali combinate e i grandi portali di lavorazione possono arrivare ad avere anche decine di gradi di libertà o assi diversi. I tipi più comuni sono

- 2 assi: movimento solo su X e Y. Pezzi limitati a lavorazioni piane senza variazioni di profondità.
- 2.5 assi (due assi e mezzo): si tratta di macchine in grado di operare su tutti e tre gli assi, ma soltanto a passi discreti su uno di essi (generalmente l'asse Z), cioè gestiscono l'interpolazione soltanto su due assi. Sono oggi il tipo più economico.
- 4 e 5 assi: oltre al movimento sui tre assi X, Y e Z si aggiunge l'inclinazione (ed eventualmente la rotazione) del mandrino o di una tavola rotobasculante solidale al piano macchina. Queste macchine possono realizzare praticamente qualsiasi tipo di sagoma purché la forma delle concavità non vada in interferenza con il mandrino. Queste macchine sono molto costose, e vengono usate anche per lavorazioni su pezzi di grandissime dimensioni (turbine, eliche navali, scocche per motori, stampi...)

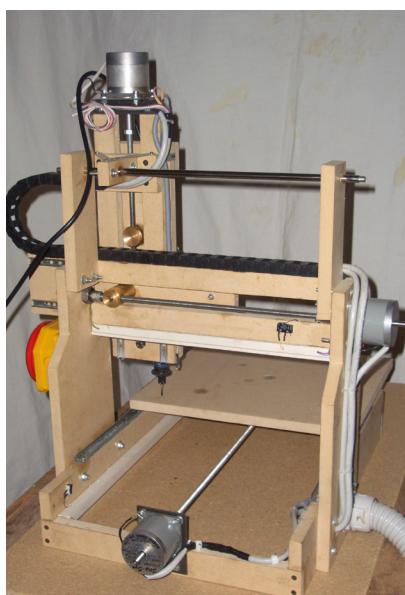


Nell'industria dei grandi trasporti si usano anche macchine dette "a 8 assi" o "a 10 assi": si tratta di centri di lavoro a due testate indipendenti, in grado di velocizzare le operazioni di lavorazione su pezzi di grandissime dimensioni come gli estrusi di alluminio e acciaio per le casse dei treni o le lamiere degli aerei.

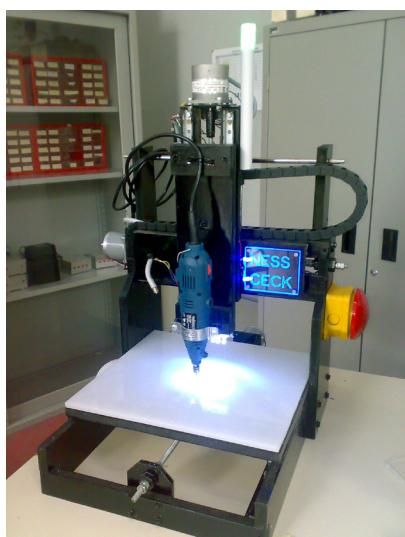
## ⇒IL NOSTRO PROGETTO - CNC 3 ASSI



PRIMA REALIZZAZIONE DELLA STRUTTURA



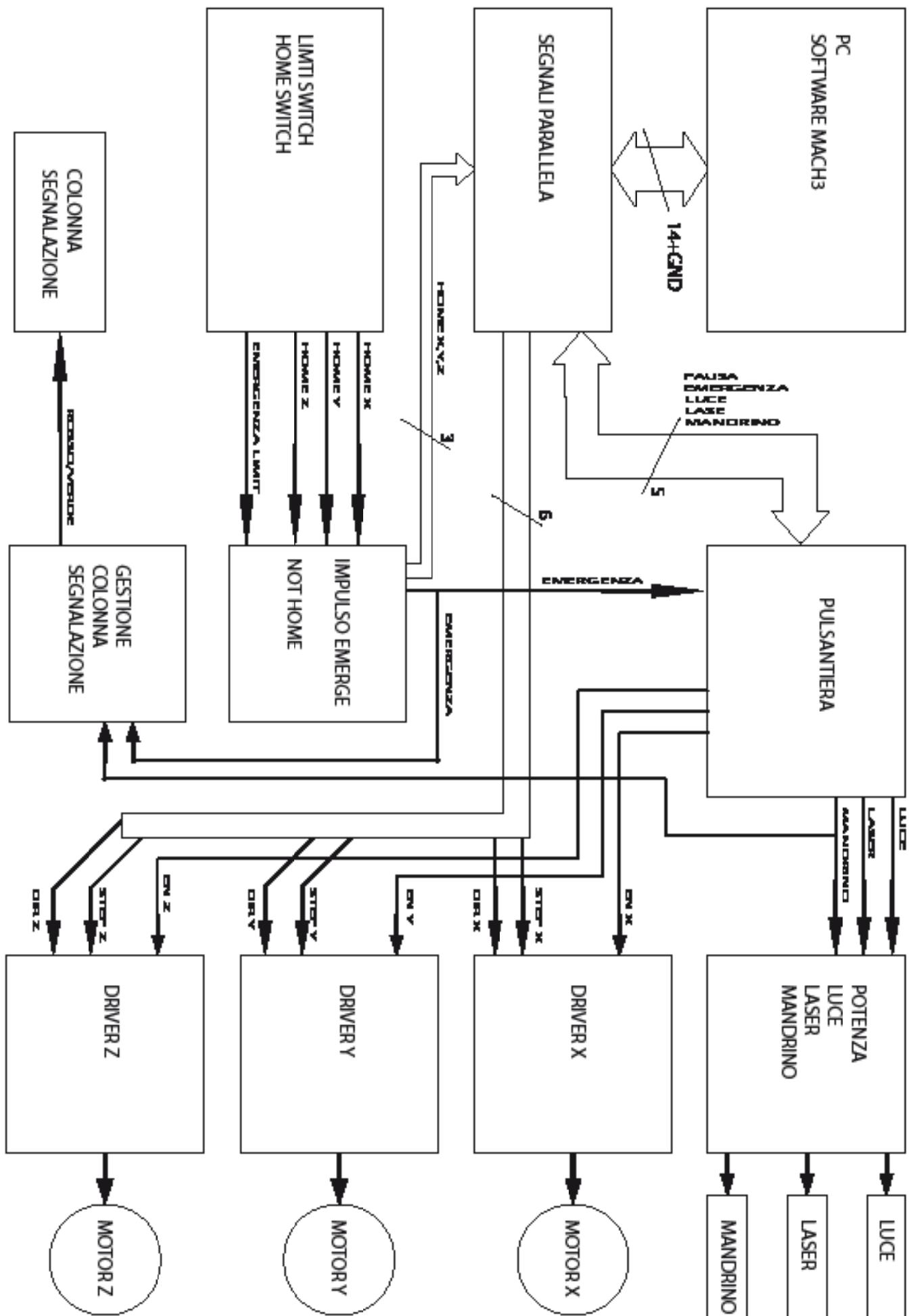
BOX DI CONTROLLO ULTIMATO



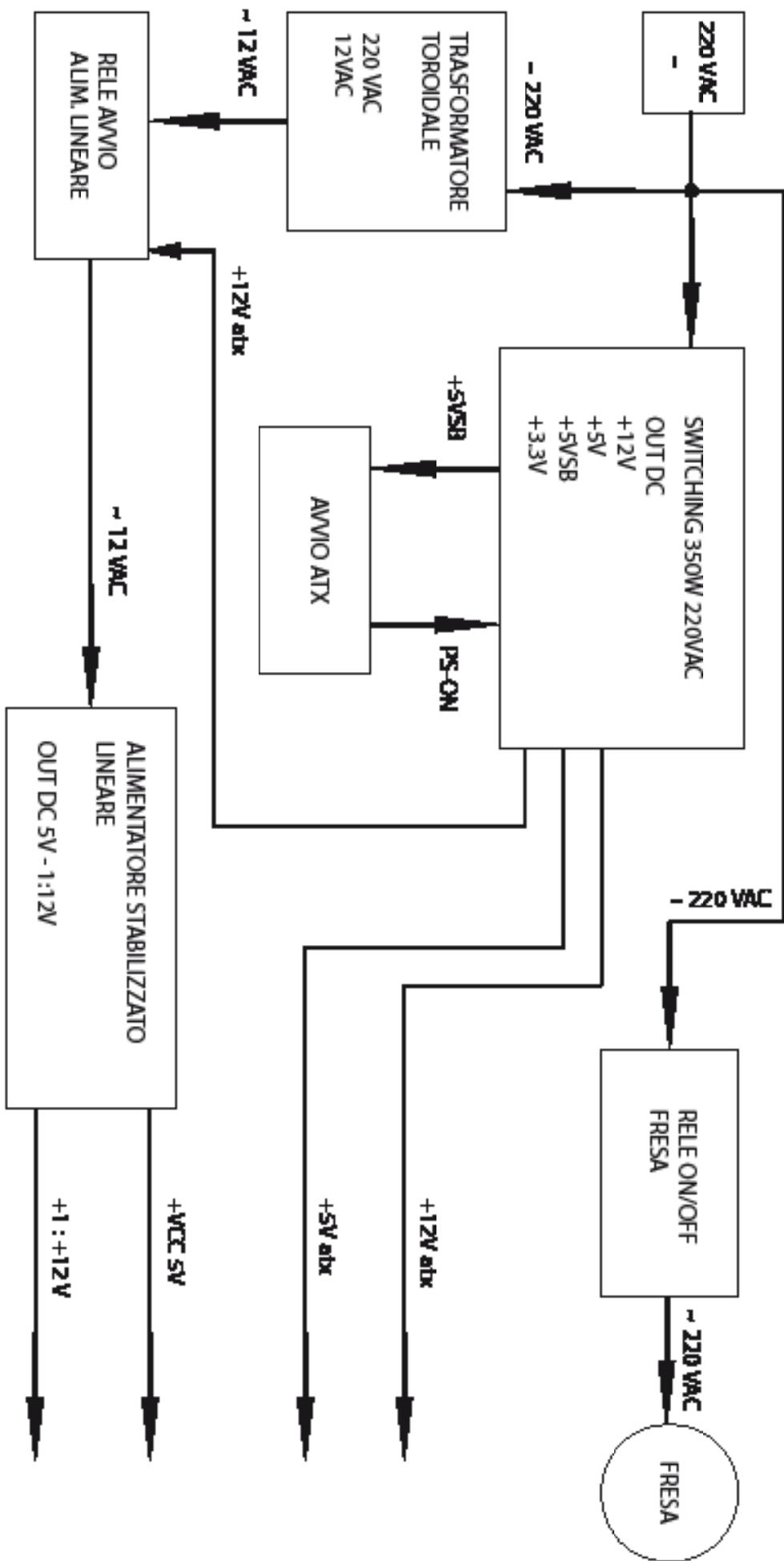
MACCHINA ULTIMATA CON TUTTE LE RIFINITURE



→ SCHEMA A BLOCCHI DEI COLLEGAMENTI TRA LE VARIE SCHEDE



↑ SCHEMA A BLOCCHI ALIMENTAZIONI



## ⇒ ACCENSIONE SPEGNIMENTO ALIMENTATORE ATX

### ► DESCRIZIONE GENERALE

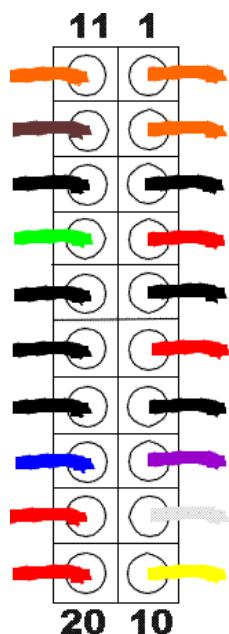
Il circuito ha il compito di avviare l'alimentatore da pc (ATX).

In modalità stanby l'alimentatore fornisce una sola tensione di 5V sul filo viola(+5VSB), questa tensione va ad alimentare il flip-flop jk e il pulsante. Il flip-flop in configurazione toogle(J,K,preset e clear collegati a 5v), cambierà di stato la sua uscita ogni volta che viene premuto il pulsante o la chiave di avvio collegata come nello schema sul piedino di clock del FF.

quando viene attivata l'uscita verrà attivato anche un relè che collegherà a massa il filo verde(PS-ON) il quale causerà l'avvio dell'alimentatore.

La presenza dell'alimentazione a 5V in modalità stand-by è segnalata dall'accensione di un led giallo, la presenza della tensione a 5V ad alimentatore avviato è segnalata da un led verde. Un led arancio invece segnalerà il corretto avviamento dell'alimentatore e che le varie tensioni d'uscita(3.3V, +5V, +12V, -5V, -12V) sono presenti in maniera corretta(non sono presenti corto circuiti o altre cause di uno scorretto mal funzionamento).

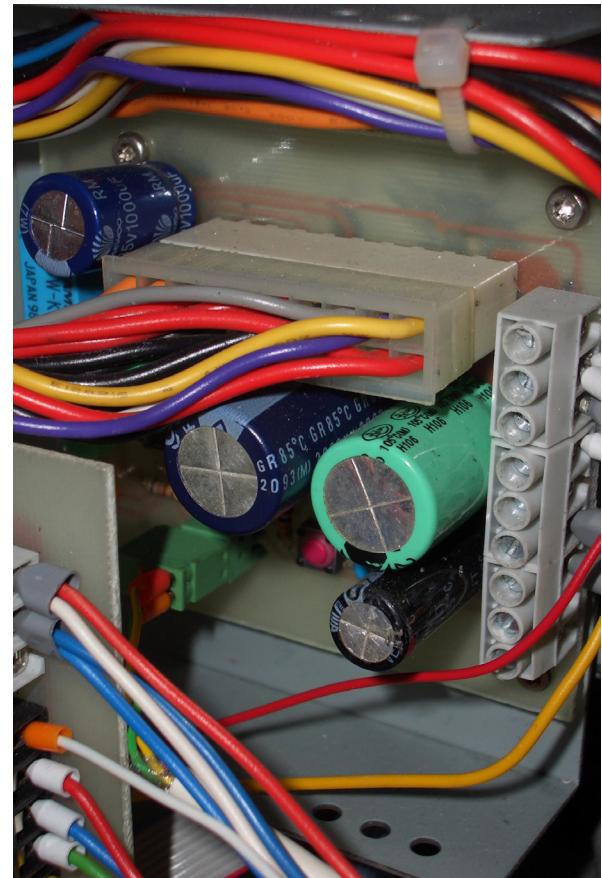
In caso contrario se è presente qualche anomalia il led arancio sarà spento.



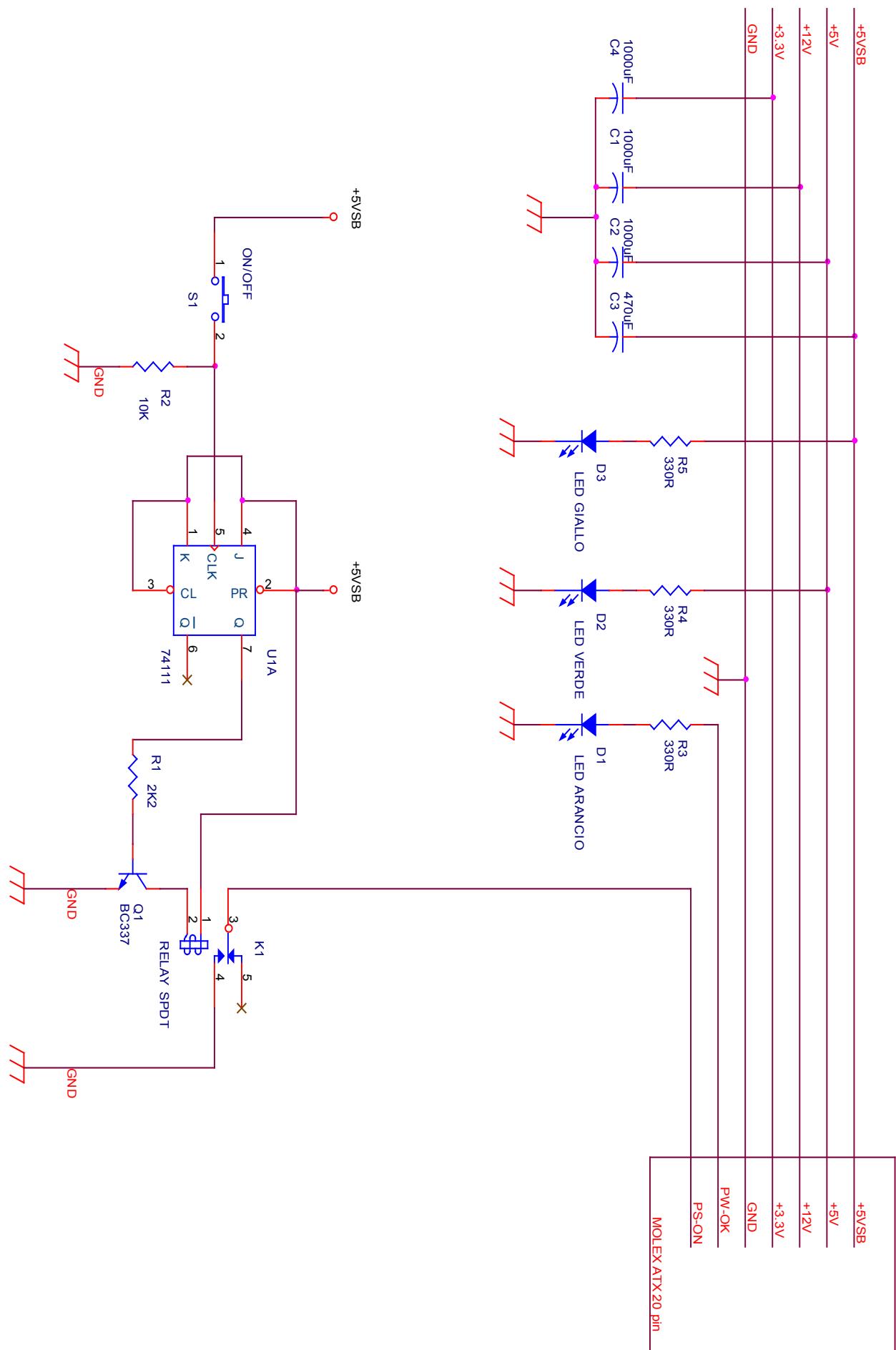
Connettore  
alimentazione  
ATX

11	1	3.3	3.3
		-12	3.3
		GND	GND
		ON	+5
		GND	GND
		GND	+5
		GND	GND
		-5	OK
		+5	VSB
		+5	+12

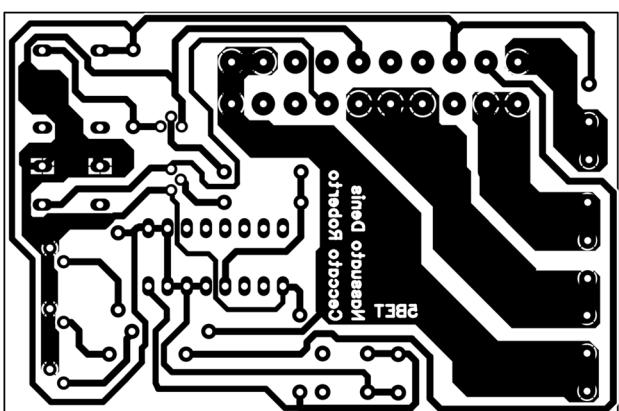
PIEDINATURA DEL CONNETTORE  
MOLEX ATX A 20 PIN.



## ► SCHEMA ELETTRICO

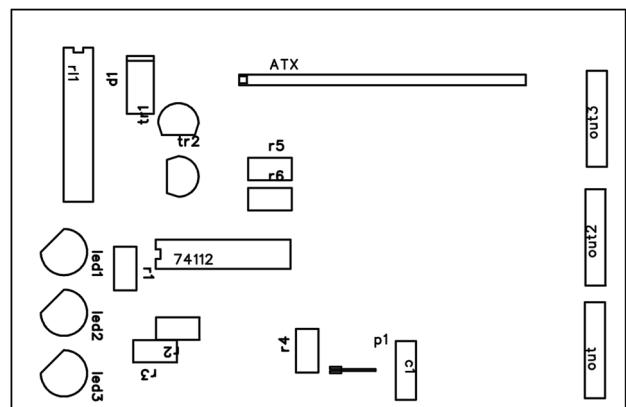


## ► PCB



1. BOTTOM COPPER - Lato rame

2. SERIGRAFIA - Posizione componenti



## ⇒ ALIMENTATORE STABILIZZATO 5VDC - 1:12VDC

### ► DESCRIZIONE GENERALE

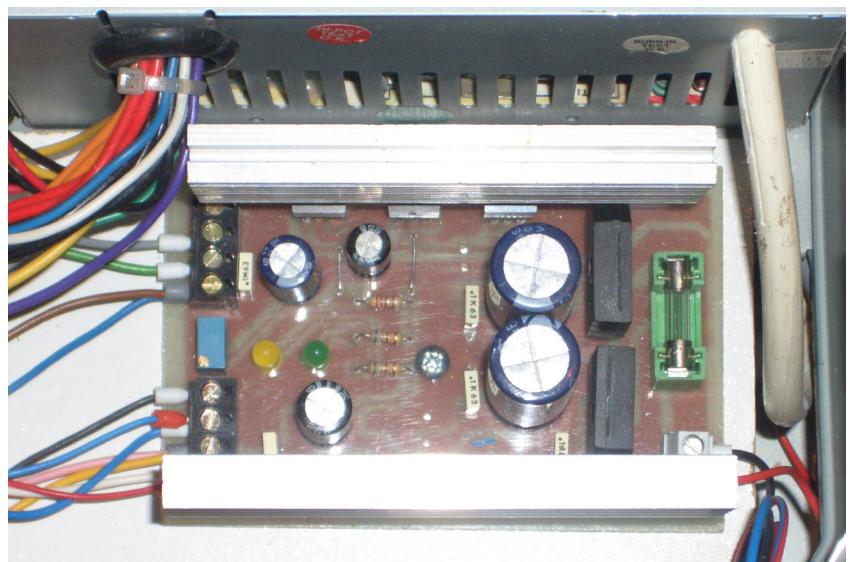
Si tratta di una tecnologia estremamente semplice ed economica, largamente usata ove la potenza richiesta sia limitata ed il costo rappresenti un limite.

Un generico alimentatore lineare è idealmente (e spesso anche praticamente) composto dai seguenti elementi collegati in cascata:

- Un trasformatore: provvede a ridurre (o in rari casi aumentare) la tensione proveniente dalla rete elettrica per avvicinarla al valore richiesto dal carico da servire.
- Un raddrizzatore: trasforma la corrente alternata fornita dal trasformatore in corrente continua. Può essere a diodo singolo o a ponte.
- Un filtro livellatore: livella la corrente unidirezionale pulsante uscente dal raddrizzatore in una corrente più uniforme e costante. Solitamente rappresentato semplicemente da un condensatore.
- Un circuito elettronico Stabilizzatore detto anche regolatore, che può spaziare da un semplice diodo zener ad un circuito integrato dedicato. Assicura che la tensione generata dall'alimentatore si mantenga costante nel tempo ed entro una stretta tolleranza rispetto al valore richiesto, al variare della tensione della rete elettrica e del carico applicato. Nel nostro caso per la tensione fissa di 5v come stabilizzatore è stato utilizzato il circuito integrato 7805 su contenitore TO-220, mentre per la tensione variabile da 1:12v il regolatore è l'integrato LM317 sempre in contenitore TO-220.

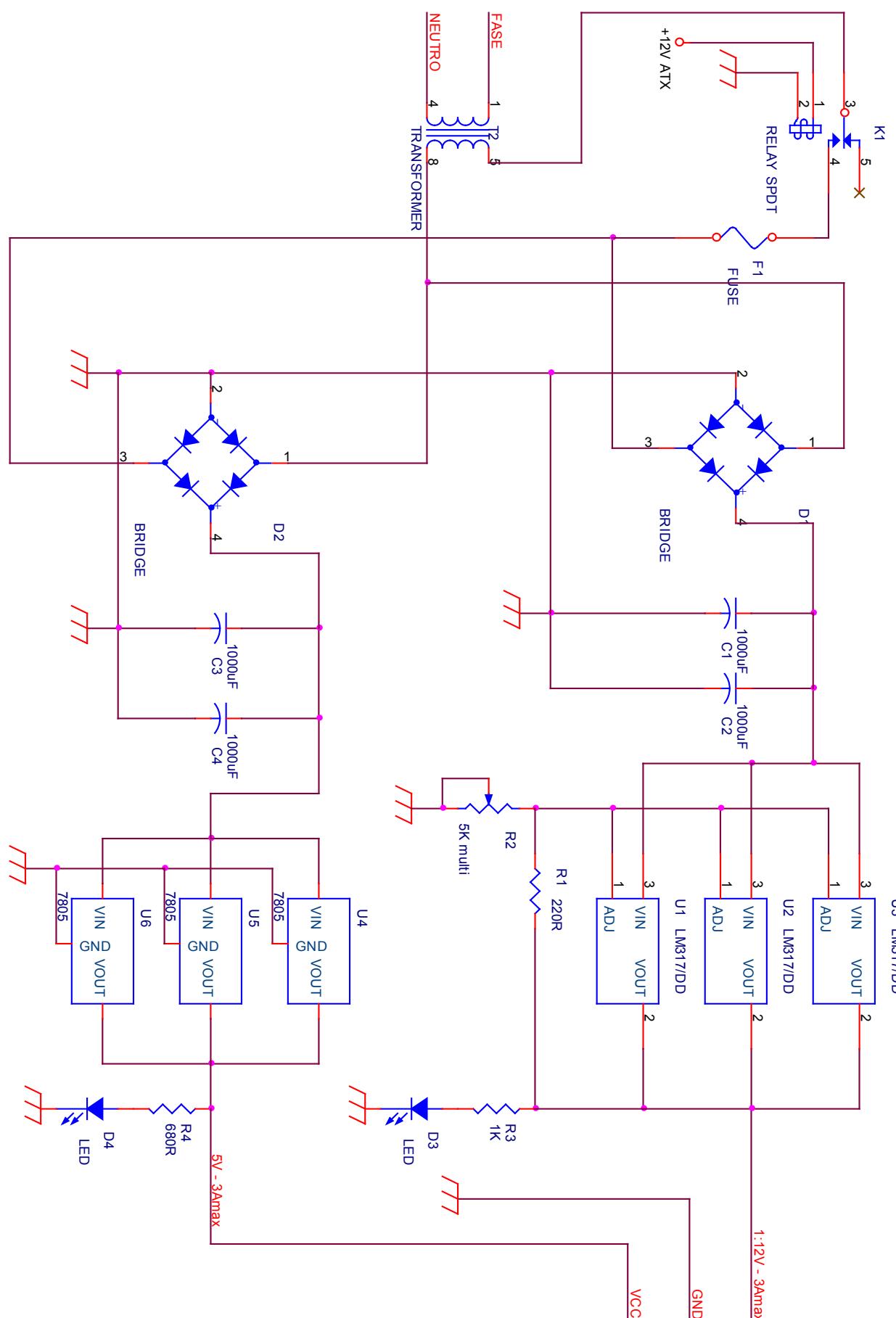
I

principali limiti di questi alimentatori risiedono nel basso rendimento energetico, che comporta, nel caso di elevate potenze gestite, un consistente sviluppo di calore, che deve essere smaltito per evitare danni all'apparato.



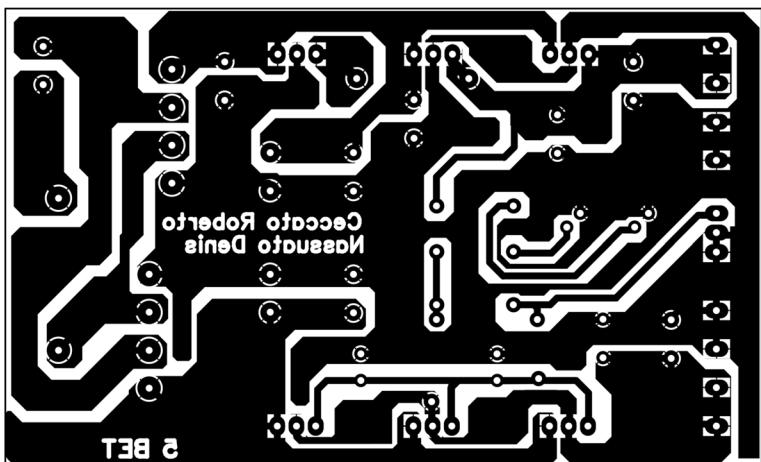
Un altro limite è nell'eccessivo incremento di dimensioni e peso all'aumentare della potenza di un alimentatore analogico.

## ► SCHEMA ELETTRICO

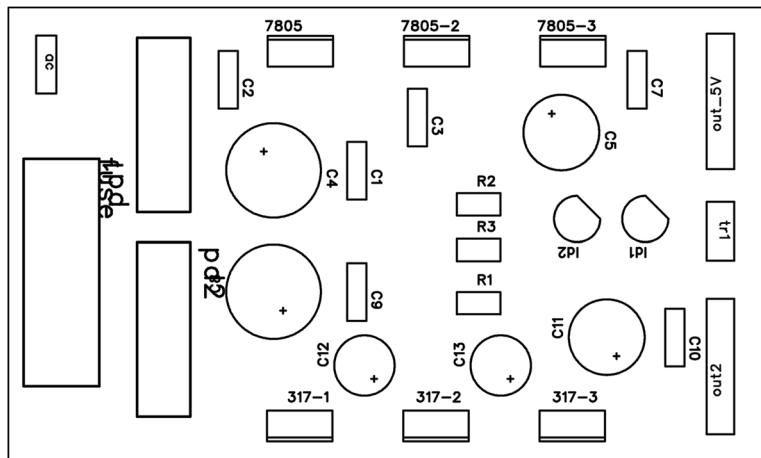
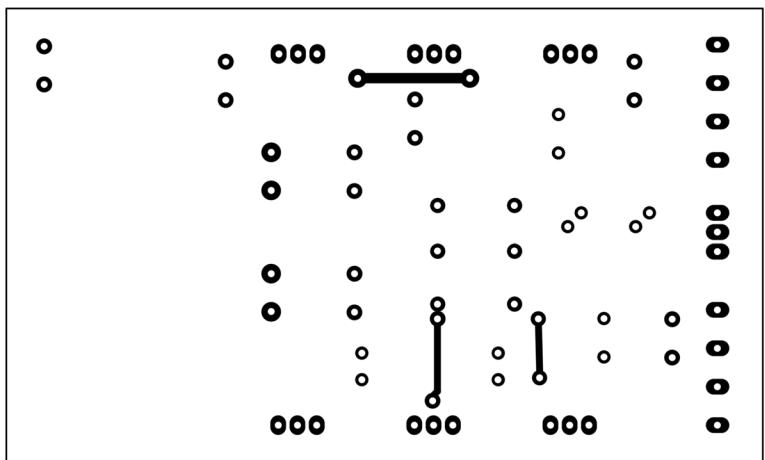


## ► PCB

1. TOP COPPER - Lato componenti



2. BOTTOM COPPER - Lato rame



3. SERIGRAFIA - Posizione componenti

## ⇒ PORTA PARALLELA (LPT)

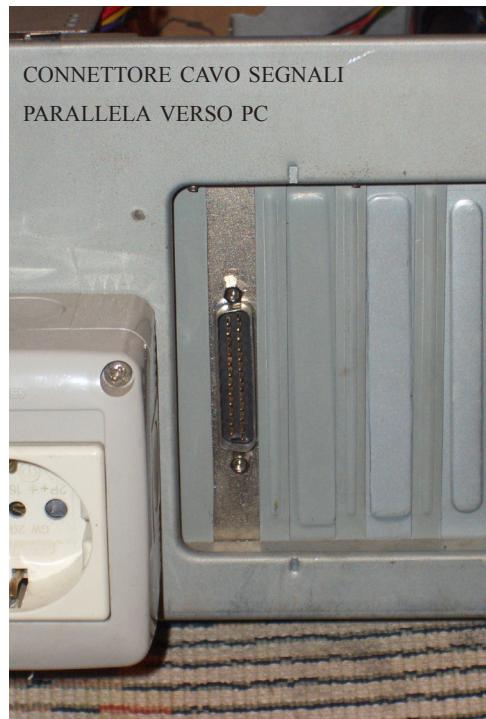
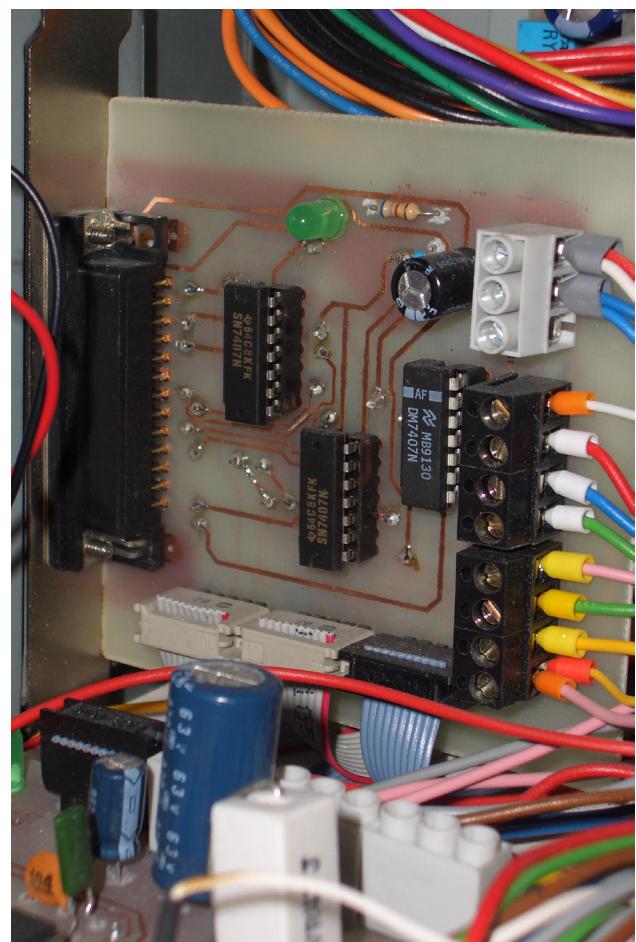
### ► DESCRIZIONE GENERALE

Il circuito disaccoppia i segnali provenienti da e verso il pc tramite la porta parallela.

Il disaccoppiamento avviene attraverso degli Hex buffer (7407) i quali portano in uscita lo stesso livello (1 o 0) che si presenta al suo ingresso.

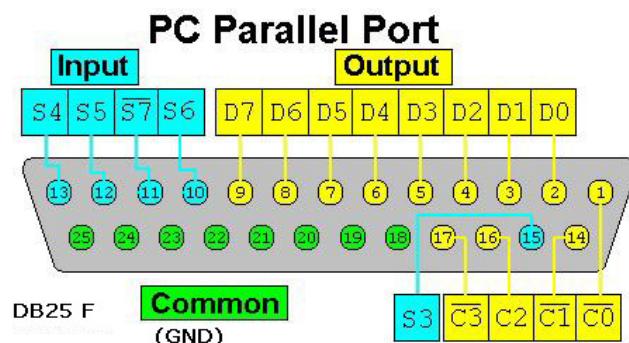
I segnali verso la macchina e provenienti dal pc sono:

- STEP ASSE X
- STEP ASSE Y
- STEP ASSE Z
- DIREZIONE ASSE X
- DIREZIONE ASSE Y
- DIREZIONE ASSE Z
- CONTROLLO ON/OFF MANDRINO
- CONTROLLO ON/OFF LASER
- CONTROLLO ON/OFF LUCE LAV.



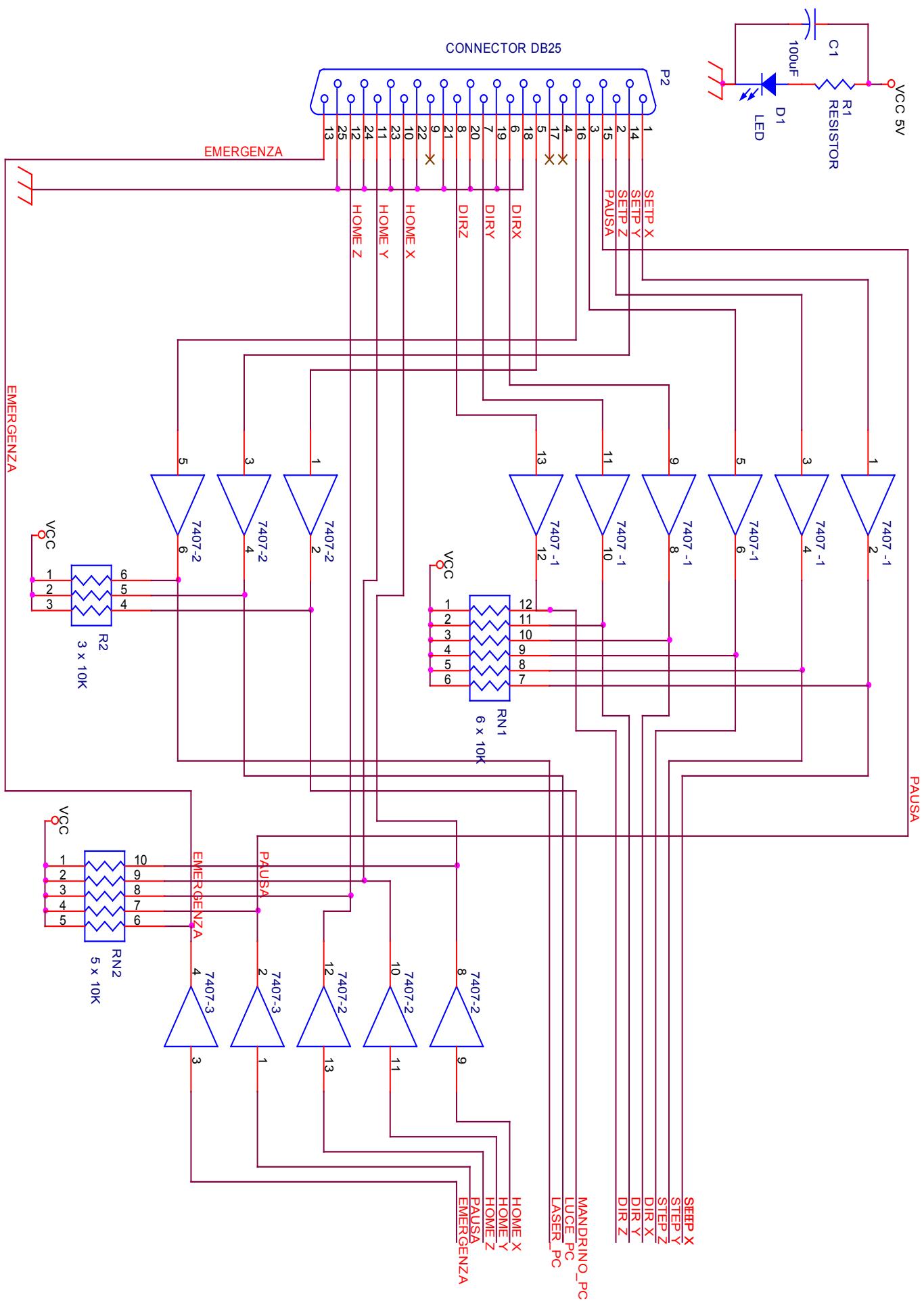
Mentre i segnali provenienti dalla macchina e diretti verso il pc sono:

- EMERGENZA GENERALE
- PAUSA LAVORAZIONE
- HOME ASSE X
- HOME ASSE Y
- HOME ASSE Z

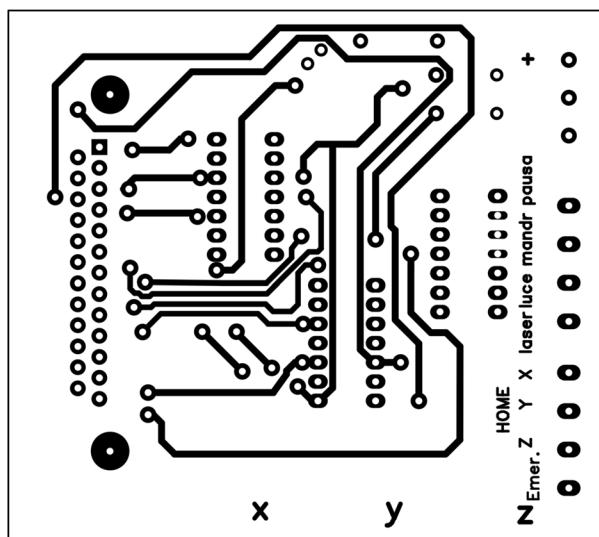


D0 ⇒ STEP X	D6 ⇒ DIR Z	S6 ⇒ HOME Y
D1 ⇒ STEP Y	D7 ⇒ NC	S7 ⇒ HOME X
D2 ⇒ STEP X	C0 ⇒ NC	C1 ⇒ LUCE
D3 ⇒ MANDRINO	S3 ⇒ PAUSA	C2 ⇒ LASER
D4 ⇒ DIR X	S4 ⇒ EMERGEN.	C3 ⇒ NC
D5 ⇒ DIR Y	S5 ⇒ HOME Z	

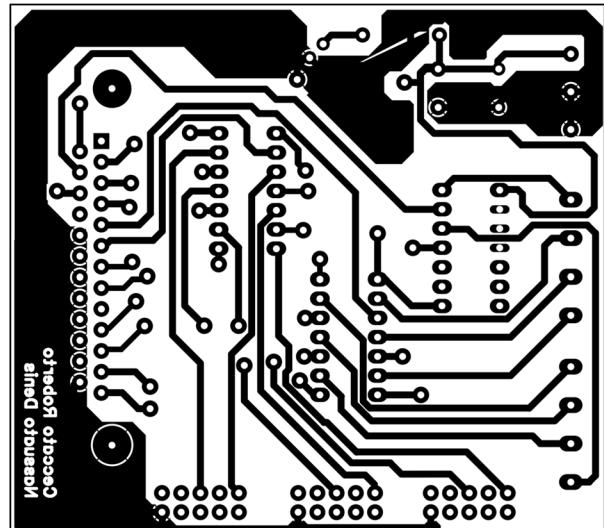
# ► SCHEMA ELETTRICO



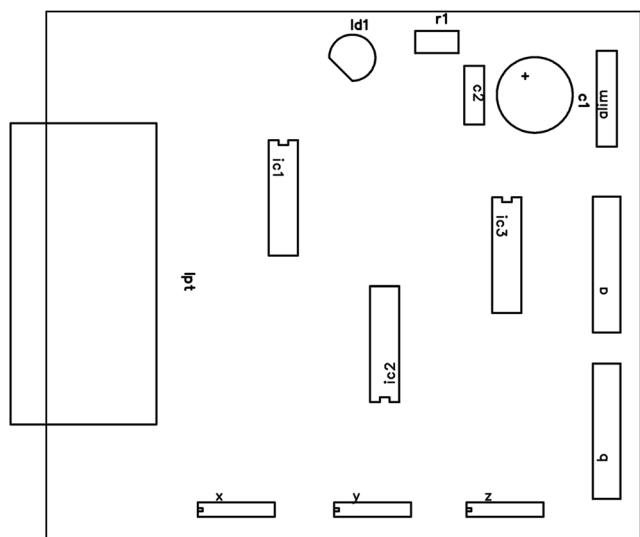
## ► PCB



1. TOP COPPER - Lato componenti



2. BOTTOM COPPER - Lato rame



3. SERIGRAFIA - Posizione componenti

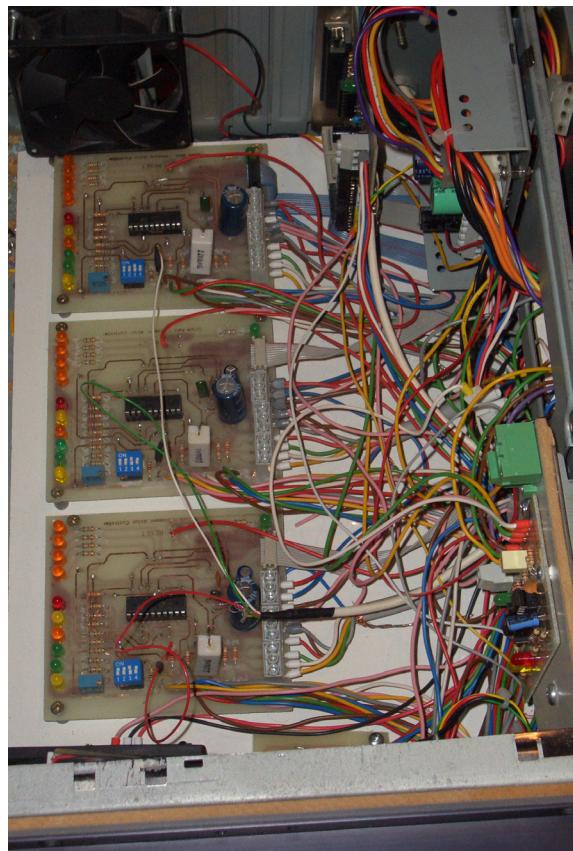
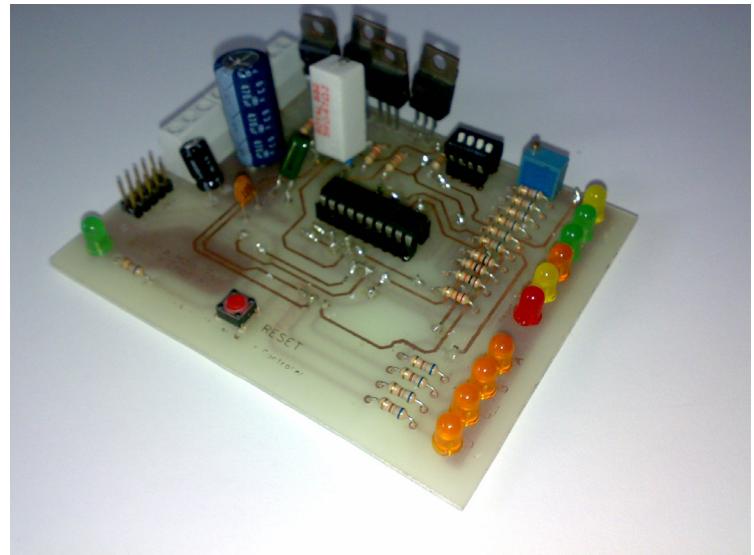
## ⇒ DRIVER MOTORE STEPPER

### ► DESCRIZIONE GENERALE

Il circuito è in grado di gestire e pilotare un singolo motore passo passo (stepper) del tipo unipolare. Il cuore di questo driver sta nell'integrato L297, il quale in base ad alcuni segnali di controllo in ingresso quali:

- STEP: impulso di passo;
- CW/CCW: segnale di direzione della rotazione;
- HALF/FULL: tipo di passo, mezzo passo o passo intero;
- ENABLE: abilitazione del driver.
- RESET: configurazione iniziale delle fasi del motore.

è in grado di gestire direttamente le quattro basi dei quattro transistor darlington, i quali a loro volta pilotano ciascuna delle quattro fasi del motore.



### ► FUNZIONAMENTO

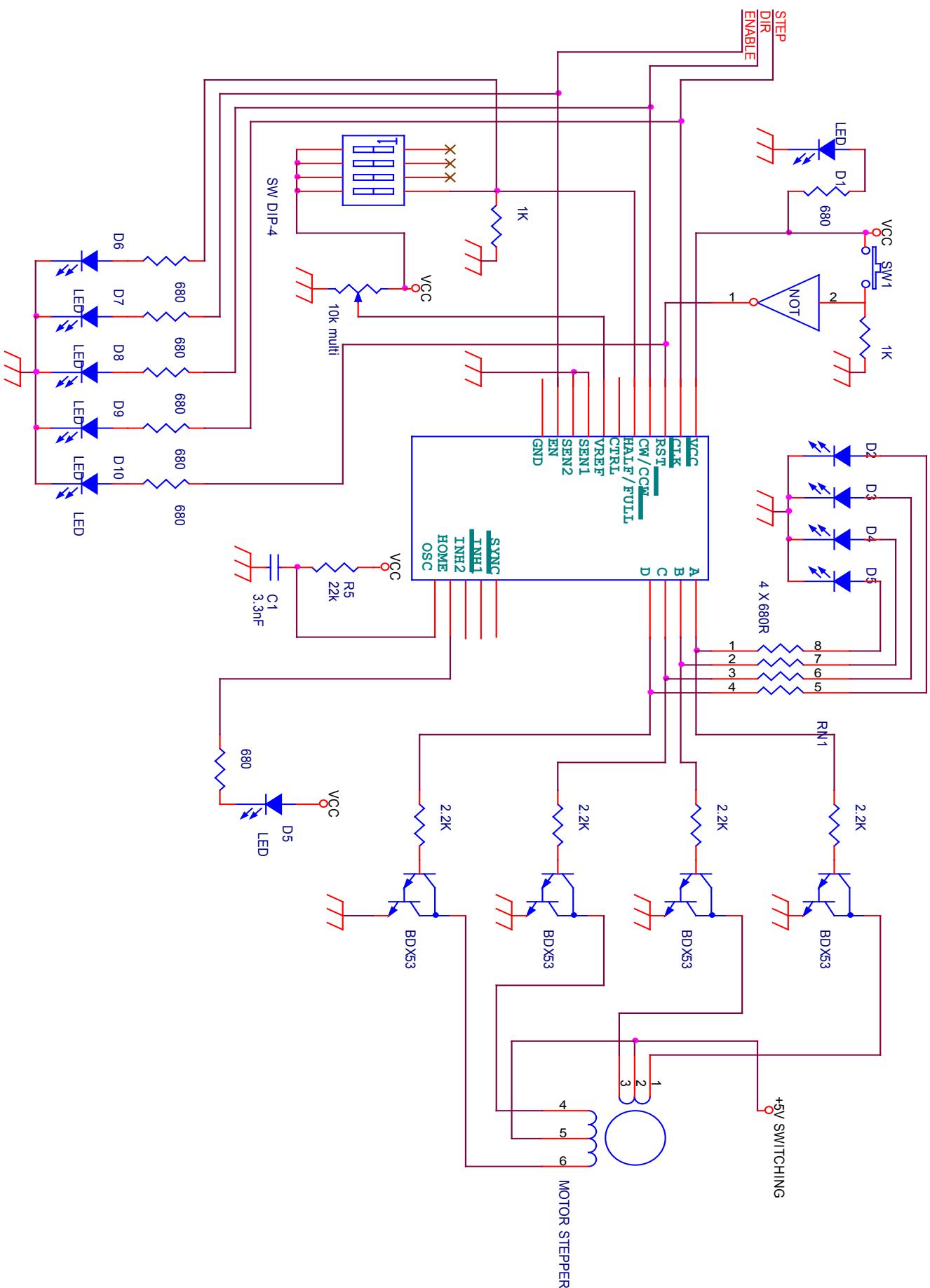
Per attivare la scheda è necessario abilitare il segnale di Enable del relativo asse tramite la pulsantiera di comando, a questo punto il driver manterrà attive le prime due fasi(A,B) del motore, se all'ingresso STEP si presenta un treno di impulsi(in questo caso proveniente dalla porta parallela del PC), l'integrato inizierà la codifica e il pilotaggio delle varie fasi del motore per farlo ruotare in base anche allo stato del segnale di direzione.

La velocità di direzione è proporzionale alla frequenza degli impulsi si step.

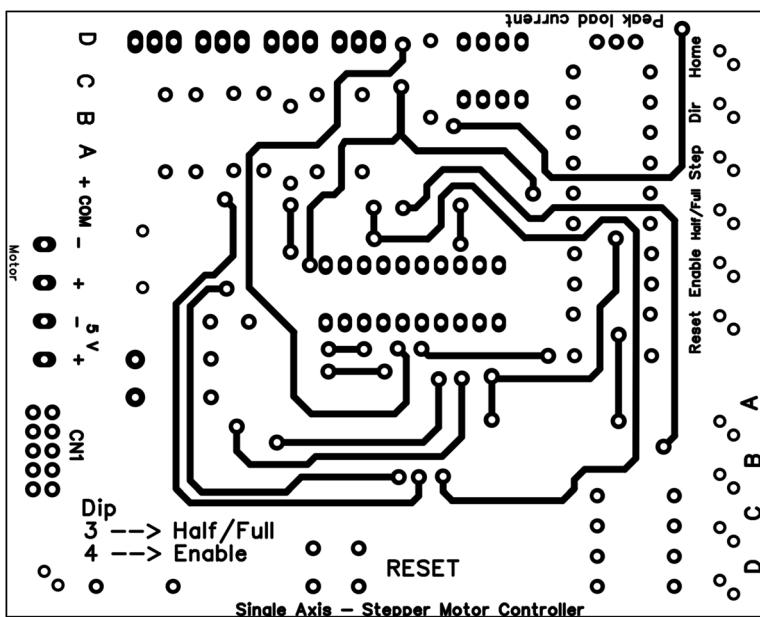
Lo stato di tutti i segnali sia di controllo che delle fasi è visualizzato tramite il relativo led presente sulla scheda.

PER IL COMANDO SONO PRESENTI TRE DRIVER DI QUESTO TIPO, UNO PER CIASCUN MOTORE DEL RELATIVO ASSE (X,Y,Z). TUTTI I SEGNALI DI ABILITAZIONE SONO GESTITI DALLA PULSANTIERA DI COMANDO, I SEGNALI DI STEP E DIREZIONE DAL PC, MENTRE IL TIPO DI PASSO È IMPOSTATO PER TUTTE E TRE LE SCHEDE IN MODALITÀ MEZZO PASSO TRAMITE IL DIP-SWITCH PRESENTE SULLA SCHEDA

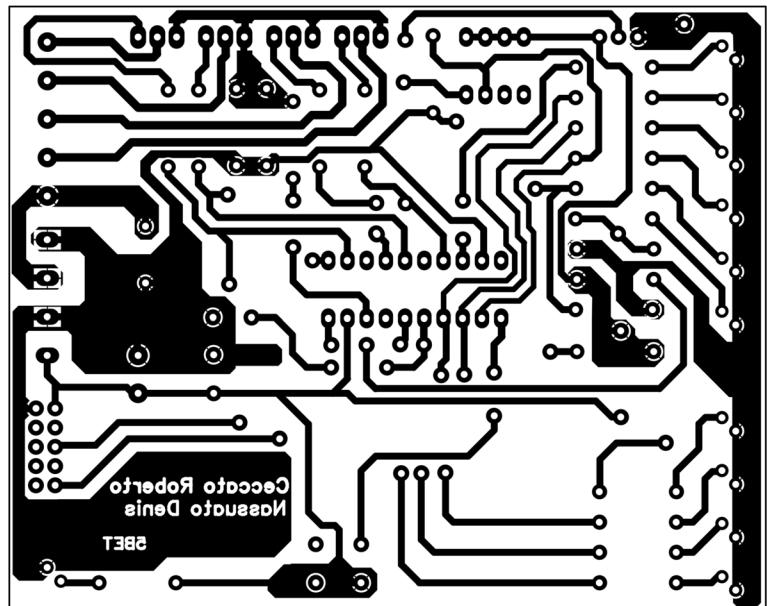
## ► SCHEMA ELETTRICO



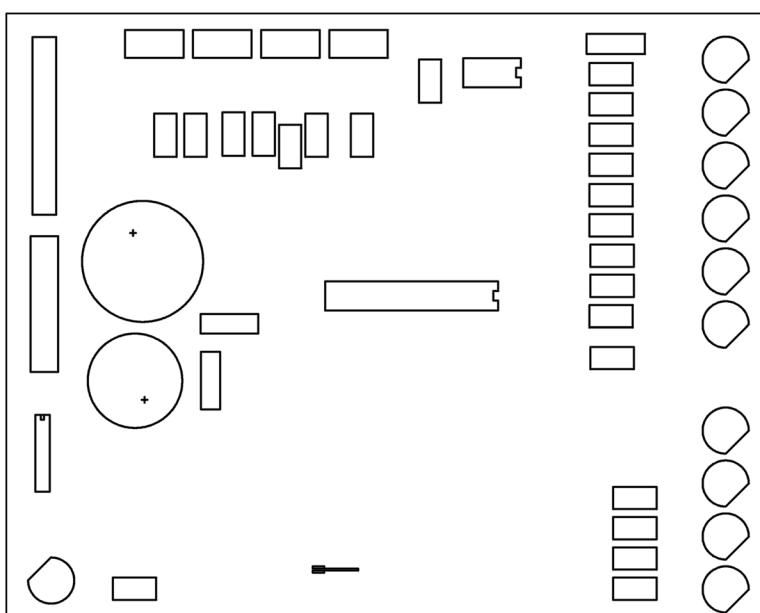
## ► PCB



1. TOP COPPER - Lato componenti



2. BOTTOM COPPER - Lato rame

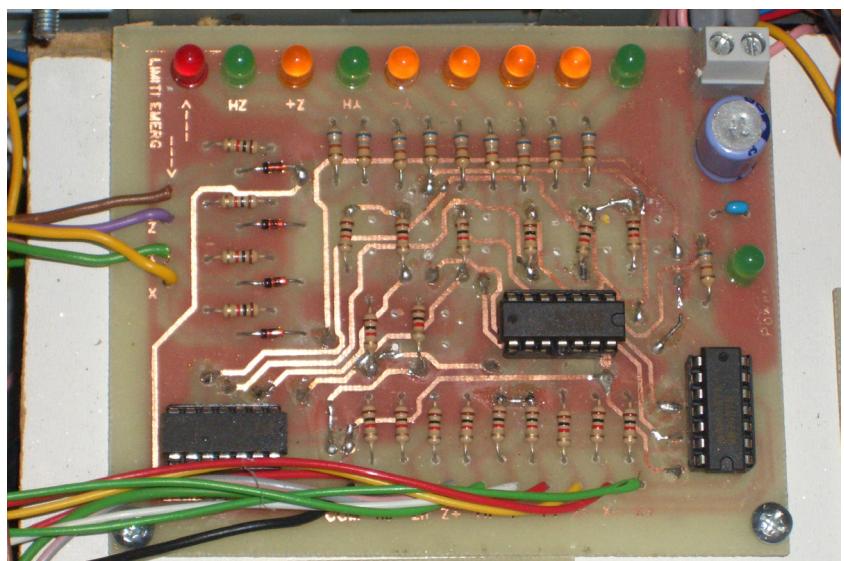


3. SERIGRAFIA - Posizione componenti

## ⇒ CONTROLLO LIMIT SWITCH E HOME SWITCH

### ► DESCRIZIONE GENERALE

Il circuito disaccoppia il segnale di ciascun finecorsa o home switch posizionati su ogni asse, in modo da eliminare i possibili disturbi provenienti dai cavi dei motori, i quali viaggiano paralleli a quelli dei finecorsa fino alla macchina.



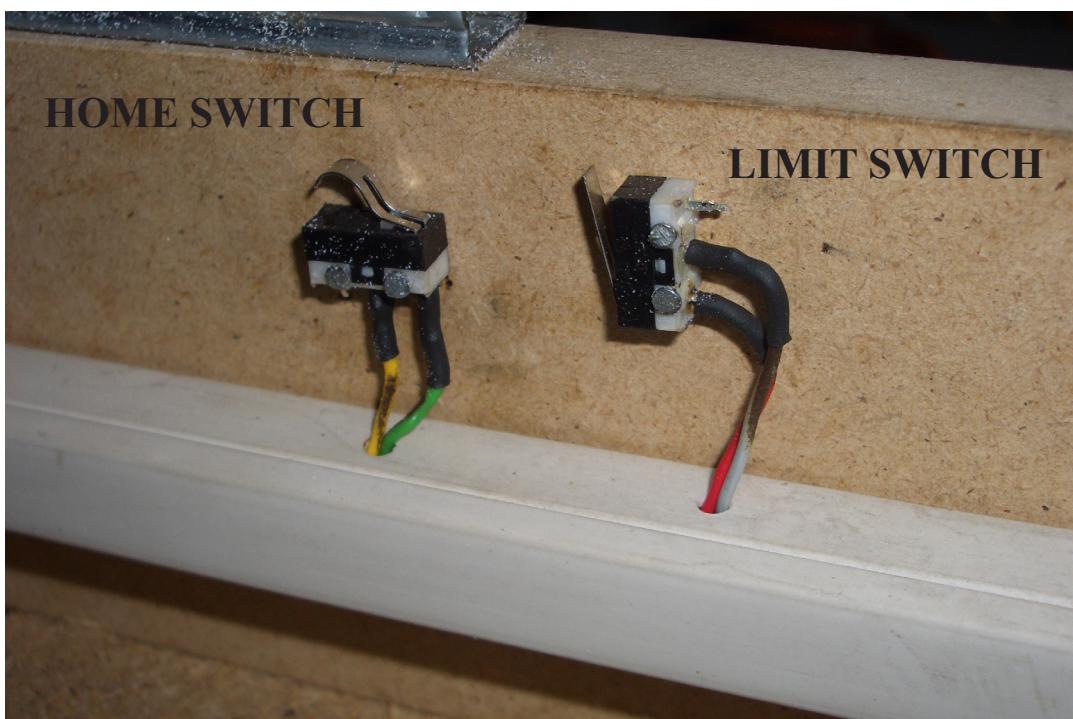
### ► FUNZIONAMENTO

I segnali disaccoppiati degli home switch vengono passati direttamente al PC, dove il software usa i medesimi segnali per referenziare la macchina a coordinate 0,0,0.

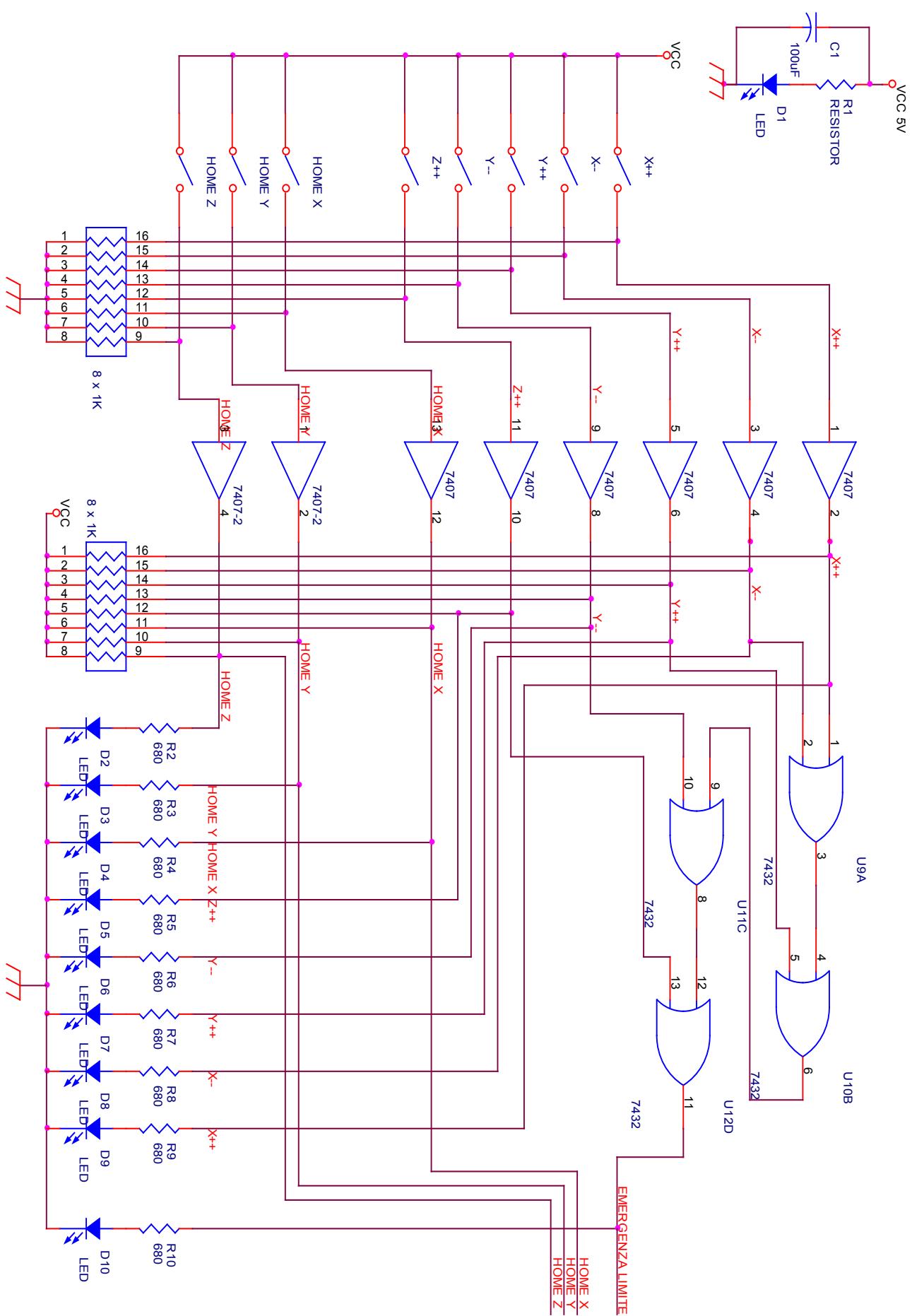
Invece i segnali provenienti dai finecorsa di limite, che sono due per l'asse X, due per l'asse Y e solamente uno per l'asse Z, vengono mandati tutti in OR logico tramite una rete logica(a porte OR), in modo che quando qualsiasi finecorsa di limite viene attivato venga a sua volta attivato un segnale unico di emergenza.

Ad ogni limit switch è collegato un led arancio, per ogni home switch è collegato un led verde mentre per l'uscita della rete a porte or, cioè l'emergenza generale relativa ai limit switch è collegato un led rosso.

Quindi ad ogni attivazione di qualsiasi switch sulla macchina comporterà l'accensione del relativo led di stato.

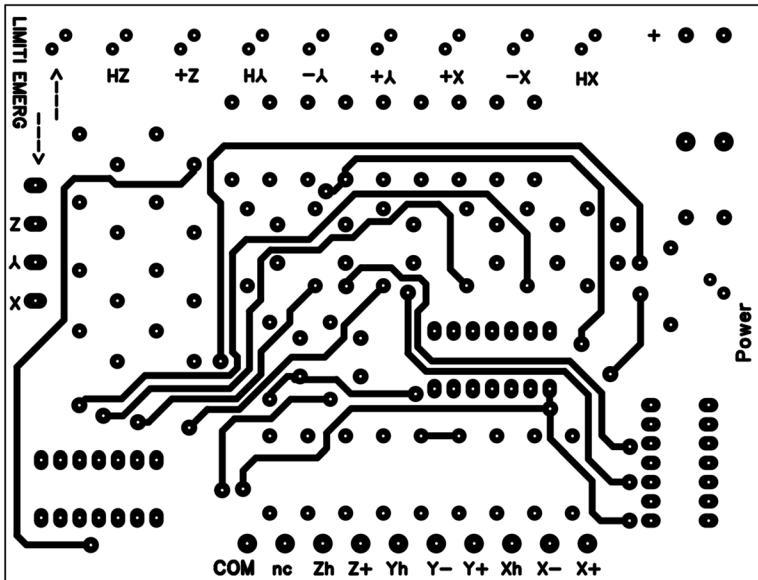


## ► SCHEMA ELETTRICO

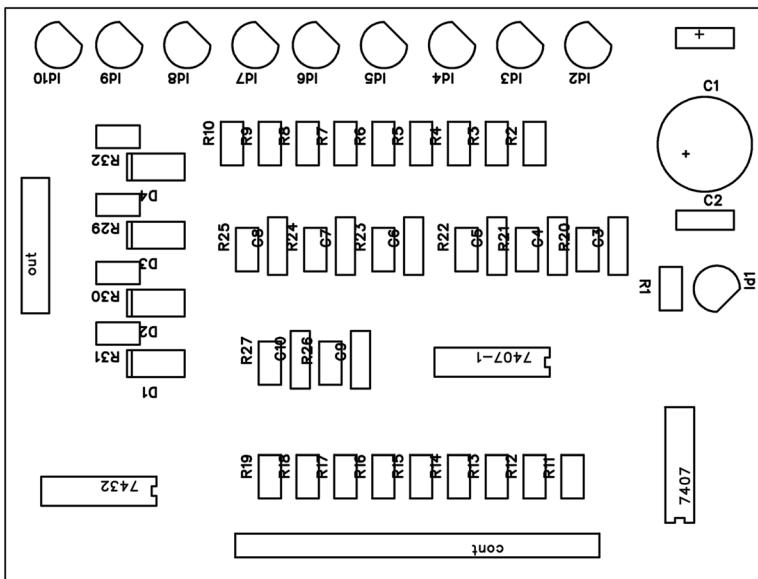
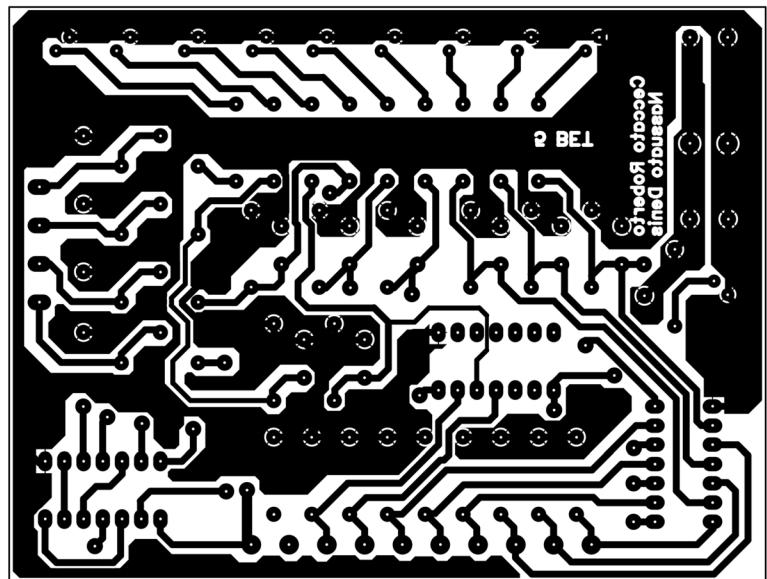


# ► PCB

1. TOP COPPER - Lato componenti



2. BOTTOM COPPER - Lato rame



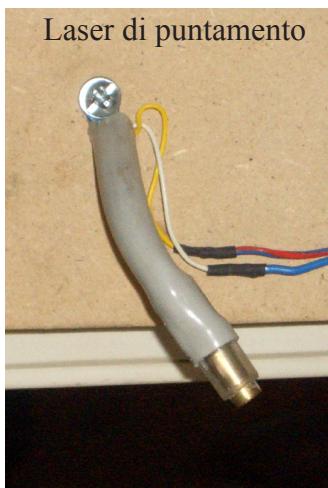
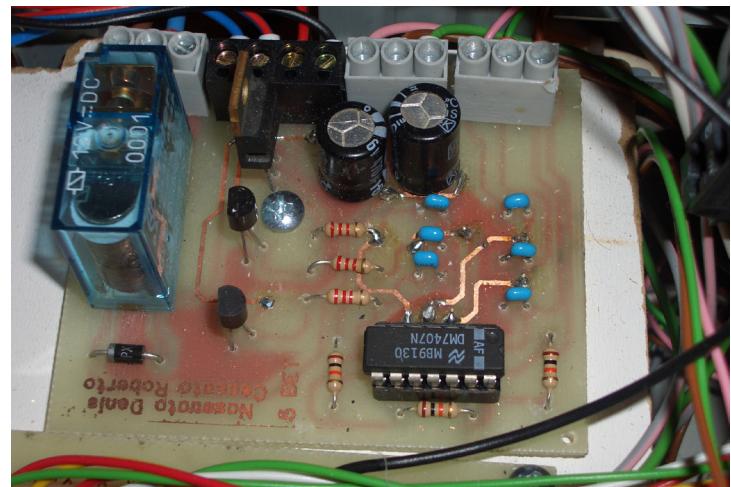
3. SERIGRAFIA - Posizione componenti

## ⇒ POTENZA DISPOSITIVI LUCE, LASER, FRESA

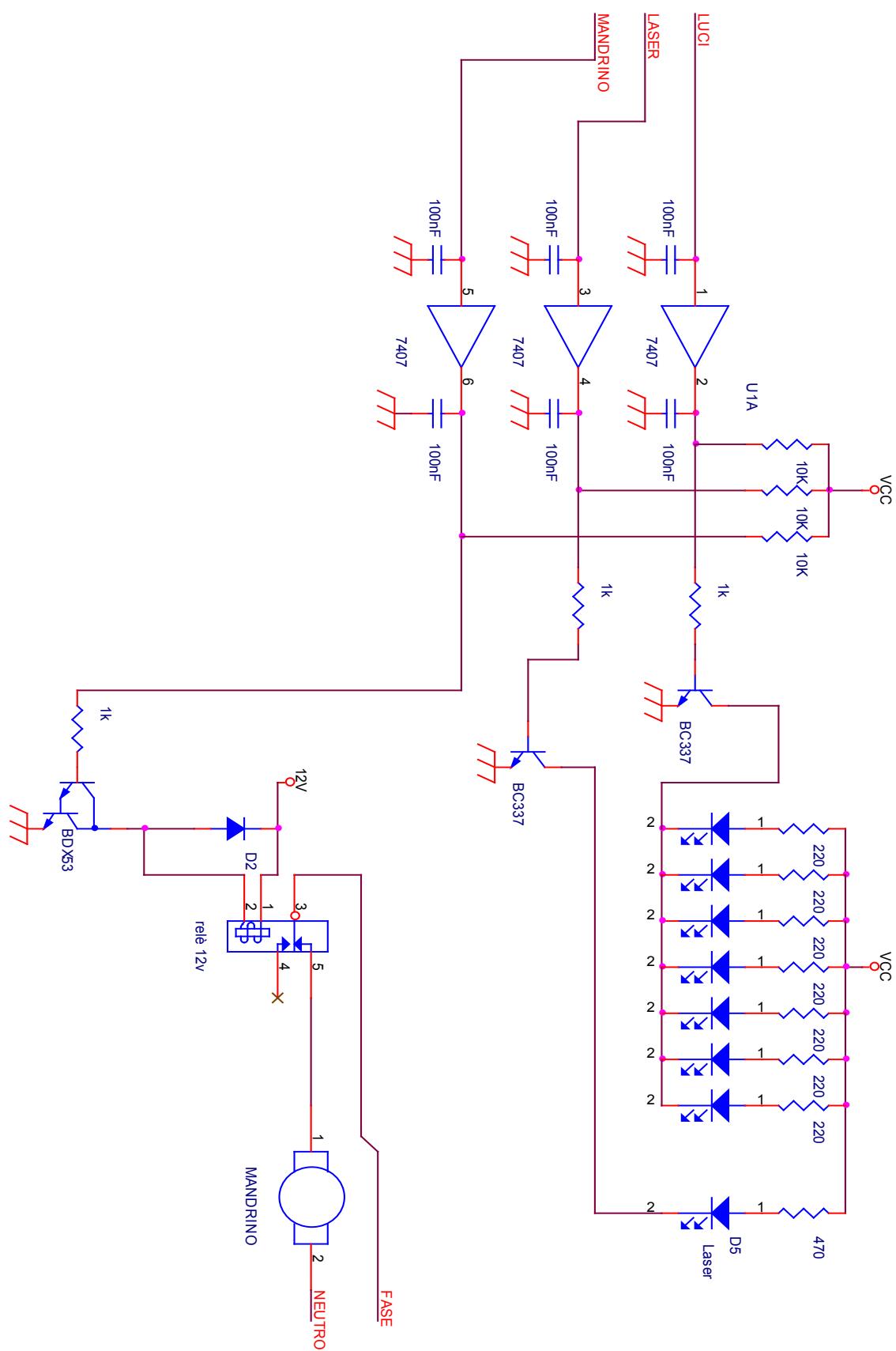
### ► DESCRIZIONE GENERALE

Il circuito è stato realizzato per riuscire a pilotare tramite ingressi digitali (TTL), 3 dispositivi (luce lavorazione, laser di puntamento e mandrino frese), mediante l'utilizzo di un transistor (bc337) per i led ad alta luminosità della zona lavoro; un altro transistor per pilotare il laser di puntamento posizione utensile ed infine un terzo transistor (BDX53) per azionare un relè il quale provvede all'interruzione dell'alimentazione del mini trapano di precisione.

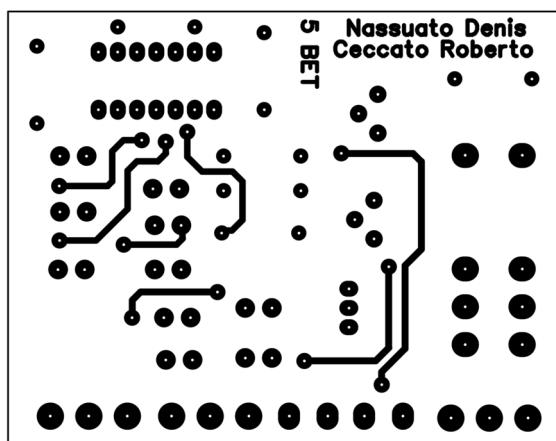
Per avere un'ampia illuminazione del piano di lavoro, sono stati collegati 7 led del tipo ad alta luminosità (10000 mcd) al di sotto dell'asse verticale Z.



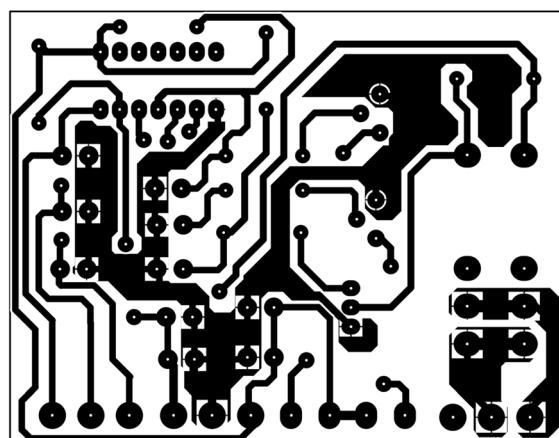
## ► SCHEMA ELETTRICO



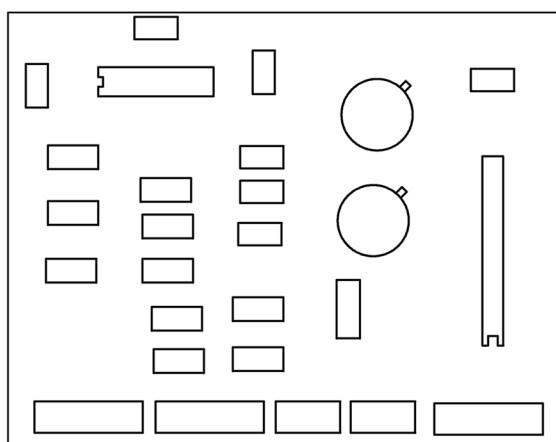
## ► PCB



1. TOP COPPER - Lato componenti



2. BOTTOM COPPER - Lato rame



3. SERIGRAFIA - Posizione componenti

## ⇒ COLONNA DI SEGNALAZIONE VISIVA E SEGNALAZIONE ACUSTICA

### ► DESCRIZIONE GENERALE

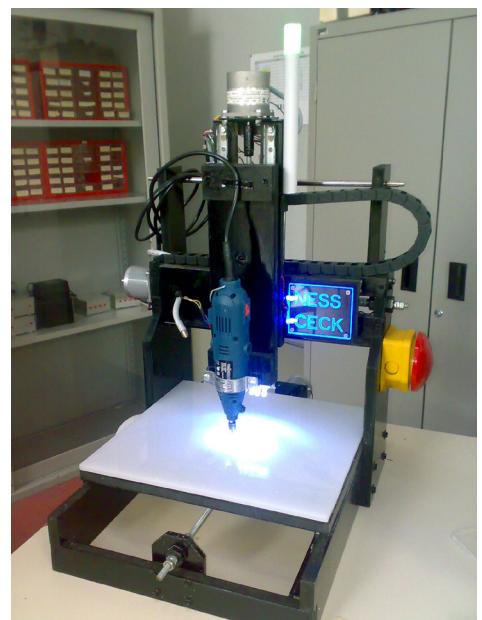
Le segnalazioni luminose e sonore sono un elemento fondamentale nella sicurezza di un impianto.

Per evitare errate interpretazioni è stata introdotta una normativa europea che attribuisce un significato univoco ai segnali visivi o sonori.

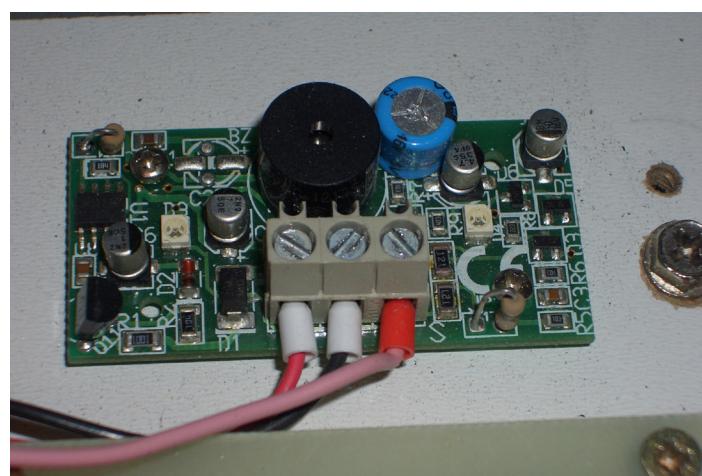
STATO MACCHINA OK



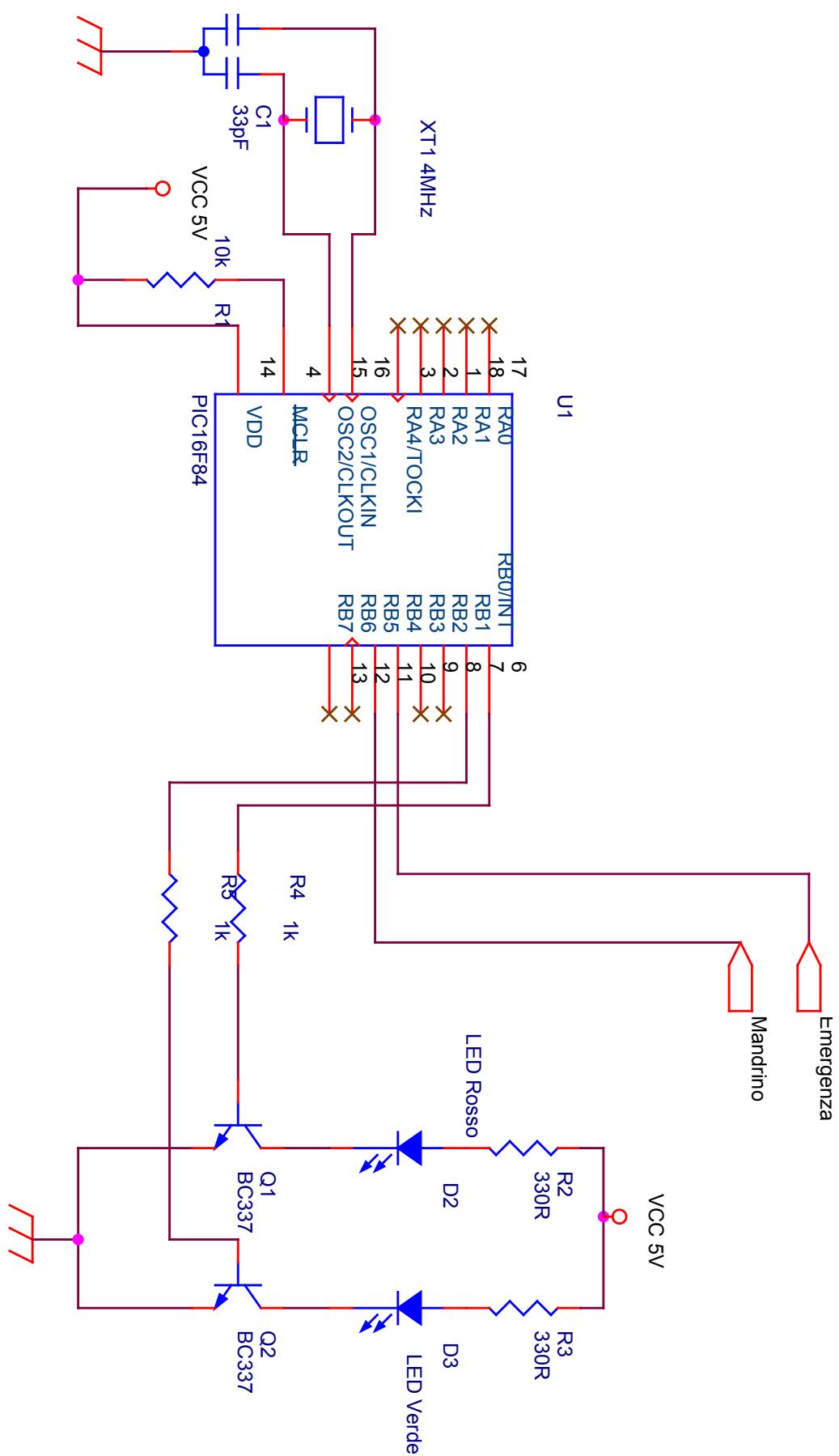
STATO MACCHINA EMERGENZA



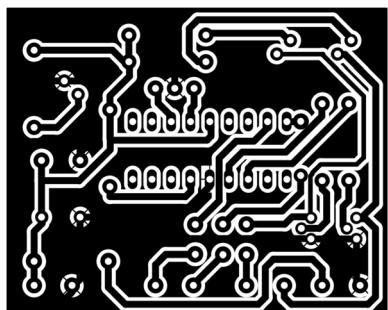
SIRENA ATTIVATA IN CASO DI  
EMERGENZA



## ► SCHEMA ELETTRICO

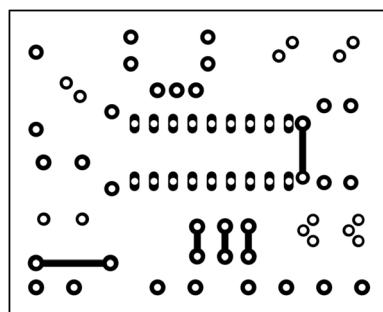


## ► PCB

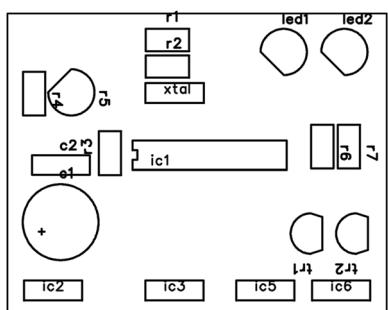


1. BOTTOM COPPER - Lato rame

2. TOP COPPER - Lato componenti



3. SERIGRAFIA - Posizione componenti



```
*****  
;*  
;* CECCATO ROBERTO - NASSUATO DENIS  
;*  
;*  
;* 5 BET  
;*  
;*  
;* SOFTWARE PIC 16F84A  
;* PER GESTIONE PULSANTIERA CONTROLLO CNC A 3 ASSI  
;*  
;*  
;*****
```

; SOFTWARE DI GESTIONE COLONNINA LUMINOSA DI SEGNALAZIONE PER:

- ; 1- STATO MACCHINA OK (LED VERDE FISSO)
- ; 2- STATO MACCHINA OK E MANDRINO ATTIVO (LED VERDE LAMPEGGIANTE)
- ; 3- STATO MACCHINA EMERGENZA (LED ROSSO FISSO)
- ; 4- STATO MACCHINA EMERGENZA CON MADRINO
- ; ANCORA ATTIVO (ROSSO/VERDE ALTERNATI)

; TABELLA PIEDINATURA INGRESSI USCITE PIC

RA0	-->	NC
RA1	-->	NC
RA2	-->	NC
RA3	-->	NC
RA4	-->	NC
<hr/>		
RB0	-->	SEGNALE DI ATTIVAZIONE LED ROSSO
RB1	-->	SEGNAL DI ATTIVAZIONE LED VERDE
RB2	-->	NC
RB3	-->	NC
RB4	-->	INGRESSO SEGNAL ATTIVAZIONE MANDRINO
RB5	-->	INGRESSO SEGNAL EMERGENZA MACCHINA
RB6	-->	NC
RB7	-->	NC
<hr/>		

;----- DIRETTIVE -----

```
PROCESSOR 16F84A  
RADIX DEC  
INCLUDE "P16F84A.INC"  
_CONFIG 3FF1H
```

;WatchDog Timer	OFF
;Power Up Timer	OFF
;Data Code Protect	OFF

----- Registri temporanei -----

Count EQU	20H	;registro contatore delay
Count1 EQU	21H	;registro contatore delay
test tempo EQU	22H	;registro per il nà di ritardi da eseguire per attendere il minimo di attesa per l'avvio del test uscite

----- Reset Vector -----

----- Start point at CPU reset -----

ORG 00H

----- INIZIO -----

INIZIO	BSF	STATUS,RP0	;Bank 1
	MOVLW	00H	;imposto PORTA
	MOVWF	TRISA	;con RA0:RA5 uscite
	MOVLW	0F0H	;imposto PORTB
gressi	MOVWF	TRISB	;con RB0,RB1,RB3,RB7 uscite e RB2,RB4:RB6 ingressi
	BCF	STATUS,RP0	;Bank 0
	CLRF	PORTB	;uscite di PORTB a zero
	CLRF	PORTA	;uscite di PORTA a zero

----- PROGRAM START -----

	BCF	PORTB,0	;spengo led rosso se attivo
	BCF	PORTB,1	;spengo led verde se attivo
MAIN	BTFS	PORTB,5	;testo ingresso emergenza
	GOTO	EME	;se attivo salto al controllo mandrino
	BTFFS	PORTB,5	;e attivazione led rosso o rosso/verde
	GOTO	V	;altrimenti testo nuovamente l'emergenza
			;se non è attivo salto al controllo mandrino
			;e attivazione led verde o verde lampeggiante
	GOTO	MAIN	;torna al programma principale MAIN LOOP

----- STATO EMERGENZA -----

EME	BTFS	PORTB,4	;testo se è attivo il mandrino
	GOTO	RV	;se si lampeggio rosso/verde
	BCF	PORTB,1	;se no spengo led verde
	BSF	PORTB,0	;accendo led rosso
	GOTO	MAIN	;torna al MAIN LOOP

----- STATO OK -----

V	BTFSC	PORTB,4	;testo se è attivo il mandrino
	GOTO	VL	;se si lampeggio verde
	BCF	PORTB,0	;se no spengo led rosso
	BSF	PORTB,1	;accendo led verde
	GOTO	MAIN	;torna al MAIN LOOP

----- EMERGE E MANDRINO ATT -----

RV	BCF	PORTB,0	;spengo led rosso
	BCF	PORTB,1	;spengo led verde
	BSF	PORTB,1	;accendo led verde
	CALL	DELAY	
	CALL	DELAY	
	CALL	DELAY	;attendo circa 800ms
	CALL	DELAY	
	BCF	PORTB,1	;spengo led verde
	BSF	PORTB,0	;accendo led rosso
	CALL	DELAY	
	CALL	DELAY	
	CALL	DELAY	;attendo circa 800ms
	CALL	DELAY	
	BCF	PORTB,0	;spengo led rosso
	GOTO	MAIN	;torna al MAIN LOOP

----- MANDRINO ATTIVO -----

VL	BCF	PORTB,0	;spengo led rosso
	BCF	PORTB,1	;accendo led verde
	CALL	DELAY	
	CALL	DELAY	
	CALL	DELAY	;attendo circa 800ms
	CALL	DELAY	
	BSF	PORTB,1	;accendo led verde
	CALL	DELAY	
	CALL	DELAY	
	CALL	DELAY	;attendo circa 800ms
	CALL	DELAY	
	BCF	PORTB,1	;spengo led verde
	GOTO	MAIN	;torna al MAIN LOOP

----- DELAY circa 200 ms -----

DELAY

CLRF Count  
CLRF Count1

DELAYLOOP2

DECFSZ Count,1  
GOTO DELAYLOOP2

DECFSZ Count1,1  
GOTO DELAYLOOP2

RETURN

----- FINE PROGRAMMA -----

END

## ⇒ PULSANTIERA DI CONTROLLO

### ► DESCRIZIONE GENERALE

La pulsantiera di controllo ha il compito di gestire un totale di 8 segnali di comando per la macchina. Essa è gestita internamente da un microcontrollore Microchip, un PIC 16F872. Il µC opera ad una frequenza di clock dettata dal quarzo a 4MHz.

I segnali nello specifico sono:

- Enable drive motore asse X
- Enable drive motore asse Y
- Enable drive motore asse Z
- Luce piano di lavorazione
- Laser di punzattamento posizione utensile
- Mandrino fresa
- Segnale di pausa lavorazione
- Segnale di emergenza generale macchina

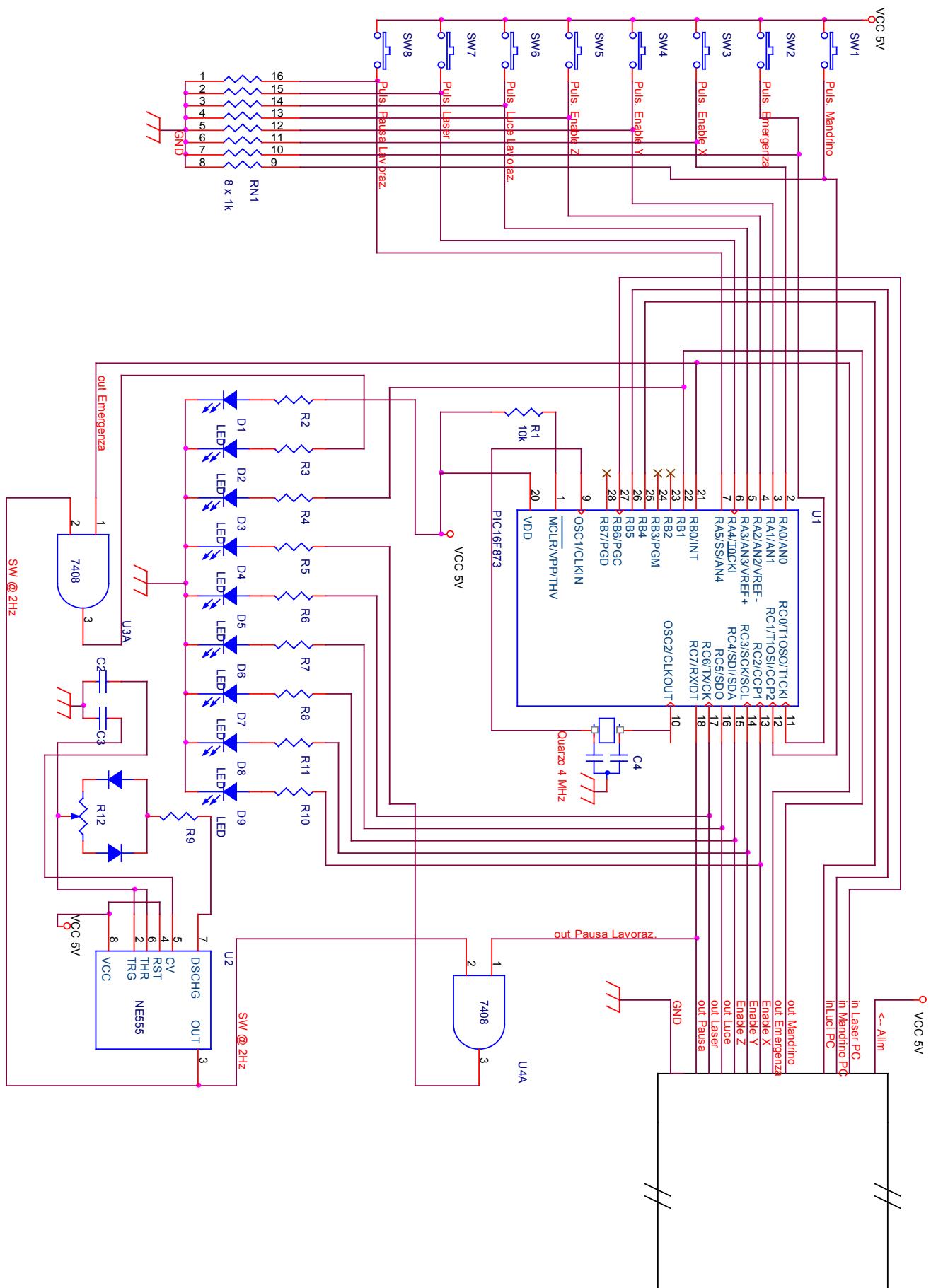
Ogni segnale può essere attivato o disattivato tramite il relativo pulsante, allo stesso tempo un led segnalera lo stato della linea.

### ► FUNZIONAMENTO

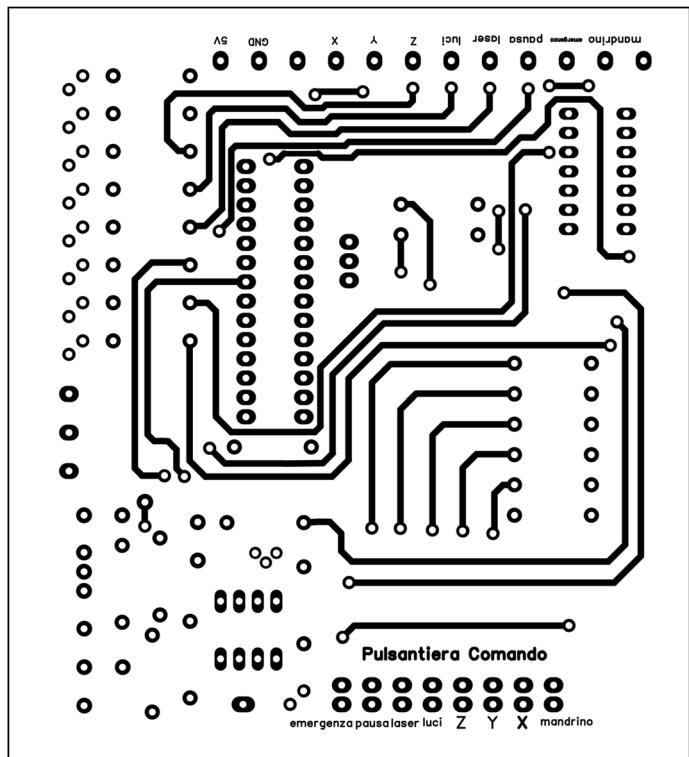
Nel momento in cui la macchina viene accesa il microcontrollore di gestione manderà attivo il segnale di emergenza e non è possibile attivare nessun comando affinchè non verrà premuto il pulsante di emergenza posto sulla pulsantiera stessa o sulla macchina. A questo punto, dopo aver disattivato l'emergenza, verranno abilitati i 3 drive dei motori e inoltre sarà possibile gestire tutte le altre linee di controllo.



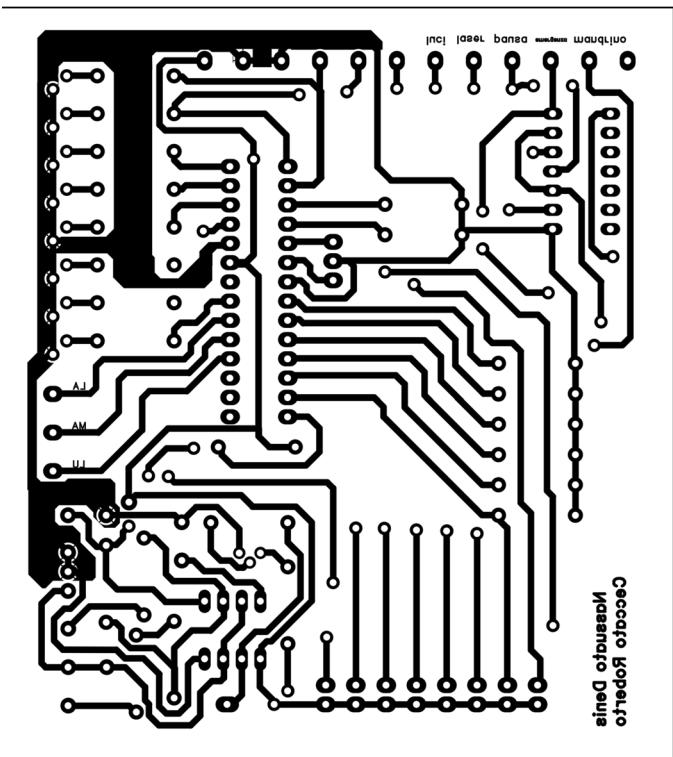
# ► SCHEMA ELETTRICO



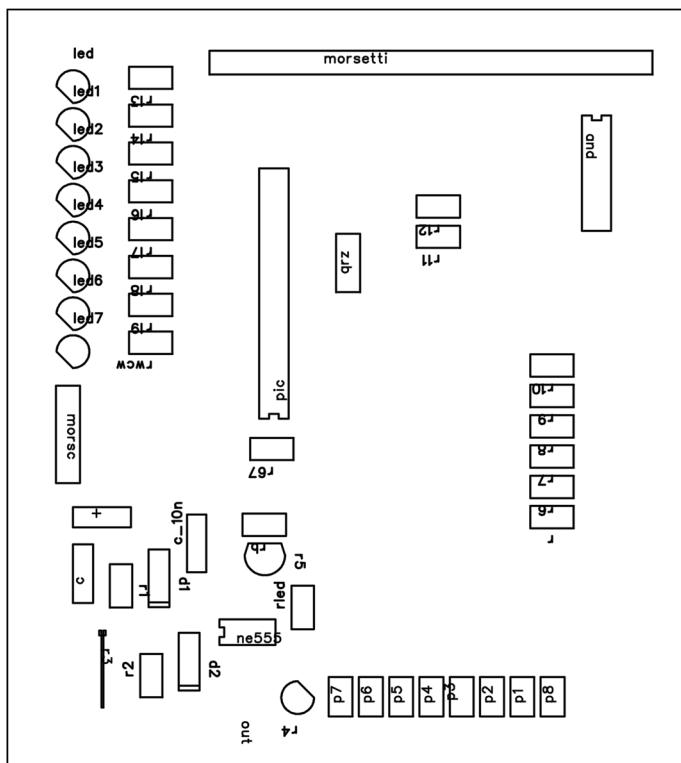
## ► PCB



1. TOP COPPER - Lato componenti



2. BOTTOM COPPER - Lato rame



3. SERIGRAFIA - Posizione componenti

\*\*\*\*\*  
.\* CECCATO ROBERTO - NASSUATO DENIS  
.\*  
.\*

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\*\*\*\*\*  
.\* SOFTWARE PIC 16F87X (16F872)  
.\* PER GESTIONE PULSANTIERA CONTROLLO CNC A 3 ASSI  
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SOFTWARE DI GESTIONE PULSANTIERA MACCHINA CNC A 3 ASSI  
LA PULSANTIERA COMANDA 8 SEGNALI DIVERSI PER IL CONTROLLO DI:

- 1- ABILITAZIONE SCHEDA MOTORE ASSE X
- 2- ABILITAZIONE SCHEDA MOTORE ASSE Y
- 3- ABILITAZIONE SCHEDA MOTORE ASSE Z
- 4- ON/OFF LUCE LAVORAZIONE TRAMITE PULSANTE O ESTERNO
- 5- ON/OFF LASER PUNTAMENTO POSIZIONE REALE UTENSILE SUL PIANO DI LAVORO
- 6- ON/OFF MANDRINO UTENSILE
- 7- ON/OFF SEGNALE DI PAUSA DURANTE LA LAVORAZIONE (USATO PER PIC O ALTRO)
- 8- SEGNALE FUNGO DI EMERGENZA

TABELLA PIEDINATURA INGRESSI USCITE PIC

RA0	⇒	PULSANTE ENABLE ASSE X
RA1	⇒	PULSANTE ENABLE ASSE Y
RA2	⇒	PULSANTE ENABLE ASSE Z
RA3	⇒	PULSANTE ON/OFF LUCI LAVORAZIONE
RA4	⇒	PULSANTE ON/OFF LASER PUNTAMENTO
RA5	⇒	PULSANTE ON/OFF PAUSA LAVORAZIONE
<hr/>		
RB0	⇒	USCITA SEGNALE EMERGENZA
RB1	⇒	USCITA ON/OFF MANDRINO
RB2	⇒	LIBERO
RB3	⇒	LIBERO
RB4	⇒	INGRESSO ESTERNO ON/OFF LUCE LAVORAZIONE x interrupt
RB5	⇒	INGRESSO ESTERNO ON/OFF MANDRINO UTENSILE x interrupt
RB6	⇒	INGRESSO ESTERNO ON/OFF LASER PUNTAMENTO x interrupt
RB7	⇒	LIBERO
<hr/>		
RC0	⇒	PULSANTE EMERGENZA
RC1	⇒	PULSANTE ON/OFF MANDRINO UTENSILE
RC2	⇒	USCITA ABILITAZIONE ASSE X
RC3	⇒	USCITA ABILITAZIONE ASSE Y
RC4	⇒	USCITA ABILITAZIONE ASSE Z
RC5	⇒	USCITA ON/OFF LUCI LAVORAZIONE
RC6	⇒	USCITA ON/OFF LASER PUNTAMENTO
RC7	⇒	USCITA ON/OFF PAUSA LAVORAZIONE
<hr/>		

----- DIRETTIVE -----

```
PROCESSOR 16F872
RADIX DEC
INCLUDE "P16F872.INC"
__CONFIG 3FF9H
```

;WatchDog Time	OFF
;Power Up Timer	ON
;Brown Out Reset	ON
;Low Voltage Programming	ON
;Data Code Protect	OFF
;Flash Program Memory Write Enable	ON
;Background Debugger Mode	OFF

----- Registri temporanei -----

Count EQU	20H	;registro contatore delay
Count1 EQU	21H	;registro contatore delay
test EQU	22H	;registro per il nà di ritardi da eseguire per attendere il tempo ;minimo di attesa per l'avvio del test uscite

----- Reset Vector -----

----- Start point at CPU reset -----

```
ORG 00H
```

```
GOTO INIZIO
```

----- Start point at INTERRUPT EVENT -----

ORG	04H	
BTFSS	INTCON,RBIF	;verifico se è avvenuto l'interrupt ; sul cambiamento di livello RB4:RB7
RETFIE		
BCF	INTCON,RBIF	;se sì:
CALL	DELAY	;attende circa 1 ms
BTFS	PORTB,6	
CALL	INT_LSR	;gestione accensione laser da pc
CALL	DELAY	
BTFS	PORTB,5	
CALL	INT_MAND	;gestione accensione mandrino da pc
CALL	DELAY	
BTFS	PORTB,4	
CALL	INT_LUCI	;gestione accensione luce da pc
CALL	DELAY	
RETFIE		

----- INIZIO -----

INIZIO	BSF BCF	STATUS,RP0 STATUS,RP1	;Bank 1
	MOVLW MOVWF	06H ADCON1	;Configura i pin di PORTA ;come ingressi digitali
	MOVLW MOVWF	03FH TRISA	;imposto PORTA ‘0011111’ ;con RA0:RA5 ingressi
gressi	MOVLW MOVWF	074H TRISB	;imposto PORTB ‘01110100’ ;con RB0,RB1,RB3,RB7 uscite e RB2,RB4:RB6 in-
	MOVLW MOVWF	003H TRISC	;imposto PORTC ‘00000011’ ;con RC0,RC1 ingressi e RC2:RC7 uscite
	BCF BCF	STATUS,RP0 STATUS,RP1	;Bank 0
	CLRF CLRF	PORTB PORTC	;uscite di PORTB a zero ;uscite di PORTC a zero
	BSF BSF	INTCON,GIE INTCON,RBIE	;abilita interrupt generale ;abilita interrupt sul cambiamento di livello RB4:RB7

----- PROGRAM START -----

START	BSF CALL	PORTB,0 DELAY_200ms	;attivo EMERGENZA ;attende 200ms
WAIT	BTFSS GOTO CALL	PORTC,0 WAIT EM	;rimane in attesa ;della pressione del tasto di emergenza ;se viene premuto cambia stato all'uscita portandola a 0
	CALL DELAY		;attende circa 1 ms
	BSF CALL BSF CALL BSF CALL	PORTC,2 DELAY_200ms PORTC,3 DELAY_200ms PORTC,4 DELAY_200ms	;abilita l'asse X ;attende 200ms ;abilita l'asse Y ;attende 200ms ;abilita l'asse Z ;attende 200ms
	GOTO	MAIN	;salta al programma principale ;di gestione della pulsantiera

----- MAIN -----

MAIN

BTFSC	PORTA,0	;testa il pulsante di abilitazione asse X se premuto
CALL	ENX	;cambia di stato la relativa uscita
BTFSC	PORTA,1	;altrimenti testa il pulsante di abilitazione
		;dell'asse Y se premuto
CALL	ENY	;cambia di stato la relativa uscita
BTFSC	PORTA,2	;altrimenti testa il pulsante di abilitazione
		;dell'asse Z se premuto
CALL	ENZ	;cambia di stato la relativa uscita
BTFSC	PORTA,3	;altrimenti testa il pulsante di on/off luci se premuto
CALL	LUCI	;cambia di stato la relativa uscita
BTFSC	PORTA,4	;altrimenti testa il pulsante di on/off laser se premuto
CALL	LASER	;cambia di stato la relativa uscita
BTFSC	PORTA,5	;altrimenti testa il pulsante di pausa se premuto
CALL	PAUSA	;cambia di stato la relativa uscita
BTFSC	PORTC,0	;altrimenti testa il pulsante di emergenza se premuto
CALL	EM	;cambia di stato la relativa uscita
BTFSC	PORTC,1	;altrimenti testa il pulsante di
		;on/off mandrino utensile se premuto
CALL	MANDR	;cambia di stato la relativa uscita
GOTO	MAIN	;ricomincia la sequenza di controllo pulsantiera

----- SUBROUTINE GESTIONE SEGNALI -----

----- abilitazione scheda asse X -----

ENX	BTFSC	PORTA,0	;se pulsante ancora premuto
	GOTO	ENX	;rimane in attesa
	CALL	DELAY	;altrimenti
	BTFSC	PORTC,2	;controlla se l'uscita è a livello basso
	GOTO	ELSE_ENX	;se no viene posta a 0 con salto all'else
	BSF	PORTC,2	;se si viene posta a 1
	GOTO	ENDIF_ENX	;salta alla fine
ELSE_ENX			
	BCF	PORTC,2	;else per la disattivazione dell'uscita se a valore 1
ENDIF_ENX			
	CALL	DELAY	
	RETURN		;ritorna la programma principale

----- abilitazione scheda asse Y -----

ENY	BTFS	PORTA,1	;se pulsante ancora premuto
	GOTO	ENY	;rimane in attesa
	CALL	DELAY	;altrimenti
	BTFS	PORTC,3	;controlla se l'uscita è a livello basso
	GOTO	ELSE_ENY	;se no viene posta a 0 con salto all'else
	BSF	PORTC,3	;se si viene posta a 1
	GOTO	ENDIF_ENY	;salta alla fine
ELSE_ENY			
	BCF	PORTC,3	;else per la disattivazione dell'uscita se a valore 1
ENDIF_ENY			
	CALL	DELAY	
		RETURN	;ritorna la programma principale

----- Abilitazione scheda asse Y -----

ENZ	BTFS	PORTA,2	;se pulsante ancora premuto
	GOTO	ENZ	;rimane in attesa
	CALL	DELAY	;altrimenti
	BTFS	PORTC,4	;controlla se l'uscita è a livello basso
	GOTO	ELSE_ENZ	;se no viene posta a 0 con salto all'else
	BSF	PORTC,4	;se si viene posta a 1
	GOTO	ENDIF_ENZ	;salta alla fine
ELSE_ENZ			
	BCF	PORTC,4	;else per la disattivazione dell'uscita se a valore 1
ENDIF_ENZ			
	CALL	DELAY	
		RETURN	;ritorna la programma principale

----- ON OFF LUCI -----

LUCI	BTFS	PORTA,3	;se pulsante ancora premuto
	GOTO	LUCI	;rimane in attesa
	BTFS	PORTB,4	;se già attivo da pc esce
	RETURN		;tornando al programma principale
	CALL	DELAY	;altrimenti
	BTFS	PORTC,5	;controlla se l'uscita è a livello basso
	GOTO	ELSE_LUCI	;se no viene posta a 0 con salto all'else
	BSF	PORTC,5	;se si viene posta a 1
	GOTO	ENDIF_LUCI	;salta alla fine
ELSE_LUCI			
	BCF	PORTC,5	;else per la disattivazione dell'uscita se a valore 1
ENDIF_LUCI			
	CALL	DELAY	
		RETURN	;ritorna la programma principale

----- ON OFF LASER -----

LASER	BTFS	C PORTA,4	;se pulsante ancora premuto
	GOTO	LASER	;rimane in attesa
	BTFS	PORTB,2	;se già attivo da pc esce
	RETURN		;tornando al programma principale
	CALL	DELAY	;altrimenti
	BTFS	PORTC,6	;controlla se l'uscita è a livello basso
	GOTO	ELSE_LSR	;se no viene posta a 0 con salto all'else
	BSF	PORTC,6	;se si viene posta a 1
	GOTO	ENDIF_LSR	;salta alla fine
ELSE_LSR	BCF	PORTC,6	
ENDIF_LSR	CALL	DELAY	
	RETURN		;ritorna la programma principale

----- ON OFF MANDRINO -----

MANDR	BTFS	C PORTC,1	;se pulsante ancora premuto
	GOTO	MANDR	;rimane in attesa
	BTFS	PORTB,6	;se già attivo da pc esce
	RETURN		;tornando al programma principale
	CALL	DELAY	;altrimenti
	BTFS	PORTB,1	;controlla se l'uscita è a livello basso
	GOTO	ELSE_MAN	;se no viene posta a 0 con salto all'else
	BSF	PORTB,1	;se si viene posta a 1
	GOTO	ENDIF_MAN	;salta alla fine
ELSE_MAN	BCF	PORTB,1	
ENDIF_MAN	CALL	DELAY	
	RETURN		;ritorna la programma principale

----- ATTIVAZIONE EMERGENZA -----

EM	BTFS	C PORTC,0	;se pulsante ancora premuto
	GOTO	EM	;rimane in attesa
	CALL	DELAY	;altrimenti
	BTFS	PORTB,0	;controlla se l'uscita è a livello basso
	GOTO	ELSE_EM	;se no viene posta a 0 con salto all'else
	BSF	PORTB,0	;se si viene posta a 1
	GOTO	ENDIF_EM	;salta alla fine
ELSE_EM	BCF	PORTB,0	
ENDIF_EM	CALL	DELAY	
	RETURN		;ritorna la programma principale

----- ON OFF PAUSA -----

PAUSA	BTFS	PORTA,5	;se pulsante ancora premuto
	GOTO	PAUSA	;rimane in attesa
	CALL	DELAY	;altrimenti
	BTFS	PORTC,7	;controlla se l'uscita è a livello basso
	GOTO	ELSE_PAUSA	;se no viene posta a 0 con salto all'else
	BSF	PORTC,7	;se si viene posta a 1
	GOTO	ENDIF_PAUSA	;salta alla fine
ELSE_PAUSA			
	BCF	PORTC,7	;else per la disattivazione dell'uscita se a valore 1
ENDIF_PAUSA			
	CALL	DELAY	
		RETURN	;ritorna la programma principale

----- ON OFF LUCI FROM INTERRUPT -----

INT_LUCI	BTFS	PORTB,4	;controlla se l'ingresso è a livello alto
	GOTO	ELSE_EXT_LUCI	;se no viene posta a 0 con salto all'else
	BSF	PORTC,5	;se si viene posta a 1
	GOTO	ENDIF_EXT_LUCI	;salta alla fine
ELSE_EXT_LUCI			
	BCF	PORTC,5	;else per la disattivazione dell'uscita
ENDIF_EXT_LUCI			
		RETURN	

----- ON OFF MANDRINO FROM INTERRUPT -----

INT_MAND	BTFS	PORTB,5	;controlla se l'ingresso è a livello alto
	GOTO	ELSE_EXT_MAND	;se no viene posta a 0 con salto all'else
	BSF	PORTB,1	;se si viene posta a 1
	GOTO	ENDIF_EXT_MAND	;salta alla fine
ELSE_EXT_MAND			
	BCF	PORTB,1	;else per la disattivazione dell'uscita
ENDIF_EXT_MAND			
		RETURN	

----- ON OFF LASER FROM INTERRUPT -----

INT_LSR	BTFS	PORTB,6	;controlla se l'ingresso è a livello alto
	GOTO	ELSE_EXT_LSR	;se no viene posta a 0 con salto all'else
	BSF	PORTC,6	;se si viene posta a 1
	GOTO	ENDIF_EXT_LSR	;salta alla fine
ELSE_EXT_LSR			
	BCF	PORTC,6	;else per la disattivazione dell'uscita
ENDIF_EXT_LSR			
		RETURN	

----- DELAY 0.765 ms-----

DELAY

CLRF	Count
CLRF	Count1

DELAYLOOP

DECFSZ	Count,1
GOTO	DELAYLOOP

;DECFSZ	Count1,1
;GOTO	DELAYLOOP

RETURN

----- DELAY circa 200 ms-----

DELAY\_200ms

CLRF	Count
CLRF	Count1

DELAYLOOP2

DECFSZ	Count,1
GOTO	DELAYLOOP2

DECFSZ	Count1,1
GOTO	DELAYLOOP2

RETURN

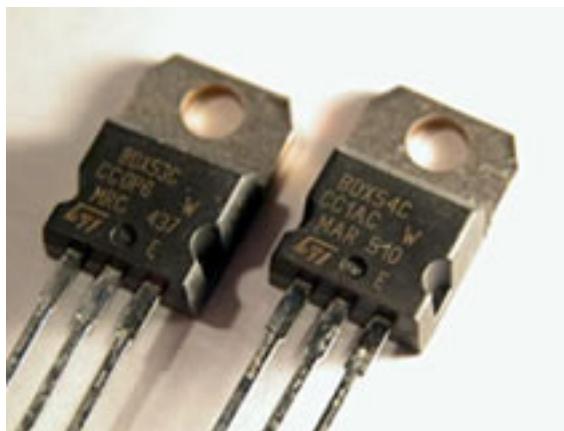
----- FINE PROGRAMMA -----

END

# DATASHEET

## DEI PRINCIPALI COMPONENTI UTILIZZATI

- 7407
- 7408
- 7432
- L297
- BDX 53
- PIC 16F872
- PIC 16F84A



**SN54LS07, SN74LS07, SN74LS17**  
**HEX BUFFERS/DRIVERS WITH**  
**OPEN-COLLECTOR HIGH-VOLTAGE OUTPUTS**

SDLS021A, D3517, MAY 1990-REVISED AUGUST 1991

- Converts TTL-Voltage Levels to MOS Levels
- High Sink-Current Capability
- Input Clamping Diodes Simplify System Design
- Open-Collector Driver for Indicator Lamps and Relays
- Package Options Include "Small Outline" Packages, Ceramic Chip Carriers, and Standard and Ceramic 300-mil DIPs

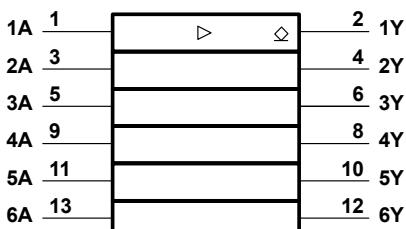
### description

These monolithic hex buffers/drivers feature high-voltage open-collector outputs to interface with high-level circuits or for driving high-current loads. They are also characterized for use as buffers for driving TTL inputs. The 'LS07 has a rated output voltage of 30 V and the 'LS17 has a rated output voltage of 15 V. The maximum sink current is 30 mA for the SN54LS07 and 40 mA for the SN74LS07 and SN74LS17.

These circuits are compatible with most TTL families. Inputs are diode-clamped to minimize transmission-line effects, which simplifies design. Typical power dissipation is 140 mW and average propagation delay time is 12 ns.

The SN54LS07 is characterized over the full military temperature range of -55°C to 125°C. The SN74LS07 and SN74LS17 are characterized for operation from 0°C to 70°C.

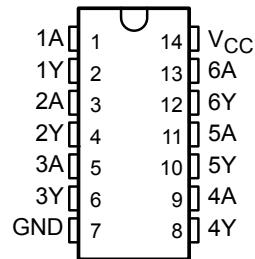
### logic symbol†



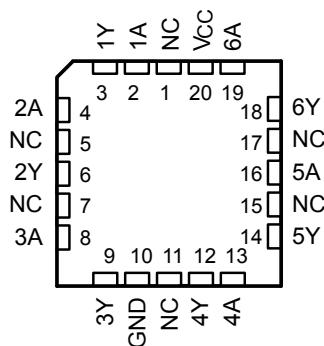
† This symbol is in accordance with ANSI/IEEE Std 91-1984 and IEC Publication 617-12.

Pin numbers shown are for D, J, and N packages.

**SN54LS07 . . . J PACKAGE**  
**SN74LS07, SN74LS17 . . . D OR N PACKAGE**  
**(TOP VIEW)**

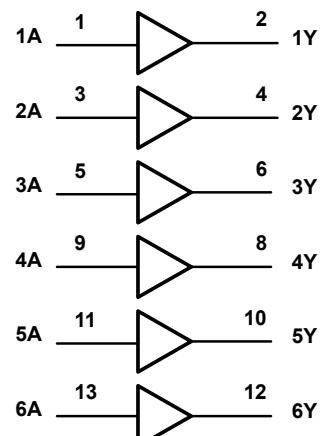


**SN54LS07 . . . FK PACKAGE**  
**(TOP VIEW)**



NC – No internal connection

### logic diagram (positive logic)

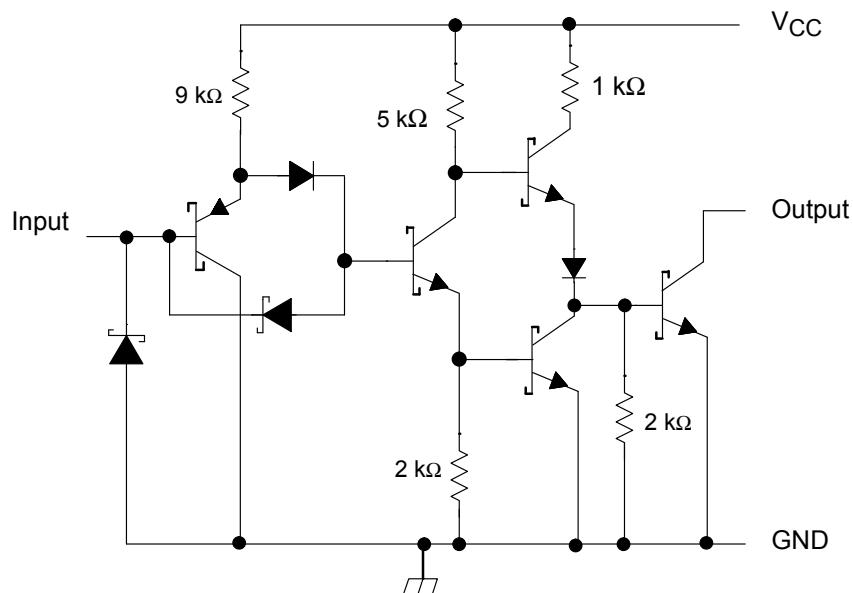


# SN54LS07, SN74LS07, SN74LS17

## HEX BUFFERS/DRIVERS WITH OPEN-COLLECTOR HIGH-VOLTAGE OUTPUTS

---

schematic (each gate)



Resistor values shown are nominal.

### absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, $V_{CC}$	7 V
Input voltage, $V_I$ (see Note 1)	5.5 V
Output voltage, $V_O$ (see Notes 1 and 2): SN54LS07, SN74LS07	30 V
SN74LS17	15 V
Operating free-air temperature range: SN54LS07	-55°C to 125°C
SN54LS07, SN74LS17	0°C to 70°C
Storage temperature range	-65°C to 150°C

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. Voltage values are with respect to network ground terminal.

2. This is the maximum voltage that should be applied to any output when it is in the off state.

### recommended operating conditions

		SN54LS07			SN74LS07 SN74LS17			UNIT
		MIN	NOM	MAX	MIN	NOM	MAX	
$V_{CC}$	Supply voltage	4.5	5	5.5	4.75	5	5.25	V
$V_{IH}$	High-level input voltage	2			2			V
$V_{IL}$	Low-level input voltage		0.8			0.8		V
$V_{OH}$	High-level output voltage	'LS07	30		30		30	V
			'LS17			15		
$I_{OL}$	Low-level output current		30		30		40	mA
$T_A$	Operating free-air temperature	-55		125	0		70	°C

SN54LS07, SN74LS07, SN74LS17  
HEX BUFFERS/DRIVERS WITH  
OPEN-COLLECTOR HIGH-VOLTAGE OUTPUTS

---

**electrical characteristics over recommended operating free-air temperature range (unless otherwise noted)**

PARAMETER	TEST CONDITIONS <sup>†</sup>	SN54LS07			SN74LS07 SN74LS17			UNIT
		MIN	TYP <sup>‡</sup>	MAX	MIN	TYP <sup>‡</sup>	MAX	
V <sub>IK</sub>	V <sub>CC</sub> = MIN, I <sub>I</sub> = -12 mA			-1.5			-1.5	V
I <sub>OH</sub>	V <sub>CC</sub> = MIN, V <sub>IH</sub> = 2 V	'LS07, V <sub>OH</sub> = 30 V		0.25			0.25	mA
		'LS17, V <sub>OH</sub> = 15 V		0.25			0.25	
V <sub>OL</sub>	V <sub>CC</sub> = MIN, V <sub>IL</sub> = 0.8 V	I <sub>OL</sub> = 16 mA		0.4			0.4	V
		I <sub>OL</sub> = MAX\$		0.7			0.7	
I <sub>I</sub>	V <sub>CC</sub> = MAX, V <sub>I</sub> = 7 V			1			1	mA
I <sub>IH</sub>	V <sub>CC</sub> = MAX, V <sub>I</sub> = 2.4 V			20			20	μA
I <sub>IL</sub>	V <sub>CC</sub> = MAX, V <sub>I</sub> = 0.4 V			-0.2			-0.2	mA
I <sub>CCH</sub>	V <sub>CC</sub> = MAX			14			14	mA
I <sub>CCL</sub>	V <sub>CC</sub> = MAX			45			45	mA

<sup>†</sup> For conditions shown as MIN or MAX, use the appropriate value specified under recommended operating conditions.

<sup>‡</sup> All typical values are at V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C.

\$ I<sub>OL</sub> = 30 mA for SN54 series parts and 40 mA for SN74 series parts.

**switching characteristics, V<sub>CC</sub> = 5 V, T<sub>A</sub> = 25°C (see Note 3)**

PARAMETER	FROM (INPUT)	TO (OUTPUT)	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub>	A	Y	R <sub>L</sub> = 110 Ω, C <sub>L</sub> = 15 pF	6	10		ns
t <sub>PHL</sub>				19	30		

NOTE 3: Load circuit and voltage waveforms are shown in Section 1 of *TTL Logic Data Book*, 1988.



POST OFFICE BOX 655303 • DALLAS, TEXAS 75265

# 54AC/74AC08 • 74ACT08

## Quad 2-Input AND Gate

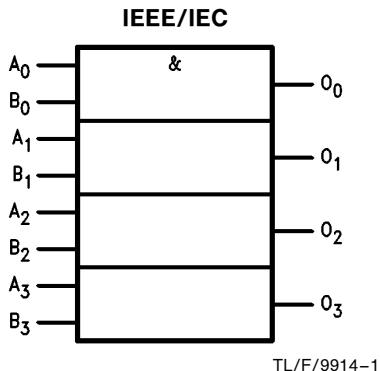
### General Description

The 'AC/'ACT08 contains four, 2-input AND gates.

### Features

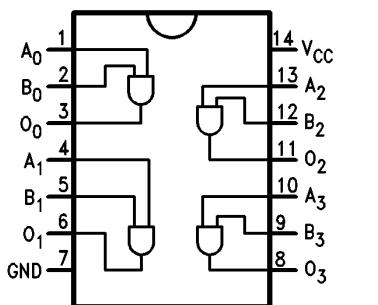
- $I_{CC}$  reduced by 50% on 54AC/74AC only
- Outputs source/sink 24 mA
- 'ACT08 has TTL-compatible inputs
- Standard Military Drawing (SMD)
  - 'AC08: 5962-87615
- For Military 54ACT08 device, see 54ACTQ08

### Logic Symbols



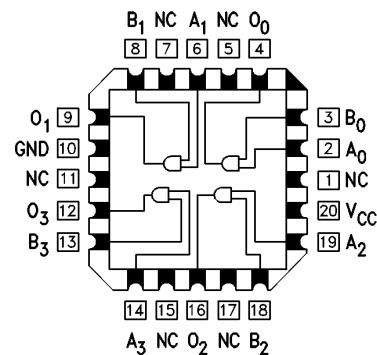
TL/F/9914-1

**Pin Assignment  
for DIP, Flatpak and SOIC**



TL/F/9914-3

**Pin Assignment  
for LCC**



TL/F/9914-2

Pin Names	Description
$A_n, B_n$	Inputs
$O_n$	Outputs

## Absolute Maximum Rating (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage ( $V_{CC}$ )	−0.5V to +7.0V	
DC Input Diode Current ( $I_{IK}$ )	$V_I = -0.5V$ $V_I = V_{CC} + 0.5V$	
	−20 mA	+20 mA
DC Input Voltage ( $V_I$ )	−0.5V to $V_{CC} + 0.5V$	
DC Output Diode Current ( $I_{OK}$ )	$V_O = -0.5V$ $V_O = V_{CC} + 0.5V$	
	−20 mA	+20 mA
DC Output Voltage ( $V_O$ )	−0.5V to $V_{CC} + 0.5V$	
DC Output Source or Sink Current ( $I_O$ )	$\pm 50\text{ mA}$	
DC $V_{CC}$ or Ground Current per Output Pin ( $I_{CC}$ or $I_{GND}$ )	$\pm 50\text{ mA}$	
Storage Temperature ( $T_{STG}$ )	−65°C to +150°C	
Junction Temperature ( $T_J$ )	CDIP 175°C PDIP 140°C	

**Note 1:** Absolute maximum ratings are those values beyond which damage to the device may occur. The databook specifications should be met, without exception, to ensure that the system design is reliable over its power supply, temperature, and output/input loading variables. National does not recommend operation of FACT™ circuits outside databook specifications.

## Recommended Operating Conditions

Supply Voltage ( $V_{CC}$ )	'AC	2.0V to 6.0V
	'ACT	4.5V to 5.5V
Input Voltage ( $V_I$ )		0V to $V_{CC}$
Output Voltage ( $V_O$ )		0V to $V_{CC}$
Operating Temperature ( $T_A$ )	74AC/ACT	−40°C to +85°C
	54AC	−55°C to +125°C
Minimum Input Edge Rate ( $\Delta V/\Delta t$ )	'AC Devices	
	$V_{IN}$ from 30% to 70% of $V_{CC}$	
	$V_{CC}$ @ 3.3V, 4.5V, 5.5V	125 mV/ns
Minimum Input Edge Rate ( $\Delta V/\Delta t$ )	'ACT Devices	
	$V_{IN}$ from 0.8V to 2.0V	
	$V_{CC}$ @ 4.5V, 5.5V	125 mV/ns

## DC Characteristics for 'AC Family Devices

Symbol	Parameter	$V_{CC}$ (V)	74AC		54AC	74AC	Units	Conditions
			$T_A = +25^\circ C$		$T_A = -55^\circ C \text{ to } +125^\circ C$	$T_A = -40^\circ C \text{ to } +85^\circ C$		
			Typ	Guaranteed Limits				
$V_{IH}$	Minimum High Level Input Voltage	3.0 4.5 5.5	1.5 2.25 2.75	2.1 3.15 3.85	2.1 3.15 3.85	2.1 3.15 3.85	V	$V_{OUT} = 0.1V$ or $V_{CC} - 0.1V$
$V_{IL}$	Maximum Low Level Input Voltage	3.0 4.5 5.5	1.5 2.25 2.75	0.9 1.35 1.65	0.9 1.35 1.65	0.9 1.35 1.65	V	$V_{OUT} = 0.1V$ or $V_{CC} - 0.1V$
$V_{OH}$	Minimum High Level Output Voltage	3.0 4.5 5.5	2.99 4.49 5.49	2.9 4.4 5.4	2.9 4.4 5.4	2.9 4.4 5.4	V	$I_{OUT} = -50\text{ }\mu A$
		3.0 4.5 5.5		2.56 3.86 4.86	2.4 3.7 4.7	2.46 3.76 4.76	V	* $V_{IN} = V_{IL}$ or $V_{IH}$ −12 mA $I_{OH}$ −24 mA −24 mA
$V_{OL}$	Maximum Low Level Output Voltage	3.0 4.5 5.5	0.002 0.001 0.001	0.1 0.1 0.1	0.1 0.1 0.1	0.1 0.1 0.1	V	$I_{OUT} = 50\text{ }\mu A$
		3.0 4.5 5.5		0.36 0.36 0.36	0.5 0.5 0.5	0.44 0.44 0.44	V	* $V_{IN} = V_{IL}$ or $V_{IH}$ 12 mA $I_{OL}$ 24 mA 24 mA
$I_{IN}$	Maximum Input Leakage Current	5.5		$\pm 0.1$	$\pm 1.0$	$\pm 1.0$	$\mu A$	$V_I = V_{CC}, GND$

\*All outputs loaded; thresholds on input associated with output under test.

## DC Characteristics for 'AC Family Devices (Continued)

Symbol	Parameter	V <sub>CC</sub> (V)	74AC		54AC		74AC		Units	Conditions		
			T <sub>A</sub> = +25°C		T <sub>A</sub> = -55°C to +125°C		T <sub>A</sub> = -40°C to +85°C					
			Typ	Guaranteed Limits								
I <sub>OLD</sub>	†Minimum Dynamic Output Current	5.5				50		75	mA	V <sub>OLD</sub> = 1.65V Max		
I <sub>OHD</sub>		5.5				-50		-75	mA	V <sub>OHD</sub> = 3.85V Min		
I <sub>CC</sub>	Maximum Quiescent Supply Current	5.5		2.0	40.0		20.0	μA	V <sub>IN</sub> = V <sub>CC</sub> or GND			

†Maximum test duration 2.0 ms, one output loaded at a time.

Note: I<sub>IN</sub> and I<sub>CC</sub> @ 3.0V are guaranteed to be less than or equal to the respective limit @ 5.5V V<sub>CC</sub>.

I<sub>CC</sub> for 54AC @ 25°C is identical to 74AC @ 25°C.

## DC Characteristics for 'ACT Family Devices

Symbol	Parameter	V <sub>CC</sub> (V)	74ACT		74ACT		Units	Conditions		
			T <sub>A</sub> = +25°C		T <sub>A</sub> = -40°C to +85°C					
			Typ	Guaranteed Limits						
V <sub>IH</sub>	Minimum High Level Input Voltage	4.5 5.5	1.5 1.5	2.0 2.0	2.0 2.0		V	V <sub>OUT</sub> = 0.1V or V <sub>CC</sub> - 0.1V		
V <sub>IL</sub>	Maximum Low Level Input Voltage	4.5 5.5	1.5 1.5	0.8 0.8	0.8 0.8		V	V <sub>OUT</sub> = 0.1V or V <sub>CC</sub> - 0.1V		
V <sub>OH</sub>	Minimum High Level Output Voltage	4.5 5.5	4.49 5.49	4.4 5.4	4.4 5.4		V	I <sub>OUT</sub> = -50 μA		
		4.5 5.5		3.86 4.86	3.76 4.76		V	*V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IH</sub> I <sub>OH</sub> -24 mA		
V <sub>OL</sub>	Maximum Low Level Output Voltage	4.5 5.5	0.001 0.001	0.1 0.1	0.1 0.1		V	I <sub>OUT</sub> = 50 μA		
		4.5 5.5		0.36 0.36	0.44 0.44		V	*V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IH</sub> I <sub>OL</sub> 24 mA		
I <sub>IN</sub>	Maximum Input Leakage Current	5.5		±0.1	±1.0		μA	V <sub>I</sub> = V <sub>CC</sub> , GND		
I <sub>CCT</sub>	Maximum I <sub>CC</sub> /Input	5.5	0.6		1.5		mA	V <sub>I</sub> = V <sub>CC</sub> - 2.1V		
I <sub>OLD</sub>	†Minimum Dynamic Output Current	5.5			75		mA	V <sub>OLD</sub> = 1.65V Max		
		5.5			-75		mA	V <sub>OHD</sub> = 3.85V Min		
I <sub>CC</sub>	Maximum Quiescent Supply Current	5.5		4.0	40.0		μA	V <sub>IN</sub> = V <sub>CC</sub> or GND		

\*All outputs loaded; thresholds on input associated with output under test.

†Maximum test duration 2.0 ms, one output loaded at a time.

# 54AC/74AC32 • 74ACT32 Quad 2-Input OR Gate

## General Description

The 'AC/'ACT32 contains four, 2-input OR gates.

## Features

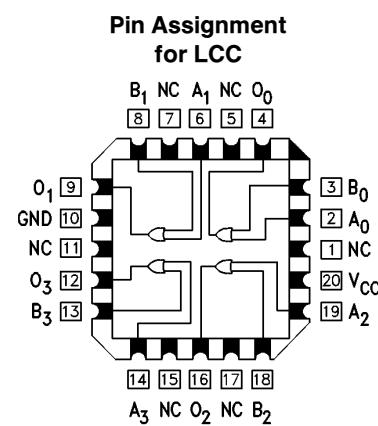
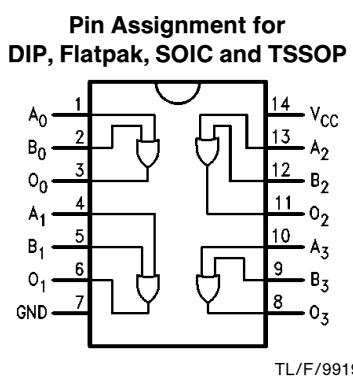
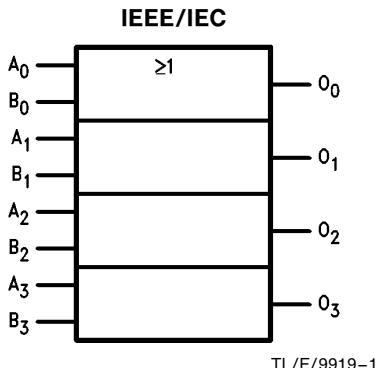
- $I_{CC}$  reduced by 50% on 54AC/74AC only
- Outputs source/sink 24 mA
- 'ACT32 has TTL-compatible inputs
- Standard Military Drawing (SMD)
  - 'AC32: 5962-87614
- For Military 54ACT32, see the 54ACTQ32

Commercial	Military	Package Number	Package Description
74ACT32PC		N14A	14-Lead Molded Dual-In-Line (0.300" Wide)
74ACT32SC (Note 1)		M14A	14-Lead Molded Small Outline (0.150" Wide), JEDEC
74ACT32MTC (Note 1)		MTC14	14-Lead Molded Thin Shrink Small Outline Package, JEDEC
	54ACT32DM (Note 2)	J14A	14-Lead Ceramic Dual-In-Line
	54ACT32FM (Note 2)	W14B	14-Lead Cerpak
	54ACT32LM (Note 2)	E20A	20-Lead Ceramic Leadless Chip Carrier, Type C

**Note 1:** Devices also available in 13" Tape and Reel. Use suffix SCX, SJX, and MTCX.

**Note 2:** Military grade device with environmental and burn-in processing, use suffix DMQB, FMQB and LMQB.

## Logic Symbol



Pin Names	Description
$A_n, B_n$ $O_n$	Inputs Outputs

FACT™ is a trademark of National Semiconductor Corporation.

## Absolute Maximum Rating (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage ( $V_{CC}$ )	$-0.5V$ to $+7.0V$
DC Input Diode Current ( $I_{IK}$ )	
$V_I = -0.5V$	$-20\text{ mA}$
$V_I = V_{CC} + 0.5V$	$+20\text{ mA}$
DC Input Voltage ( $V_I$ )	$-0.5V$ to $V_{CC} + 0.5V$
DC Output Diode Current ( $I_{OK}$ )	
$V_O = -0.5V$	$-20\text{ mA}$
$V_O = V_{CC} + 0.5V$	$+20\text{ mA}$
DC Output Voltage ( $V_O$ )	$-0.5V$ to $V_{CC} + 0.5V$
DC Output Source or Sink Current ( $I_O$ )	$\pm 50\text{ mA}$
DC $V_{CC}$ or Ground Current per Output Pin ( $I_{CC}$ or $I_{GND}$ )	$\pm 50\text{ mA}$
Storage Temperature ( $T_{STG}$ )	$-65^{\circ}\text{C}$ to $+150^{\circ}\text{C}$
Junction Temperature ( $T_J$ )	
CDIP	$175^{\circ}\text{C}$
PDIP	$140^{\circ}\text{C}$

**Note 1:** Absolute maximum ratings are those values beyond which damage to the device may occur. The databook specifications should be met, without exception, to ensure that the system design is reliable over its power supply, temperature, and output/input loading variables. National does not recommend operation of FACT™ circuits outside databook specifications.

## Recommended Operating Conditions

Supply Voltage ( $V_{CC}$ )	$2.0V$ to $6.0V$
'AC	$4.5V$ to $5.5V$
'ACT	
Input Voltage ( $V_I$ )	$0V$ to $V_{CC}$
Output Voltage ( $V_O$ )	$0V$ to $V_{CC}$
Operating Temperature ( $T_A$ )	
74AC/ACT	$-40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$
54AC	$-55^{\circ}\text{C}$ to $+125^{\circ}\text{C}$
Minimum Input Edge Rate ( $\Delta V/\Delta t$ )	
'AC Devices	
$V_{IN}$ from $30\%$ to $70\%$ of $V_{CC}$	
$V_{CC}$ @ $3.3V, 4.5V, 5.5V$	$125\text{ mV/ns}$
'ACT Devices	
$V_{IN}$ from $0.8V$ to $2.0V$	
$V_{CC}$ @ $4.5V, 5.5V$	$125\text{ mV/ns}$

## DC Characteristics for 'AC Family Devices

Symbol	Parameter	$V_{CC}$ (V)	74AC		54AC	74AC	Units	Conditions
			$T_A = +25^{\circ}\text{C}$		$T_A = -55^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	$T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$		
			Typ	Guaranteed Limits				
$V_{IH}$	Minimum High Level Input Voltage	3.0 4.5 5.5	1.5 2.25 2.75	2.1 3.15 3.85	2.1 3.15 3.85	2.1 3.15 3.85	V	$V_{OUT} = 0.1V$ or $V_{CC} - 0.1V$
$V_{IL}$	Maximum Low Level Input Voltage	3.0 4.5 5.5	1.5 2.25 2.75	0.9 1.35 1.65	0.9 1.35 1.65	0.9 1.35 1.65	V	$V_{OUT} = 0.1V$ or $V_{CC} - 0.1V$
$V_{OH}$	Minimum High Level Output Voltage	3.0 4.5 5.5	2.99 4.49 5.49	2.9 4.4 5.4	2.9 4.4 5.4	2.9 4.4 5.4	V	$I_{OUT} = -50\text{ }\mu\text{A}$
		3.0 4.5 5.5		2.56 3.86 4.86	2.4 3.7 4.7	2.46 3.76 4.76	V	$*V_{IN} = V_{IL}$ or $V_{IH}$ $-12\text{ mA}$ $I_{OH} = -24\text{ mA}$ $-24\text{ mA}$
$V_{OL}$	Maximum Low Level Output Voltage	3.0 4.5 5.5	0.002 0.001 0.001	0.1 0.1 0.1	0.1 0.1 0.1	0.1 0.1 0.1	V	$I_{OUT} = 50\text{ }\mu\text{A}$
		3.0 4.5 5.5		0.36 0.36 0.36	0.5 0.5 0.5	0.44 0.44 0.44	V	$*V_{IN} = V_{IL}$ or $V_{IH}$ $12\text{ mA}$ $I_{OL} = 24\text{ mA}$ $24\text{ mA}$
$I_{IN}$	Maximum Input Leakage Current	5.5		$\pm 0.1$	$\pm 1.0$	$\pm 1.0$	$\mu\text{A}$	$V_I = V_{CC}, \text{GND}$

\*All outputs loaded; thresholds on input associated with output under test.

## DC Characteristics for 'AC Family Devices (Continued)

Symbol	Parameter	V <sub>CC</sub> (V)	74AC		54AC		74AC		Units	Conditions		
			T <sub>A</sub> = + 25°C		T <sub>A</sub> = - 55°C to + 125°C		T <sub>A</sub> = - 40°C to + 85°C					
			Typ	Guaranteed Limits								
I <sub>OLD</sub>	†Minimum Dynamic Output Current	5.5			50		75		mA	V <sub>OLD</sub> = 1.65V Max		
I <sub>OHD</sub>		5.5			-50		-75		mA	V <sub>OHD</sub> = 3.85V Min		
I <sub>CC</sub>	Maximum Quiescent Supply Current	5.5		2.0	40.0		20.0		μA	V <sub>IN</sub> = V <sub>CC</sub> or GND		

†Maximum test duration 2.0 ms, one output loaded at a time.

Note: I<sub>IN</sub> and I<sub>CC</sub> @ 3.0V are guaranteed to be less than or equal to the respective limit @ 5.5V V<sub>CC</sub>.

I<sub>CC</sub> for 54AC @ 25°C is identical to 74AC @ 25°C.

## DC Characteristics for 'ACT Family Devices

Symbol	Parameter	V <sub>CC</sub> (V)	74ACT		74ACT		Units	Conditions		
			T <sub>A</sub> = + 25°C		T <sub>A</sub> = - 40°C to + 85°C					
			Typ	Guaranteed Limits						
V <sub>IH</sub>	Minimum High Level Input Voltage	4.5	1.5	2.0	2.0		V	V <sub>OUT</sub> = 0.1V or V <sub>CC</sub> - 0.1V		
		5.5	1.5	2.0	2.0					
V <sub>IL</sub>	Maximum Low Level Input Voltage	4.5	1.5	0.8	0.8		V	V <sub>OUT</sub> = 0.1V or V <sub>CC</sub> - 0.1V		
		5.5	1.5	0.8	0.8					
V <sub>OH</sub>	Minimum High Level Output Voltage	4.5	4.49	4.4	4.4		V	I <sub>OUT</sub> = - 50 μA		
		5.5	5.49	5.4	5.4					
V <sub>OL</sub>		4.5		3.86	3.76		V	*V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IH</sub> I <sub>OH</sub> - 24 mA		
		5.5		4.86	4.76					
I <sub>IN</sub>	Maximum Input Leakage Current	4.5	0.001	0.1	0.1		V	I <sub>OUT</sub> = 50 μA		
		5.5	0.001	0.1	0.1					
I <sub>CCT</sub>	Maximum I <sub>CC</sub> /Input	4.5		0.36	0.44		V	*V <sub>IN</sub> = V <sub>IL</sub> or V <sub>IH</sub> I <sub>OL</sub> 24 mA		
		5.5	0.6	0.36	0.44					
I <sub>OLD</sub>	†Minimum Dynamic Output Current	5.5			75		mA	V <sub>OLD</sub> = 1.65V Max		
		5.5			-75					
I <sub>CC</sub>	Maximum Quiescent Supply Current	5.5		4.0	40.0		μA	V <sub>IN</sub> = V <sub>CC</sub> or GND		

\*All outputs loaded; thresholds on input associated with output under test.

†Maximum test duration 2.0 ms, one output loaded at a time.



**SGS-THOMSON**  
MICROELECTRONICS

**BDX53A/53B/53C**  
**BDX54A/54B/54C**

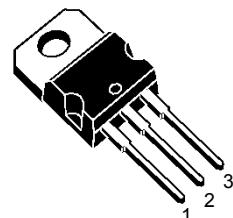
## COMPLEMENTARY SILICON POWER DARLINGTON TRANSISTORS

- BDX53B, BDX53C, BDX54B AND BDX54C ARE SGS-THOMSON PREFERRED SALES TYPES

### DESCRIPTION

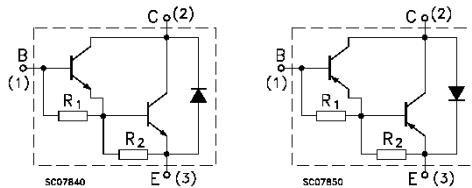
The BDX53A, BDX53B and BDX53C are silicon epitaxial-base NPN power transistors in monolithic Darlington configuration and are mounted in Jedec TO-220 plastic package. They are intended for use in hammer drivers, audio amplifiers and other medium power linear and switching applications.

The complementary PNP types are the BDX54A, BDX54B and BDX54C respectively.



TO-220

### INTERNAL SCHEMATIC DIAGRAM



R<sub>1</sub> Typ. = 10 kΩ      R<sub>2</sub> Typ. = 150 Ω

### ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter					Unit
		NPN	BDX53A	BDX53B	BDX53C	
		PNP	BDX54A	BDX54B	BDX54C	
V <sub>CBO</sub>	Collector-Base Voltage ( $I_E = 0$ )		60	80	100	V
V <sub>CEO</sub>	Collector-Emitter Voltage ( $I_B = 0$ )		60	80	100	V
V <sub>EBO</sub>	Emitter-base Voltage ( $I_C = 0$ )			5		V
I <sub>C</sub>	Collector Current			8		A
I <sub>CM</sub>	Collector Peak Current (repetitive)			12		A
I <sub>B</sub>	Base Current			0.2		A
P <sub>tot</sub>	Total Dissipation at $T_c \leq 25^\circ\text{C}$			60		W
T <sub>stg</sub>	Storage Temperature			-65 to 150		°C
T <sub>j</sub>	Max. Operating Junction Temperature			150		°C

October 1995

## BDX53A/53B/53C-BDX54A/54B/54C

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### THERMAL DATA

$R_{thj\text{-case}}$	Thermal Resistance Junction-case	Max	2.08	$^{\circ}\text{C}/\text{W}$
$R_{thj\text{-amb}}$	Thermal Resistance Junction-ambient	Max	70	$^{\circ}\text{C}/\text{W}$

ELECTRICAL CHARACTERISTICS ( $T_{\text{case}} = 25^{\circ}\text{C}$  unless otherwise specified)

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
$I_{\text{CBO}}$	Collector Cut-off Current ( $I_E = 0$ )	for BDX53A/54A	$V_{\text{CB}} = 60\text{ V}$			0.2	mA
		for BDX53B/54B	$V_{\text{CB}} = 80\text{ V}$			0.2	mA
		for BDX53C/54C	$V_{\text{CB}} = 100\text{ V}$			0.2	mA
$I_{\text{CEO}}$	Collector Cut-off Current ( $I_B = 0$ )	for BDX53A/54A	$V_{\text{CB}} = 30\text{ V}$			0.5	mA
		for BDX53B/54B	$V_{\text{CB}} = 40\text{ V}$			0.5	mA
		for BDX53C/54C	$V_{\text{CB}} = 50\text{ V}$			0.5	mA
$I_{\text{EBO}}$	Emitter Cut-off Current ( $I_C = 0$ )	$V_{\text{EB}} = 5\text{ V}$				2	mA
$V_{\text{CEO(sus)}}^*$	Collector-Emitter Sustaining Voltage ( $I_B = 0$ )	$I_C = 100\text{ mA}$	for BDX53A/54A	60			V
			for BDX53B/53B	80			V
			for BDX53C/54C	100			V
$V_{\text{CE(sat)}}^*$	Collector-emitter Saturation Voltage	$I_C = 3\text{ A}$	$I_B = 12\text{ mA}$			2	V
$V_{\text{BE(sat)}}^*$	Base-emitter Saturation Voltage	$I_C = 3\text{ A}$	$I_B = 12\text{ mA}$			2.5	V
$h_{\text{FE}}^*$	DC Current Gain	$I_C = 3\text{ A}$	$V_{\text{CE}} = 3\text{ V}$	750			
$V_F^*$	Parallel-diode Forward Voltage	$I_F = 3\text{ A}$			1.8	2.5	V
		$I_F = 8\text{ A}$			2.5		V

\* Pulsed: Pulse duration = 300  $\mu\text{s}$ , duty cycle 1.5 %  
For PNP types voltage and current values are negative.

## 18-pin Flash/EEPROM 8-Bit Microcontrollers

### Devices Included in this Data Sheet:

- PIC16F83
- PIC16F84
- PIC16CR83
- PIC16CR84
- Extended voltage range devices available (PIC16LF8X, PIC16LCR8X)

### High Performance RISC CPU Features:

- Only 35 single word instructions to learn
- All instructions single cycle except for program branches which are two-cycle
- Operating speed: DC - 10 MHz clock input  
DC - 400 ns instruction cycle

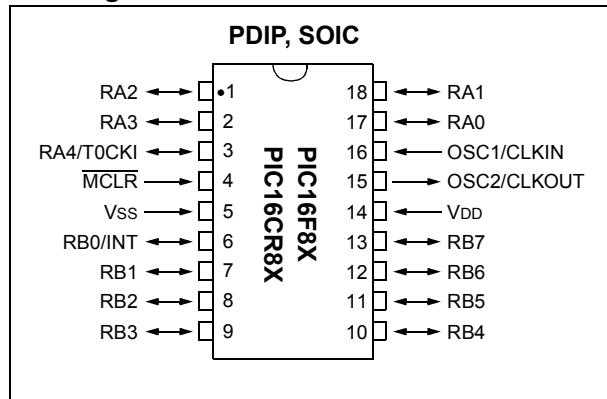
Device	Program Memory (words)	Data RAM (bytes)	Data EEPROM (bytes)	Max. Freq (MHz)
PIC16F83	512 Flash	36	64	10
PIC16F84	1 K Flash	68	64	10
PIC16CR83	512 ROM	36	64	10
PIC16CR84	1 K ROM	68	64	10

- 14-bit wide instructions
- 8-bit wide data path
- 15 special function hardware registers
- Eight-level deep hardware stack
- Direct, indirect and relative addressing modes
- Four interrupt sources:
  - External RB0/INT pin
  - TMR0 timer overflow
  - PORTB<7:4> interrupt on change
  - Data EEPROM write complete
- 1000 erase/write cycles Flash program memory
- 10,000,000 erase/write cycles EEPROM data memory
- EEPROM Data Retention > 40 years

### Peripheral Features:

- 13 I/O pins with individual direction control
- High current sink/source for direct LED drive
  - 25 mA sink max. per pin
  - 20 mA source max. per pin
- TMR0: 8-bit timer/counter with 8-bit programmable prescaler

### Pin Diagrams



### Special Microcontroller Features:

- In-Circuit Serial Programming (ICSP™) - via two pins (ROM devices support only Data EEPROM programming)
- Power-on Reset (POR)
- Power-up Timer (PWRT)
- Oscillator Start-up Timer (OST)
- Watchdog Timer (WDT) with its own on-chip RC oscillator for reliable operation
- Code-protection
- Power saving SLEEP mode
- Selectable oscillator options

### CMOS Flash/EEPROM Technology:

- Low-power, high-speed technology
- Fully static design
- Wide operating voltage range:
  - Commercial: 2.0V to 6.0V
  - Industrial: 2.0V to 6.0V
- Low power consumption:
  - < 2 mA typical @ 5V, 4 MHz
  - 15 µA typical @ 2V, 32 kHz
  - < 1 µA typical standby current @ 2V

# PIC16F8X

PIC16CXX devices contain an 8-bit ALU and working register. The ALU is a general purpose arithmetic unit. It performs arithmetic and Boolean functions between data in the working register and any register file.

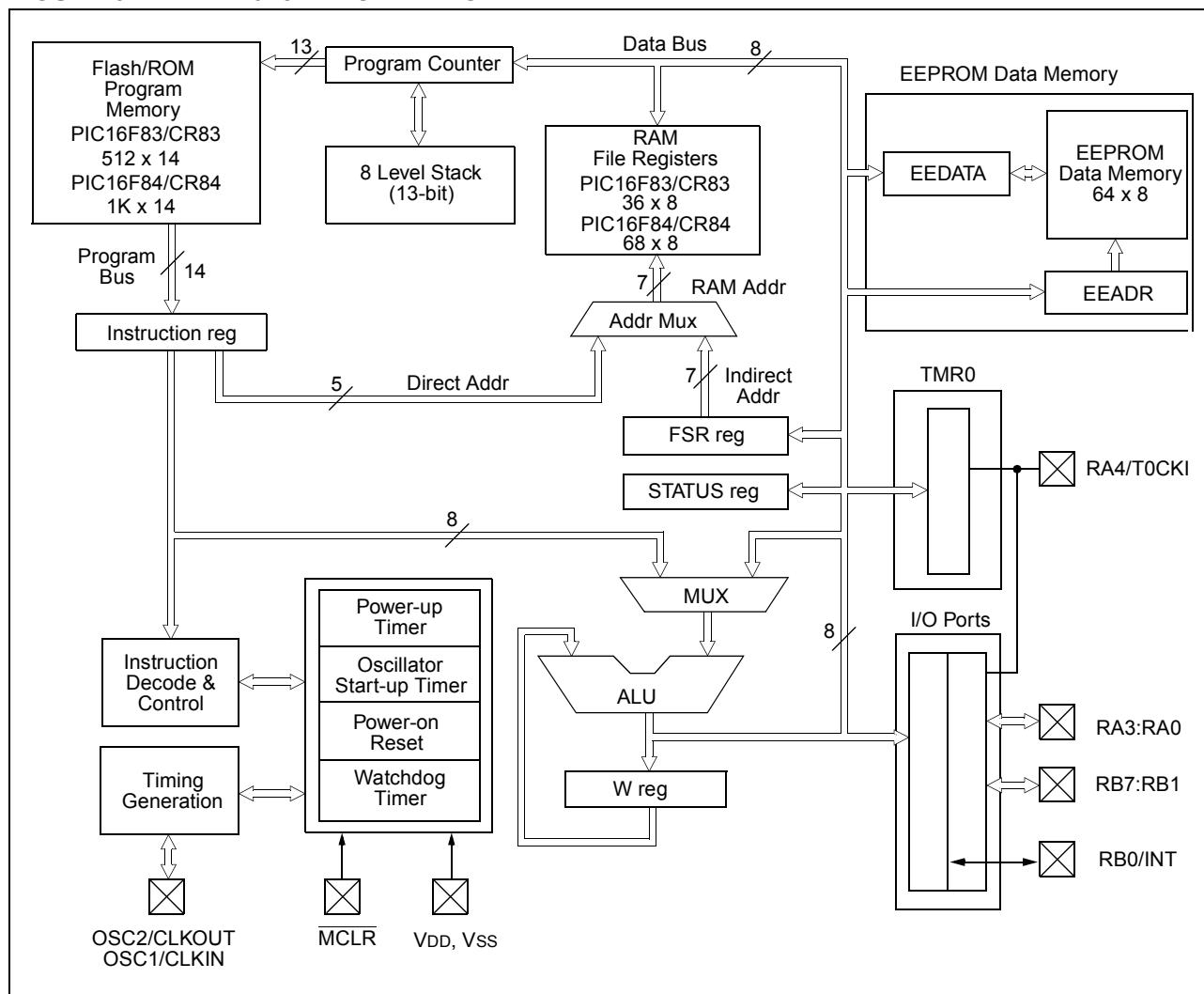
The ALU is 8-bits wide and capable of addition, subtraction, shift and logical operations. Unless otherwise mentioned, arithmetic operations are two's complement in nature. In two-operand instructions, typically one operand is the working register (W register), and the other operand is a file register or an immediate constant. In single operand instructions, the operand is either the W register or a file register.

The W register is an 8-bit working register used for ALU operations. It is not an addressable register.

Depending on the instruction executed, the ALU may affect the values of the Carry (C), Digit Carry (DC), and Zero (Z) bits in the STATUS register. The C and DC bits operate as a borrow and digit borrow out bit, respectively, in subtraction. See the SUBLW and SUBWF instructions for examples.

A simplified block diagram for the PIC16F8X is shown in Figure 3-1, its corresponding pin description is shown in Table 3-1.

**FIGURE 3-1: PIC16F8X BLOCK DIAGRAM**



**TABLE 3-1 PIC16F8X PINOUT DESCRIPTION**

Pin Name	DIP No.	SOIC No.	I/O/P Type	Buffer Type	Description
OSC1/CLKIN	16	16	I	ST/CMOS <sup>(3)</sup>	Oscillator crystal input/external clock source input.
OSC2/CLKOUT	15	15	O	—	Oscillator crystal output. Connects to crystal or resonator in crystal oscillator mode. In RC mode, OSC2 pin outputs CLKOUT which has 1/4 the frequency of OSC1, and denotes the instruction cycle rate.
MCLR	4	4	I/P	ST	Master clear (reset) input/programming voltage input. This pin is an active low reset to the device.
RA0	17	17	I/O	TTL	PORTA is a bi-directional I/O port.  Can also be selected to be the clock input to the TMR0 timer/counter. Output is open drain type.
RA1	18	18	I/O	TTL	
RA2	1	1	I/O	TTL	
RA3	2	2	I/O	TTL	
RA4/T0CKI	3	3	I/O	ST	
RB0/INT	6	6	I/O	TTL/ST <sup>(1)</sup>	PORTB is a bi-directional I/O port. PORTB can be software programmed for internal weak pull-up on all inputs. RB0/INT can also be selected as an external interrupt pin.
RB1	7	7	I/O	TTL	
RB2	8	8	I/O	TTL	
RB3	9	9	I/O	TTL	
RB4	10	10	I/O	TTL	Interrupt on change pin.
RB5	11	11	I/O	TTL	Interrupt on change pin.
RB6	12	12	I/O	TTL/ST <sup>(2)</sup>	Interrupt on change pin. Serial programming clock.
RB7	13	13	I/O	TTL/ST <sup>(2)</sup>	Interrupt on change pin. Serial programming data.
Vss	5	5	P	—	Ground reference for logic and I/O pins.
Vdd	14	14	P	—	Positive supply for logic and I/O pins.

Legend: I = input      O = output      I/O = Input/Output      P = power  
           — = Not used      TTL = TTL input      ST = Schmitt Trigger input

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in serial programming mode.

3: This buffer is a Schmitt Trigger input when configured in RC oscillator mode and a CMOS input otherwise.

# PIC16F8X

## 3.1 Clocking Scheme/Instruction Cycle

The clock input (from OSC1) is internally divided by four to generate four non-overlapping quadrature clocks namely Q1, Q2, Q3 and Q4. Internally, the program counter (PC) is incremented every Q1, the instruction is fetched from the program memory and latched into the instruction register in Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow is shown in Figure 3-2.

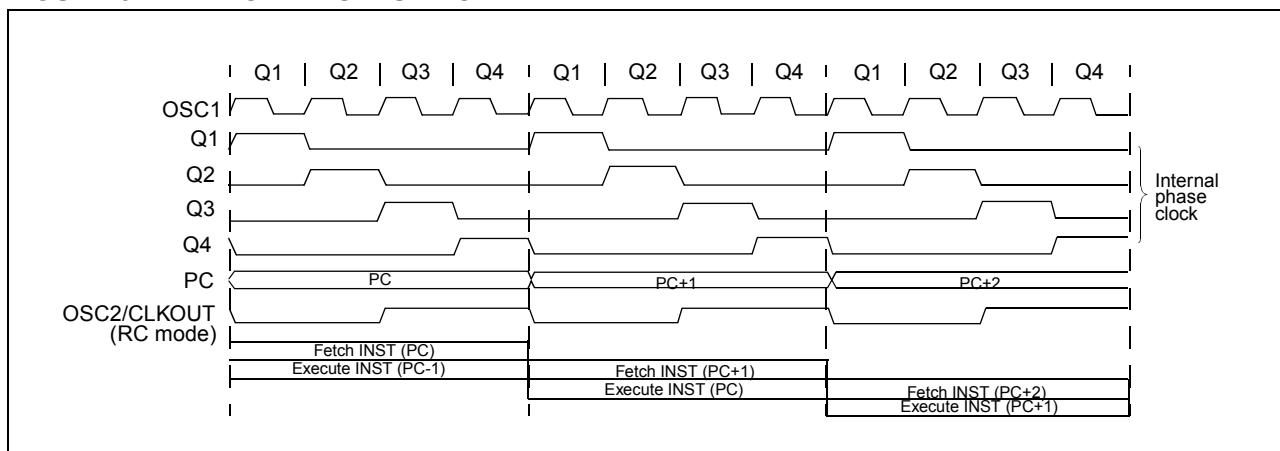
## 3.2 Instruction Flow/Pipelining

An “Instruction Cycle” consists of four Q cycles (Q1, Q2, Q3 and Q4). The instruction fetch and execute are pipelined such that fetch takes one instruction cycle while decode and execute takes another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (e.g., GOTO) then two cycles are required to complete the instruction (Example 3-1).

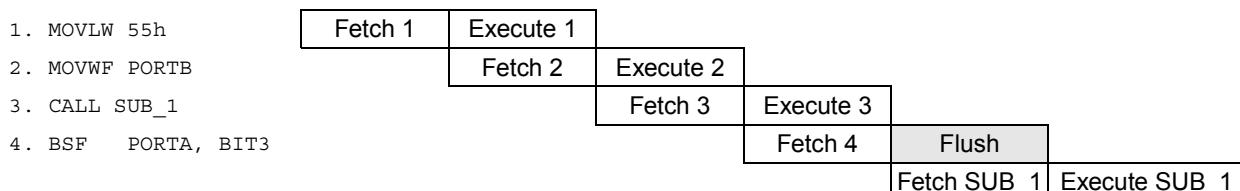
A fetch cycle begins with the Program Counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the “Instruction Register” in cycle Q1. This instruction is then decoded and executed during the Q2, Q3, and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

FIGURE 3-2: CLOCK/INSTRUCTION CYCLE

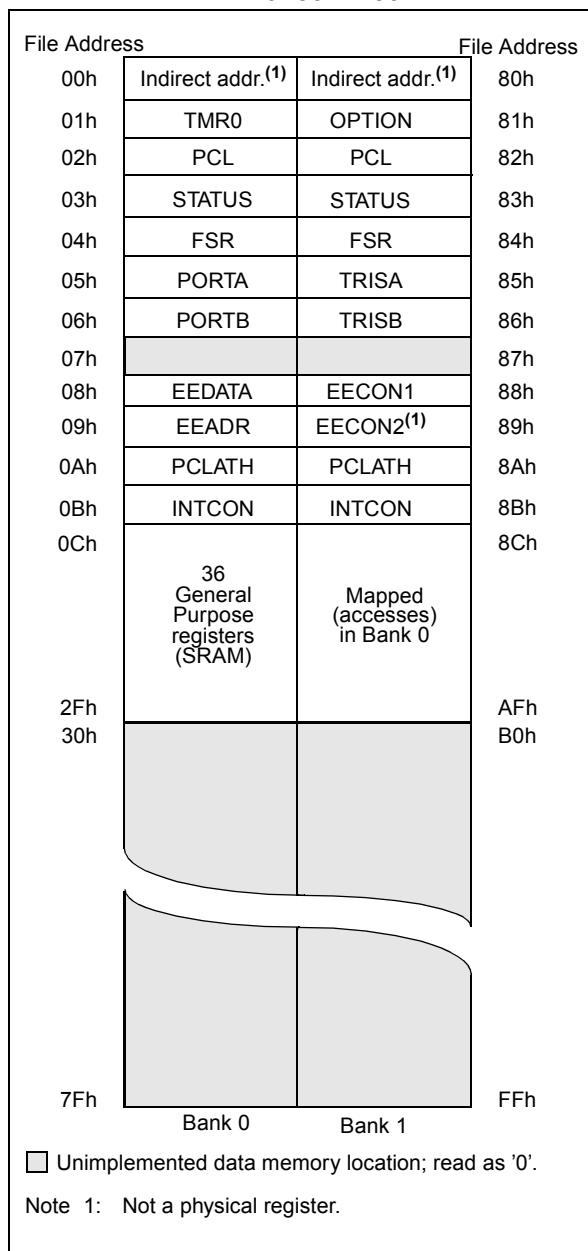


EXAMPLE 3-1: INSTRUCTION PIPELINE FLOW

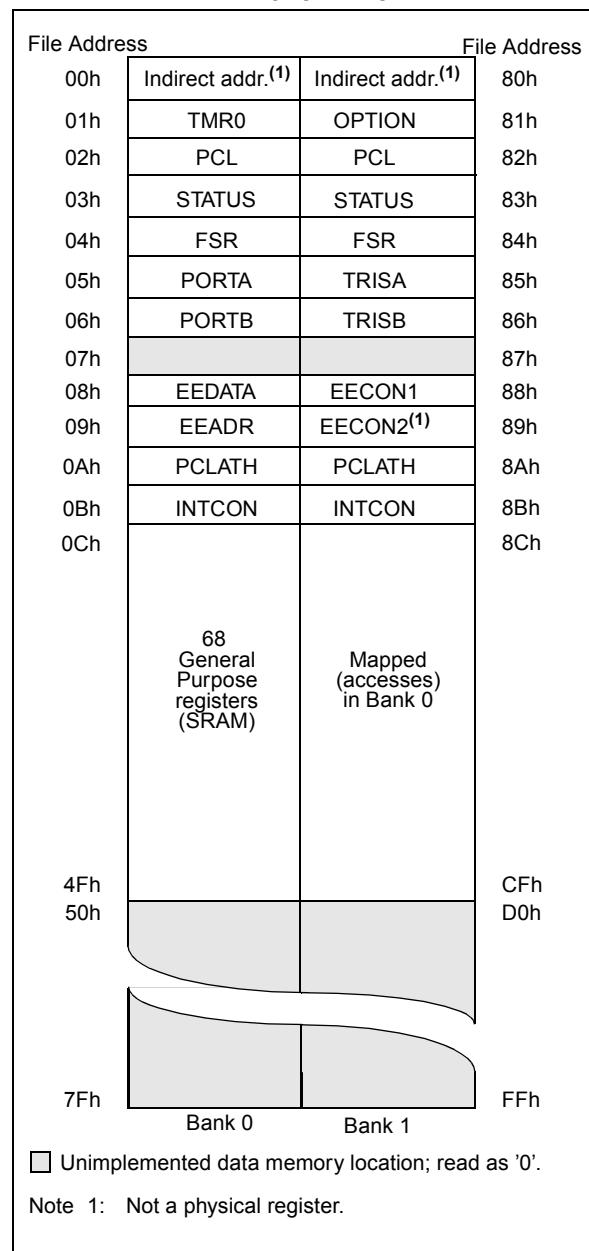


All instructions are single cycle, except for any program branches. These take two cycles since the fetch instruction is “flushed” from the pipeline while the new instruction is being fetched and then executed.

**FIGURE 4-1: REGISTER FILE MAP - PIC16F83/CR83**



**FIGURE 4-2: REGISTER FILE MAP - PIC16F84/CR84**



# PIC16F8X

---

**TABLE 4-1 REGISTER FILE SUMMARY**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets (Note3)				
<b>Bank 0</b>															
00h	INDF	Uses contents of FSR to address data memory (not a physical register)							-----	-----	-----				
01h	TMR0	8-bit real-time clock/counter							xxxx xxxx	uuuu uuuu	uuuu uuuu				
02h	PCL	Low order 8 bits of the Program Counter (PC)							0000 0000	0000 0000	0000 0000				
03h	STATUS (2)	IRP	RP1	RP0	TO	PD	Z	DC	C	0001 1xxx	000q quuu				
04h	FSR	Indirect data memory address pointer 0							xxxx xxxx	uuuu uuuu	uuuu uuuu				
05h	PORTA	—	—	—	RA4/T0CKI	RA3	RA2	RA1	RA0	---x xxxx	---u uuuu				
06h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0/INT	xxxx xxxx	uuuu uuuu				
07h		Unimplemented location, read as '0'							-----	-----	-----				
08h	EEDATA	EEPROM data register							xxxx xxxx	uuuu uuuu	uuuu uuuu				
09h	EEADR	EEPROM address register							xxxx xxxx	uuuu uuuu	uuuu uuuu				
0Ah	PCLATH	—	—	—	Write buffer for upper 5 bits of the PC (1)				---0 0000	---0 0000	---0 0000				
0Bh	INTCON	GIE	EEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 000x	0000 000u				
<b>Bank 1</b>															
80h	INDF	Uses contents of FSR to address data memory (not a physical register)							-----	-----	-----				
81h	OPTION_REG	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111				
82h	PCL	Low order 8 bits of Program Counter (PC)							0000 0000	0000 0000	0000 0000				
83h	STATUS (2)	IRP	RP1	RP0	TO	PD	Z	DC	C	0001 1xxx	000q quuu				
84h	FSR	Indirect data memory address pointer 0							xxxx xxxx	uuuu uuuu	uuuu uuuu				
85h	TRISA	—	—	—	PORTA data direction register				---1 1111	---1 1111	---1 1111				
86h	TRISB	PORTB data direction register							1111 1111	1111 1111	1111 1111				
87h		Unimplemented location, read as '0'							-----	-----	-----				
88h	EECON1	—	—	—	EEIF	WRERR	WREN	WR	RD	---0 x000	---0 q000				
89h	EECON2	EEPROM control register 2 (not a physical register)							-----	-----	-----				
0Ah	PCLATH	—	—	—	Write buffer for upper 5 bits of the PC (1)				---0 0000	---0 0000	---0 0000				
0Bh	INTCON	GIE	EEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 000x	0000 000u				

Legend: x = unknown, u = unchanged. - = unimplemented read as '0', q = value depends on condition.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a slave register for PC<12:8>. The contents of PCLATH can be transferred to the upper byte of the program counter, but the contents of PC<12:8> is never transferred to PCLATH.

2: The TO and PD status bits in the STATUS register are not affected by a MCLR reset.

3: Other (non power-up) resets include: external reset through MCLR and the Watchdog Timer Reset.

## 5.0 I/O PORTS

The PIC16F8X has two ports, PORTA and PORTB. Some port pins are multiplexed with an alternate function for other features on the device.

### 5.1 PORTA and TRISA Registers

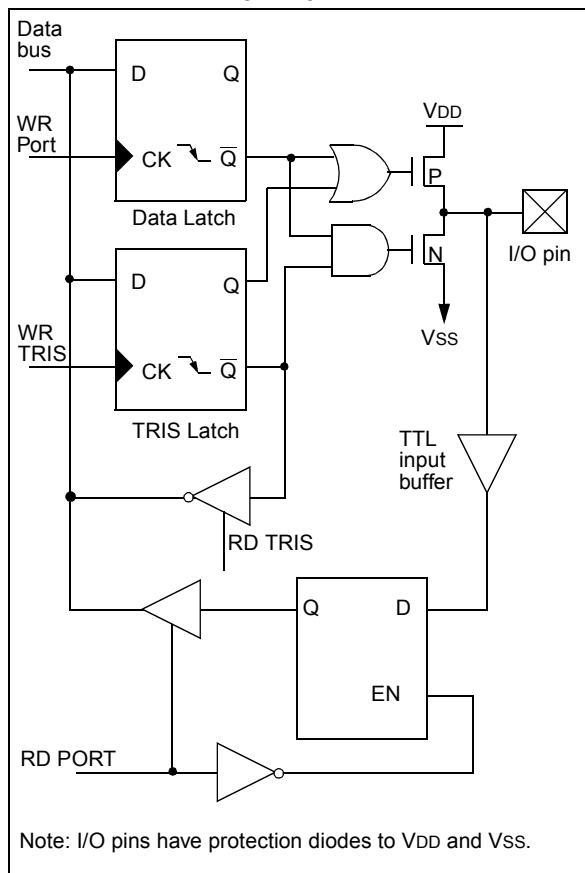
PORTA is a 5-bit wide latch. RA4 is a Schmitt Trigger input and an open drain output. All other RA port pins have TTL input levels and full CMOS output drivers. All pins have data direction bits (TRIS registers) which can configure these pins as output or input.

Setting a TRISA bit (=1) will make the corresponding PORTA pin an input, i.e., put the corresponding output driver in a hi-impedance mode. Clearing a TRISA bit (=0) will make the corresponding PORTA pin an output, i.e., put the contents of the output latch on the selected pin.

Reading the PORTA register reads the status of the pins whereas writing to it will write to the port latch. All write operations are read-modify-write operations. So a write to a port implies that the port pins are first read, then this value is modified and written to the port data latch.

The RA4 pin is multiplexed with the TMR0 clock input.

**FIGURE 5-1: BLOCK DIAGRAM OF PINS RA3:RA0**



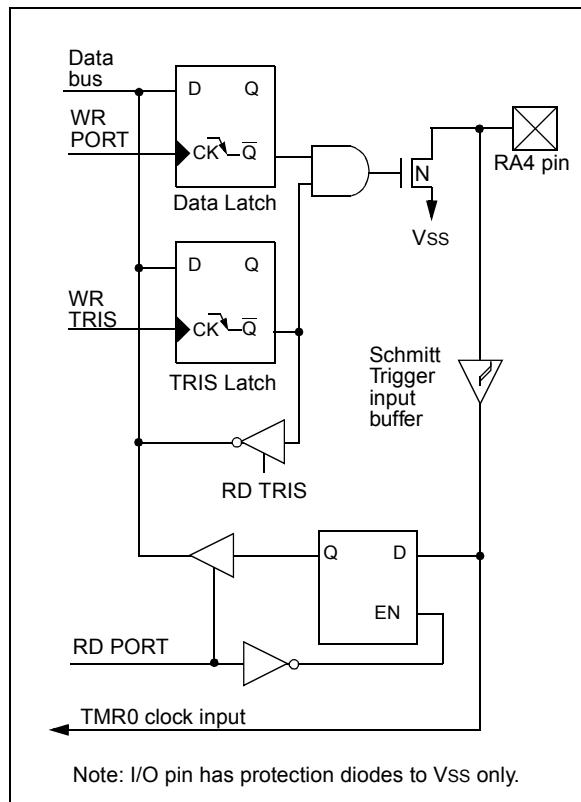
### EXAMPLE 5-1: INITIALIZING PORTA

```

CLRF    PORTA      ; Initialize PORTA by
                     ; setting output
                     ; data latches
BSF     STATUS, RP0   ; Select Bank 1
MOVLW  0x0F      ; Value used to
                     ; initialize data
                     ; direction
MOVWF  TRISA      ; Set RA<3:0> as inputs
                     ; RA4 as outputs
                     ; TRISA<7:5> are always
                     ; read as '0'.

```

**FIGURE 5-2: BLOCK DIAGRAM OF PIN RA4**



# PIC16F8X

---

**TABLE 5-1 PORTA FUNCTIONS**

Name	Bit0	Buffer Type	Function
RA0	bit0	TTL	Input/output
RA1	bit1	TTL	Input/output
RA2	bit2	TTL	Input/output
RA3	bit3	TTL	Input/output
RA4/T0CKI	bit4	ST	Input/output or external clock input for TMR0. Output is open drain type.

Legend: TTL = TTL input, ST = Schmitt Trigger input

**TABLE 5-2 SUMMARY OF REGISTERS ASSOCIATED WITH PORTA**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets
05h	PORTA	—	—	—	RA4/T0CKI	RA3	RA2	RA1	RA0	---x xxxx	---u uuuu
85h	TRISA	—	—	—	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	---1 1111	---1 1111

Legend: x = unknown, u = unchanged, - = unimplemented read as '0'. Shaded cells are unimplemented, read as '0'

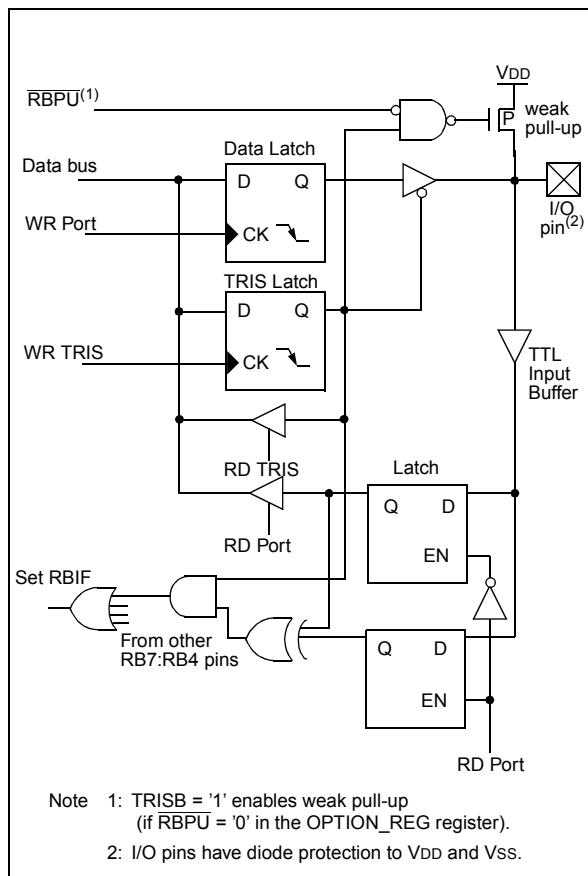
## 5.2 PORTB and TRISB Registers

PORTB is an 8-bit wide bi-directional port. The corresponding data direction register is TRISB. A '1' on any bit in the TRISB register puts the corresponding output driver in a hi-impedance mode. A '0' on any bit in the TRISB register puts the contents of the output latch on the selected pin(s).

Each of the PORTB pins have a weak internal pull-up. A single control bit can turn on all the pull-ups. This is done by clearing the RBPU (OPTION\_REG<7>) bit. The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

Four of PORTB's pins, RB7:RB4, have an interrupt on change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB4 pin configured as an output is excluded from the interrupt on change comparison). The pins value in input mode are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of the pins are OR'd together to generate the RB port change interrupt.

**FIGURE 5-3: BLOCK DIAGRAM OF PINS RB7:RB4**



This interrupt can wake the device from SLEEP. The user, in the interrupt service routine, can clear the interrupt in the following manner:

- Read (or write) PORTB. This will end the mismatch condition.
- Clear flag bit RBIF.

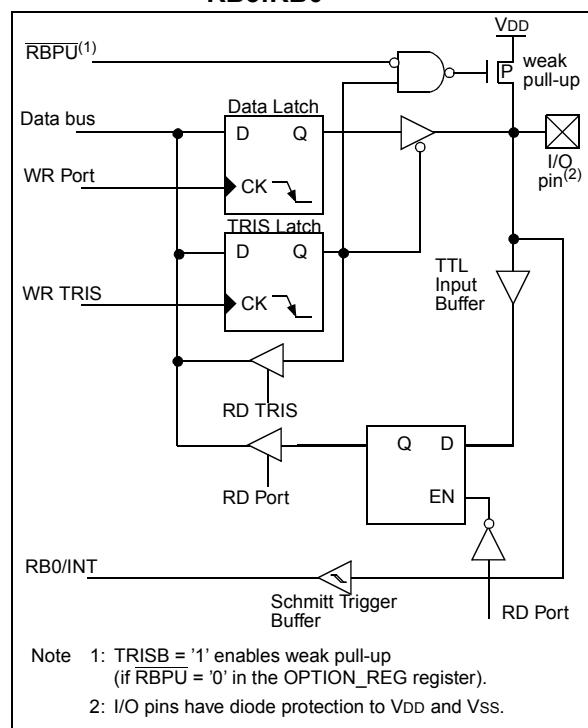
A mismatch condition will continue to set the RBIF bit. Reading PORTB will end the mismatch condition, and allow the RBIF bit to be cleared.

This interrupt on mismatch feature, together with software configurable pull-ups on these four pins allow easy interface to a key pad and make it possible for wake-up on key-depression (see AN552 in the Embedded Control Handbook).

**Note 1:** For a change on the I/O pin to be recognized, the pulse width must be at least  $T_{CY}$  (4/fosc) wide.

The interrupt on change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt on change feature. Polling of PORTB is not recommended while using the interrupt on change feature.

**FIGURE 5-4: BLOCK DIAGRAM OF PINS RB3:RB0**



# PIC16F8X

---

## EXAMPLE 5-1: INITIALIZING PORTB

```

CLRF    PORTB      ; Initialize PORTB by
                ; setting output
                ; data latches
BSF     STATUS, RP0 ; Select Bank 1
MOVLW   0xCF       ; Value used to
                ; initialize data
                ; direction
MOVWF   TRISB      ; Set RB<3:0> as inputs
                ; RB<5:4> as outputs
                ; RB<7:6> as inputs

```

**TABLE 5-3 PORTB FUNCTIONS**

Name	Bit	Buffer Type	I/O Consistency Function
RB0/INT	bit0	TTL/ST <sup>(1)</sup>	Input/output pin or external interrupt input. Internal software programmable weak pull-up.
RB1	bit1	TTL	Input/output pin. Internal software programmable weak pull-up.
RB2	bit2	TTL	Input/output pin. Internal software programmable weak pull-up.
RB3	bit3	TTL	Input/output pin. Internal software programmable weak pull-up.
RB4	bit4	TTL	Input/output pin (with interrupt on change). Internal software programmable weak pull-up.
RB5	bit5	TTL	Input/output pin (with interrupt on change). Internal software programmable weak pull-up.
RB6	bit6	TTL/ST <sup>(2)</sup>	Input/output pin (with interrupt on change). Internal software programmable weak pull-up. Serial programming clock.
RB7	bit7	TTL/ST <sup>(2)</sup>	Input/output pin (with interrupt on change). Internal software programmable weak pull-up. Serial programming data.

Legend: TTL = TTL input, ST = Schmitt Trigger.

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in serial programming mode.

**TABLE 5-4 SUMMARY OF REGISTERS ASSOCIATED WITH PORTB**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets
06h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0/INT	xxxx xxxx	uuuu uuuu
86h	TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111	1111 1111
81h	OPTION_REG	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged. Shaded cells are not used by PORTB.

# PIC16F8X

## 8.1 Configuration Bits

The configuration bits can be programmed (read as '0') or left unprogrammed (read as '1') to select various device configurations. These bits are mapped in program memory location 2007h.

Address 2007h is beyond the user program memory space and it belongs to the special test/configuration memory space (2000h - 3FFFh). This space can only be accessed during programming.

To find out how to program the PIC16C84, refer to *PIC16C84 EEPROM Memory Programming Specification* (DS30189).

**FIGURE 8-1: CONFIGURATION WORD - PIC16CR83 AND PIC16CR84**

R-u	R-u	R-u	R-u	R-u	R-u	R/P-u	R-u	R-u	R-u	R-u	R-u	R-u	R-u
CP	CP	CP	CP	CP	CP	DP	CP	CP	CP	<del>PWRTE</del>	WDTE	FOSC1	FOSC0
bit13													bit0
													<div style="border: 1px solid black; padding: 5px;"><p>R = Readable bit P = Programmable bit - n = Value at POR reset u = unchanged</p></div>
<p>bit 13:8 <b>CP</b>: Program Memory Code Protection bit 1 = Code protection off 0 = Program memory is code protected</p>													
<p>bit 7 <b>DP</b>: Data Memory Code Protection bit 1 = Code protection off 0 = Data memory is code protected</p>													
<p>bit 6:4 <b>CP</b>: Program Memory Code Protection bit 1 = Code protection off 0 = Program memory is code protected</p>													
<p>bit 3 <b>PWRTE</b>: Power-up Timer Enable bit 1 = Power-up timer is disabled 0 = Power-up timer is enabled</p>													
<p>bit 2 <b>WDTE</b>: Watchdog Timer Enable bit 1 = WDT enabled 0 = WDT disabled</p>													
<p>bit 1:0 <b>FOSC1:FOSC0</b>: Oscillator Selection bits 11 = RC oscillator 10 = HS oscillator 01 = XT oscillator 00 = LP oscillator</p>													

**FIGURE 8-2: CONFIGURATION WORD - PIC16F83 AND PIC16F84**

R/P-u	R/P-u												
CP	PWRTE	WDTE	FOSC1	FOSC0									
bit13												bit0	
												R = Readable bit P = Programmable bit - n = Value at POR reset u = unchanged	

bit 13:4 **CP**: Code Protection bit  
 1 = Code protection off  
 0 = All memory is code protected

bit 3 **PWRTE**: Power-up Timer Enable bit  
 1 = Power-up timer is disabled  
 0 = Power-up timer is enabled

bit 2 **WDTE**: Watchdog Timer Enable bit  
 1 = WDT enabled  
 0 = WDT disabled

bit 1:0 **FOSC1:FOSC0**: Oscillator Selection bits  
 11 = RC oscillator  
 10 = HS oscillator  
 01 = XT oscillator  
 00 = LP oscillator

## 8.2 Oscillator Configurations

### 8.2.1 OSCILLATOR TYPES

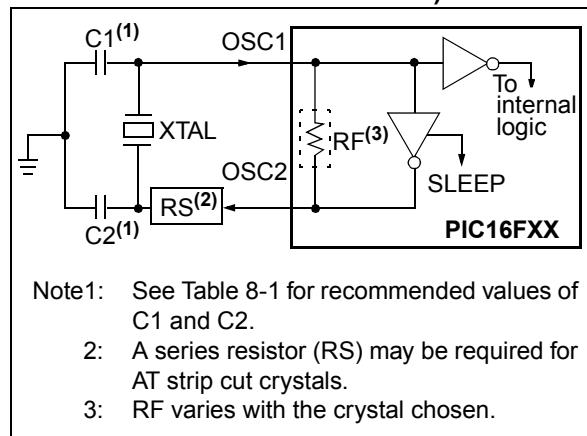
The PIC16F8X can be operated in four different oscillator modes. The user can program two configuration bits (FOSC1 and FOSC0) to select one of these four modes:

- LP Low Power Crystal
- XT Crystal/Resonator
- HS High Speed Crystal/Resonator
- RC Resistor/Capacitor

### 8.2.2 CRYSTAL OSCILLATOR / CERAMIC RESONATORS

In XT, LP or HS modes a crystal or ceramic resonator is connected to the OSC1/CLKIN and OSC2/CLKOUT pins to establish oscillation (Figure 8-3).

**FIGURE 8-3: CRYSTAL/CERAMIC RESONATOR OPERATION (HS, XT OR LP OSC CONFIGURATION)**



The PIC16F8X oscillator design requires the use of a parallel cut crystal. Use of a series cut crystal may give a frequency out of the crystal manufacturers specifications. When in XT, LP or HS modes, the device can have an external clock source to drive the OSC1/CLKIN pin (Figure 8-4).

# PIC16F8X

## 8.9 Interrupts

The PIC16F8X has 4 sources of interrupt:

- External interrupt RB0/INT pin
- TMR0 overflow interrupt
- PORTB change interrupts (pins RB7:RB4)
- Data EEPROM write complete interrupt

The interrupt control register (INTCON) records individual interrupt requests in flag bits. It also contains the individual and global interrupt enable bits.

The global interrupt enable bit, GIE (INTCON<7>) enables (if set) all un-masked interrupts or disables (if cleared) all interrupts. Individual interrupts can be disabled through their corresponding enable bits in INTCON register. Bit GIE is cleared on reset.

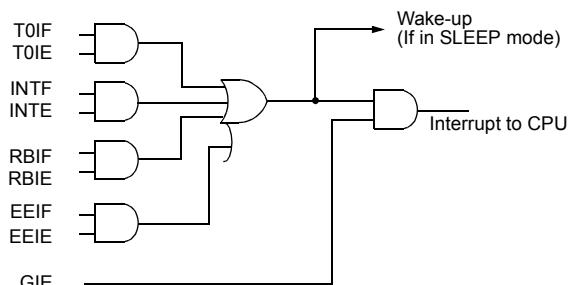
The “return from interrupt” instruction, RETFIE, exits interrupt routine as well as sets the GIE bit, which re-enables interrupts.

**FIGURE 8-16: INTERRUPT LOGIC**

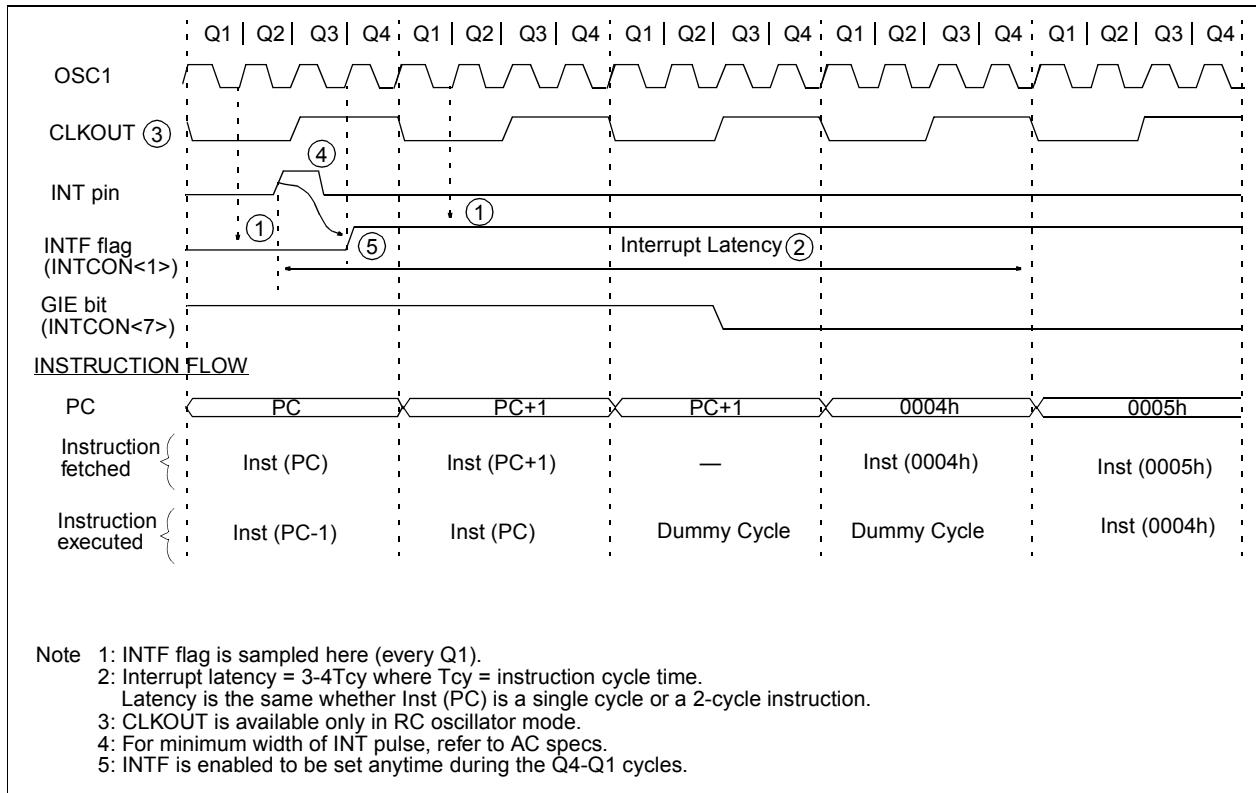
The RB0/INT pin interrupt, the RB port change interrupt and the TMR0 overflow interrupt flags are contained in the INTCON register.

When an interrupt is responded to; the GIE bit is cleared to disable any further interrupt, the return address is pushed onto the stack and the PC is loaded with 0004h. For external interrupt events, such as the RB0/INT pin or PORTB change interrupt, the interrupt latency will be three to four instruction cycles. The exact latency depends when the interrupt event occurs (Figure 8-17). The latency is the same for both one and two cycle instructions. Once in the interrupt service routine the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid infinite interrupt requests.

**Note 1:** Individual interrupt flag bits are set regardless of the status of their corresponding mask bit or the GIE bit.



**FIGURE 8-17: INT PIN INTERRUPT TIMING**



### 8.9.1 INT INTERRUPT

External interrupt on RB0/INT pin is edge triggered: either rising if INTEDG bit (OPTION\_REG<6>) is set, or falling, if INTEDG bit is clear. When a valid edge appears on the RB0/INT pin, the INTF bit (INTCON<1>) is set. This interrupt can be disabled by clearing control bit INTE (INTCON<4>). Flag bit INTF must be cleared in software via the interrupt service routine before re-enabling this interrupt. The INT interrupt can wake the processor from SLEEP (Section 8.12) only if the INTE bit was set prior to going into SLEEP. The status of the GIE bit decides whether the processor branches to the interrupt vector following wake-up.

### 8.9.2 TMR0 INTERRUPT

An overflow (FFh → 00h) in TMR0 will set flag bit T0IF (INTCON<2>). The interrupt can be enabled/disabled by setting/clearing enable bit T0IE (INTCON<5>) (Section 6.0).

### 8.9.3 PORT RB INTERRUPT

An input change on PORTB<7:4> sets flag bit RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit RBIE (INTCON<3>) (Section 5.2).

**Note 1:** For a change on the I/O pin to be recognized, the pulse width must be at least TCY wide.

## 9.0 INSTRUCTION SET SUMMARY

Each PIC16CXX instruction is a 14-bit word divided into an OPCODE which specifies the instruction type and one or more operands which further specify the operation of the instruction. The PIC16CXX instruction set summary in Table 9-2 lists **byte-oriented**, **bit-oriented**, and **literal and control** operations. Table 9-1 shows the opcode field descriptions.

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator which selects the number of the bit affected by the operation, while 'f' represents the number of the file in which the bit is located.

For **literal and control** operations, 'k' represents an eight or eleven bit constant or literal value.

**TABLE 9-1 OPCODE FIELD DESCRIPTIONS**

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1) The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1
label	Label name
TOS	Top of Stack
PC	Program Counter
PCLATH	Program Counter High Latch
GIE	Global Interrupt Enable bit
WDT	Watchdog Timer/Counter
TO	Time-out bit
PD	Power-down bit
dest	Destination either the W register or the specified register file location
[ ]	Options
( )	Contents
→	Assigned to
< >	Register bit field
∈	In the set of
<i>italics</i>	User defined term (font is courier)

The instruction set is highly orthogonal and is grouped into three basic categories:

- **Byte-oriented** operations
- **Bit-oriented** operations
- **Literal and control** operations

All instructions are executed within one single instruction cycle, unless a conditional test is true or the program counter is changed as a result of an instruction. In this case, the execution takes two instruction cycles with the second cycle executed as a NOP. One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1 µs. If a conditional test is true or the program counter is changed as a result of an instruction, the instruction execution time is 2 µs.

Table 9-2 lists the instructions recognized by the MPASM assembler.

Figure 9-1 shows the general formats that the instructions can have.

**Note:** To maintain upward compatibility with future PIC16CXX products, do not use the OPTION and TRIS instructions.

All examples use the following format to represent a hexadecimal number:

0xhh

where h signifies a hexadecimal digit.

**FIGURE 9-1: GENERAL FORMAT FOR INSTRUCTIONS**

Byte-oriented file register operations			
13	8	7	6 0
OPCODE	d	f (FILE #)	
d = 0 for destination W			
d = 1 for destination f			
f = 7-bit file register address			

Bit-oriented file register operations			
13	10	9	7 6 0
OPCODE	b (BIT #)	f (FILE #)	
b = 3-bit bit address			
f = 7-bit file register address			

Literal and control operations			
General			
13	8	7	0
OPCODE		k (literal)	
k = 8-bit immediate value			

CALL and GOTO instructions only			
13	11	10	0
OPCODE		k (literal)	
k = 11-bit immediate value			

# PIC16F8X

TABLE 9-2 PIC16FXX INSTRUCTION SET

Mnemonic, Operands	Description	Cycles	14-Bit Opcode				Status Affected	Notes
			MSb	Lsb				
<b>BYTE-ORIENTED FILE REGISTER OPERATIONS</b>								
<b>ADDWF</b> f, d	Add W and f	1	00	0111	dfff	ffff	C,DC,Z	1,2
<b>ANDWF</b> f, d	AND W with f	1	00	0101	dfff	ffff	Z	1,2
<b>CLRF</b> f	Clear f	1	00	0001	1fff	ffff	Z	2
<b>CLRW</b> -	Clear W	1	00	0001	0xxx	xxxx	Z	
<b>COMF</b> f, d	Complement f	1	00	1001	dfff	ffff	Z	1,2
<b>DECFSZ</b> f, d	Decrement f	1	00	0011	dfff	ffff	Z	1,2
<b>DECF</b> f, d	Decrement f, Skip if 0	1(2)	00	1011	dfff	ffff		1,2,3
<b>INCFSZ</b> f, d	Increment f	1	00	1010	dfff	ffff	Z	1,2
<b>INCFSZ</b> f, d	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff		1,2,3
<b>IOWF</b> f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1,2
<b>MOVF</b> f, d	Move f	1	00	1000	dfff	ffff	Z	1,2
<b>MOVWF</b> f	Move W to f	1	00	0000	1fff	ffff		
<b>NOP</b> -	No Operation	1	00	0000	0xx0	0000		
<b>RLF</b> f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	C	1,2
<b>RRF</b> f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	1,2
<b>SUBWF</b> f, d	Subtract W from f	1	00	0010	dfff	ffff	C,DC,Z	1,2
<b>SWAPF</b> f, d	Swap nibbles in f	1	00	1110	dfff	ffff		1,2
<b>XORWF</b> f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1,2
<b>BIT-ORIENTED FILE REGISTER OPERATIONS</b>								
<b>BCF</b> f, b	Bit Clear f	1	01	00bb	bfff	ffff		1,2
<b>BSF</b> f, b	Bit Set f	1	01	01bb	bfff	ffff		1,2
<b>BTFSZ</b> f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		3
<b>BTFSZ</b> f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
<b>LITERAL AND CONTROL OPERATIONS</b>								
<b>ADDLW</b> k	Add literal and W	1	11	111x	kkkk	kkkk	C,DC,Z	
<b>ANDLW</b> k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
<b>CALL</b> k	Call subroutine	2	10	0kkk	kkkk	kkkk		
<b>CLRWD</b> -	Clear Watchdog Timer	1	00	0000	0110	0100	TO,PD	
<b>GOTO</b> k	Go to address	2	10	1kkk	kkkk	kkkk		
<b>IORLW</b> k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
<b>MOVLW</b> k	Move literal to W	1	11	00xx	kkkk	kkkk		
<b>RETFIE</b> -	Return from interrupt	2	00	0000	0000	1001		
<b>RETLW</b> k	Return with literal in W	2	11	01xx	kkkk	kkkk		
<b>RETURN</b> -	Return from Subroutine	2	00	0000	0000	1000	TO,PD	
<b>SLEEP</b> -	Go into standby mode	1	00	0000	0110	0011	C,DC,Z	
<b>SUBLW</b> k	Subtract W from literal	1	11	110x	kkkk	kkkk	Z	
<b>XORLW</b> k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk		

- Note 1:** When an I/O register is modified as a function of itself ( e.g., **MOVF PORTB, 1**), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.
- 2:** If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 Module.
- 3:** If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.



MICROCHIP

# PIC16F872

## 28-Pin, 8-Bit CMOS FLASH Microcontroller

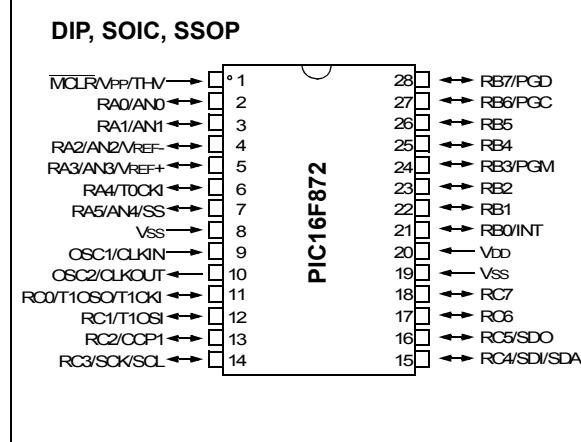
### Devices Included in this Data Sheet:

- PIC16F872

### Microcontroller Core Features:

- High-performance RISC CPU
- Only 35 single word instructions to learn
- All single cycle instructions except for program branches which are two cycle
- Operating speed: DC - 20 MHz clock input  
DC - 200 ns instruction cycle
- 2K x 14 words of FLASH Program Memory  
128 x 8 bytes of Data Memory (RAM)  
64 x 8 bytes of EEPROM Data Memory
- Pinout compatible to the PIC16C72A
- Interrupt capability (up to 10 sources)
- Eight level deep hardware stack
- Direct, indirect and relative addressing modes
- Power-on Reset (POR)
- Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Watchdog Timer (WDT) with its own on-chip RC oscillator for reliable operation
- Programmable code-protection
- Power saving SLEEP mode
- Selectable oscillator options
- Low-power, high-speed CMOS FLASH/EEPROM technology
- Fully static design
- In-Circuit Serial Programming™ (ICSP) via two pins
- Single 5V In-Circuit Serial Programming capability
- In-Circuit Debugging via two pins
- Processor read/write access to program memory
- Wide operating voltage range: 2.0V to 5.5V
- High Sink/Source Current: 25 mA
- Commercial and Industrial temperature ranges
- Low-power consumption:
  - < 2 mA typical @ 5V, 4 MHz
  - 20 µA typical @ 3V, 32 kHz
  - < 1 µA typical standby current

### Pin Diagram



### Peripheral Features:

- Timer0: 8-bit timer/counter with 8-bit prescaler
- Timer1: 16-bit timer/counter with prescaler, can be incremented during sleep via external crystal/clock
- Timer2: 8-bit timer/counter with 8-bit period register, prescaler and postscaler
- One Capture, Compare, PWM module
  - Capture is 16-bit, max. resolution is 12.5 ns
  - Compare is 16-bit, max. resolution is 200 ns
  - PWM max. resolution is 10-bit
- 10-bit multi-channel Analog-to-Digital converter
- Synchronous Serial Port (SSP) with SPI™ (Master Mode) and I<sup>2</sup>C™ (Master/Slave)
- Brown-out detection circuitry for Brown-out Reset (BOR)

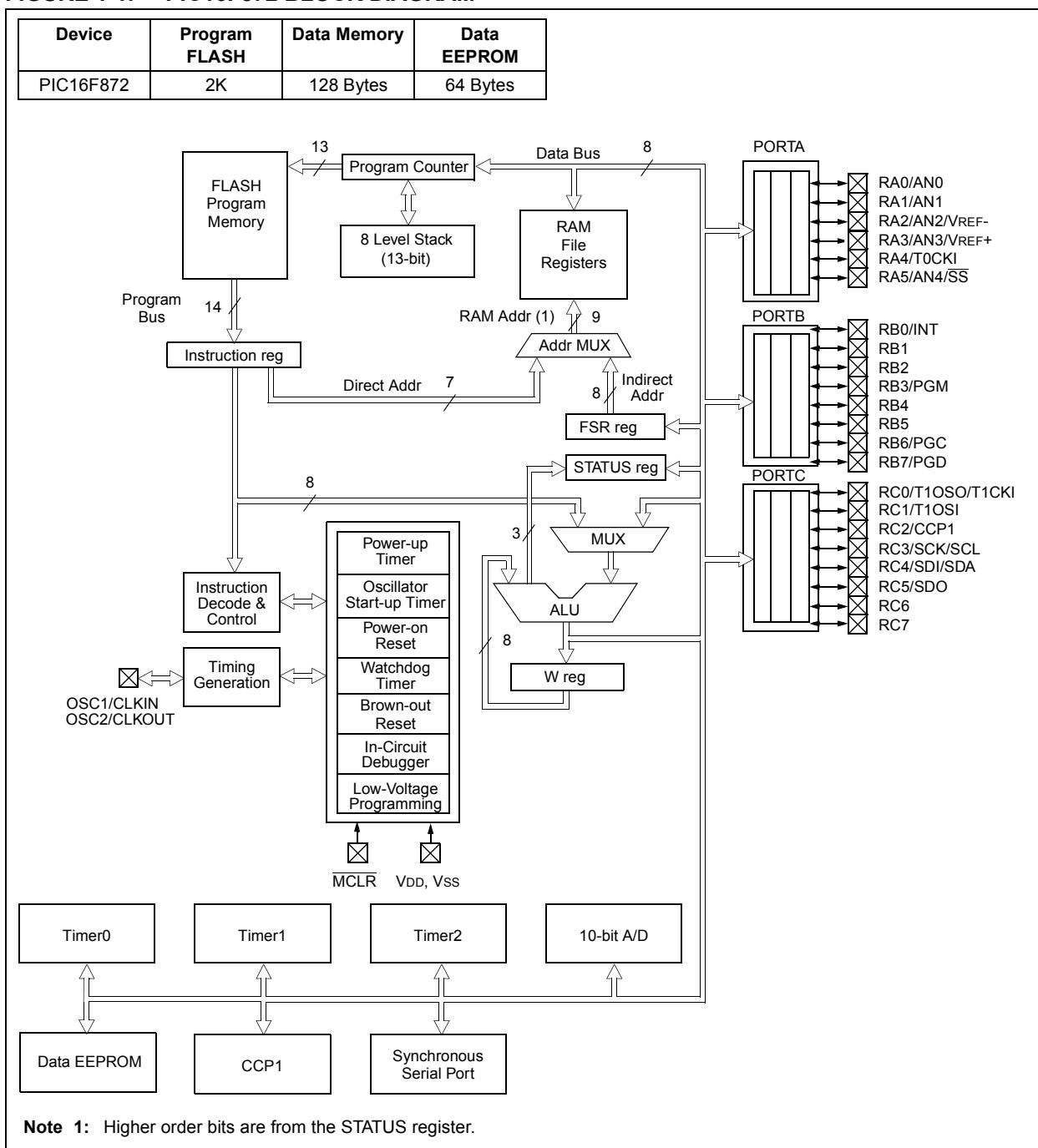
## 1.0 DEVICE OVERVIEW

This document contains device-specific information. Additional information may be found in the PICmicro™ Mid-Range Reference Manual, (DS33023), which may be obtained from your local Microchip Sales Representative or downloaded from the Microchip website. The Reference Manual should be considered a comple-

mentary document to this data sheet, and is highly recommended reading for a better understanding of the device architecture and operation of the peripheral modules.

This data sheet covers the PIC16F872 device. The PIC16F872 is a 28-pin device and its block diagram is shown in Figure 1-1.

**FIGURE 1-1: PIC16F872 BLOCK DIAGRAM**



# PIC16F872

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**TABLE 1-1: PIC16F872 PINOUT DESCRIPTION**

Pin Name	DIP Pin#	SOIC Pin#	I/O/P Type	Buffer Type	Description
OSC1/CLKIN	9	9	I	ST/CMOS <sup>(3)</sup>	Oscillator crystal input/external clock source input.
OSC2/CLKOUT	10	10	O	—	Oscillator crystal output. Connects to crystal or resonator in crystal oscillator mode. In RC mode, the OSC2 pin outputs CLKOUT, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
MCLR/VPP/THV	1	1	I/P	ST	Master clear (reset) input or programming voltage input or high voltage test mode control. This pin is an active low reset to the device.
RA0/AN0	2	2	I/O	TTL	PORTA is a bi-directional I/O port. RA0 can also be analog input0.
RA1/AN1	3	3	I/O	TTL	RA1 can also be analog input1.
RA2/AN2/VREF-	4	4	I/O	TTL	RA2 can also be analog input2 or negative analog reference voltage.
RA3/AN3/VREF+	5	5	I/O	TTL	RA3 can also be analog input3 or positive analog reference voltage.
RA4/T0CKI	6	6	I/O	ST	RA4 can also be the clock input to the Timer0 module. Output is open drain type.
RA5/SS/AN4	7	7	I/O	TTL	RA5 can also be analog input4 or the slave select for the synchronous serial port.
RB0/INT	21	21	I/O	TTL/ST <sup>(1)</sup>	PORTB is a bi-directional I/O port. PORTB can be software programmed for internal weak pull-up on all inputs. RB0 can also be the external interrupt pin.
RB1	22	22	I/O	TTL	
RB2	23	23	I/O	TTL	
RB3/PGM	24	24	I/O	TTL/ST <sup>(1)</sup>	RB3 can also be the low voltage programming input.
RB4	25	25	I/O	TTL	Interrupt on change pin.
RB5	26	26	I/O	TTL	Interrupt on change pin.
RB6/PGC	27	27	I/O	TTL/ST <sup>(2)</sup>	Interrupt on change pin or In-Circuit Debugger pin. Serial programming clock.
RB7/PGD	28	28	I/O	TTL/ST <sup>(2)</sup>	Interrupt on change pin or In-Circuit Debugger pin. Serial programming data.
RC0/T1OSO/T1CKI	11	11	I/O	ST	PORTC is a bi-directional I/O port. RC0 can also be the Timer1 oscillator output or Timer1 clock input.
RC1/T1OSI	12	12	I/O	ST	RC1 can also be the Timer1 oscillator input.
RC2/CCP1	13	13	I/O	ST	RC2 can also be the Capture1 input/Compare1 output/PWM1 output.
RC3/SCK/SCL	14	14	I/O	ST	RC3 can also be the synchronous serial clock input/output for both SPI and I <sup>2</sup> C modes.
RC4/SDI/SDA	15	15	I/O	ST	RC4 can also be the SPI Data In (SPI mode) or data I/O (I <sup>2</sup> C mode).
RC5/SDO	16	16	I/O	ST	RC5 can also be the SPI Data Out (SPI mode).
RC6	17	17	I/O	ST	
RC7	18	18	I/O	ST	
Vss	8, 19	8, 19	P	—	Ground reference for logic and I/O pins.
Vdd	20	20	P	—	Positive supply for logic and I/O pins.

Legend: I = input      O = output      I/O = input/output      P = power  
— = Not used      TTL = TTL input      ST = Schmitt Trigger input

**Note 1:** This buffer is a Schmitt Trigger input when configured as the external interrupt or LVP.

**2:** This buffer is a Schmitt Trigger input when used in serial programming mode.

**3:** This buffer is a Schmitt Trigger input when configured in RC oscillator mode and a CMOS input otherwise.

## 2.0 MEMORY ORGANIZATION

There are three memory blocks in each of these PICmicro® MCUs. The Program Memory and Data Memory have separate buses, so that concurrent access can occur, and is detailed in this section. The EEPROM data memory block is detailed in Section 4.0.

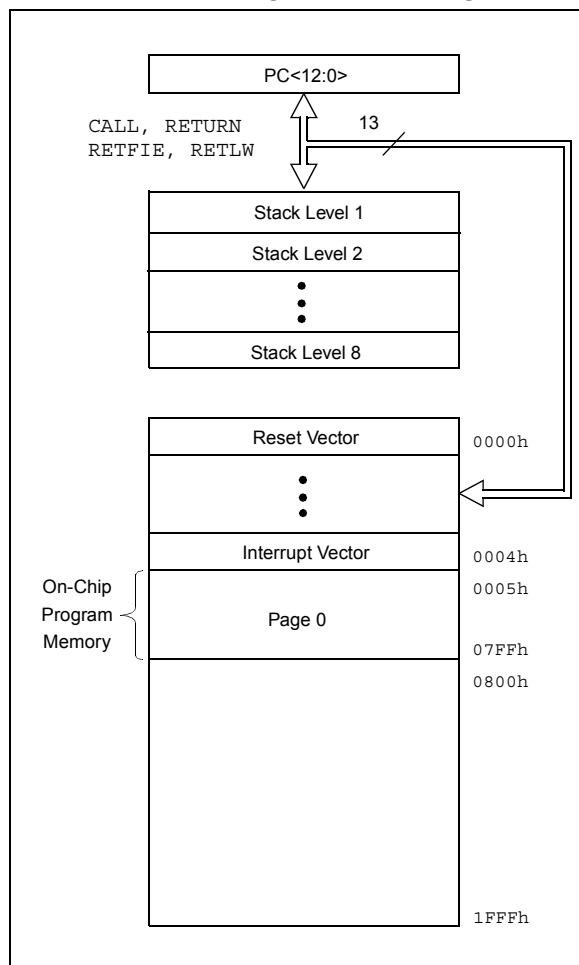
Additional information on device memory may be found in the PICmicro™ Mid-Range Reference Manual, (DS33023).

### 2.1 Program Memory Organization

The PIC16F872 devices have a 13-bit program counter capable of addressing an 8K x 14 program memory space. The PIC16F872 device has 2K x 14 words of FLASH program memory. Accessing a location above the physically implemented address will cause a wrap-around.

The reset vector is at 0000h and the interrupt vector is at 0004h.

**FIGURE 2-1: PIC16F872 PROGRAM MEMORY MAP AND STACK**



## 2.2 Data Memory Organization

The data memory is partitioned into multiple banks which contain the General Purpose Registers and the Special Function Registers. Bits RP1(STATUS<6>) and RP0 (STATUS<5>) are the bank select bits.

RP<1:0>	Bank
00	0
01	1
10	2
11	3

Each bank extends up to 7Fh (128 bytes). The lower locations of each bank are reserved for the Special Function Registers. Above the Special Function Registers are General Purpose Registers, implemented as static RAM. All implemented banks contain Special Function Registers. Some "high use" Special Function Registers from one bank may be mirrored in another bank for code reduction and quicker access.

**Note:** EEPROM Data Memory description can be found in Section 4.0 of this Data Sheet

### 2.2.1 GENERAL PURPOSE REGISTER FILE

The register file can be accessed either directly, or indirectly through the File Select Register FSR.

# PIC16F872

FIGURE 2-2: PIC16F872 REGISTER FILE MAP

						File Address					
Indirect addr.(*)	00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)					
TMR0	01h	OPTION_REG	81h	TMR0	101h	OPTION_REG					
PCL	02h	PCL	82h	PCL	102h	PCL					
STATUS	03h	STATUS	83h	STATUS	103h	STATUS					
FSR	04h	FSR	84h	FSR	104h	FSR					
PORTA	05h	TRISA	85h	PORTB	105h						
PORTB	06h	TRISB	86h		106h	TRISB					
PORTC	07h	TRISC	87h		107h						
	08h		88h		108h						
	09h		89h		109h						
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH					
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON					
PIR1	0Ch	PIE1	8Ch	EEDATA	10Ch	EECON1					
PIR2	0Dh	PIE2	8Dh	EEADR	10Dh	EECON2					
TMR1L	0Eh	PCON	8Eh	EEDATH	10Eh	Reserved <sup>(1)</sup>					
TMR1H	0Fh		8Fh	EEADRH	10Fh	Reserved <sup>(1)</sup>					
T1CON	10h		90h		110h						
TMR2	11h	SSPCON2	91h								
T2CON	12h	PR2	92h								
SSPBUF	13h	SSPADD	93h								
SSPCON	14h	SSPSTAT	94h								
CCPR1L	15h		95h								
CCPR1H	16h		96h								
CCP1CON	17h		97h								
	18h		98h								
	19h		99h								
	1Ah		9Ah								
	1Bh		9Bh								
	1Ch		9Ch								
	1Dh		9Dh								
ADRESH	1Eh	ADRESL	9Eh								
ADC0N0	1Fh	ADC0N1	9Fh								
General Purpose Register 96 Bytes	20h	General Purpose Register 32 Bytes	A0h	accesses 20h-7Fh	120h	accesses A0h - BFh					
			BFh								
			C0h								
			EFh								
			F0h								
			FFh								
		accesses 70h-7Fh									
Bank 0		Bank 1		Bank 2		Bank 3					
 Unimplemented data memory locations, read as '0'.											
* Not a physical register.											
<b>Note 1:</b> These registers are reserved; maintain these registers clear.											

## 2.2.2 SPECIAL FUNCTION REGISTERS

The Special Function Registers are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. A list of these registers is given in Table 2-1.

The Special Function Registers can be classified into two sets: core (CPU) and peripheral. Those registers associated with the core functions are described in detail in this section. Those related to the operation of the peripheral features are described in detail in the peripheral feature section.

**TABLE 2-1: SPECIAL FUNCTION REGISTER SUMMARY**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets (2)
<b>Bank 0</b>											
00h <sup>(3)</sup>	INDF									0000 0000	0000 0000
01h	TMR0									xxxx xxxx	uuuu uuuu
02h <sup>(3)</sup>	PCL									0000 0000	0000 0000
03h <sup>(3)</sup>	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	C	0001 1xxx	000q quuu
04h <sup>(3)</sup>	FSR	Indirect data memory address pointer								xxxx xxxx	uuuu uuuu
05h	PORTA	—	—							--0x 0000	--0u 0000
06h	PORTB	PORTB Data Latch when written: PORTB pins when read								xxxx xxxx	uuuu uuuu
07h	PORTC	PORTC Data Latch when written: PORTC pins when read								xxxx xxxx	uuuu uuuu
08h	—	Unimplemented								—	—
09h	—	Unimplemented								—	—
0Ah <sup>(1,3)</sup>	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					--0 0000	--0 0000
0Bh <sup>(3)</sup>	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBI	0000 000x	0000 000u
0Ch	PIR1	(4)	ADIF	(4)	(4)	SSPIF	CCP1IF	TMR2IF	TMR1IF	r0rr 0000	r0rr 0000
0Dh	PIR2	—	(4)	—	EEIF	BCLIF	—	—	(4)	-r0 0--r	-r0 0--r
0Eh	TMR1L	Holding register for the Least Significant Byte of the 16-bit TMR1 register								xxxx xxxx	uuuu uuuu
0Fh	TMR1H	Holding register for the Most Significant Byte of the 16-bit TMR1 register								xxxx xxxx	uuuu uuuu
10h	T1CON	—	—	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR1ON	--00 0000	-uuu uuuu
11h	TMR2	Timer2 module's register								0000 0000	0000 0000
12h	T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
13h	SSPBUF	Synchronous Serial Port Receive Buffer/Transmit Register								xxxx xxxx	uuuu uuuu
14h	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	0000 0000
15h	CCPR1L	Capture/Compare/PWM Register1 (LSB)								xxxx xxxx	uuuu uuuu
16h	CCPR1H	Capture/Compare/PWM Register1 (MSB)								xxxx xxxx	uuuu uuuu
17h	CCP1CON	—	—	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
18h	—	Unimplemented								—	—
19h	—	Unimplemented								—	—
1Ah	—	Unimplemented								—	—
1Bh	—	Unimplemented								—	—
1Ch	—	Unimplemented								—	—
1Dh	—	Unimplemented								—	—
1Eh	ADRESH	A/D Result Register High Byte								xxxx xxxx	uuuu uuuu
1Fh	ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/ DONE	—	ADON	0000 00-0	0000 00-0

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented read as '0', r = reserved.

Shaded locations are unimplemented, read as '0'.

**Note 1:** The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8> whose contents are transferred to the upper byte of the program counter.

**2:** Other (non power-up) resets include external reset through MCLR and Watchdog Timer Reset.

**3:** These registers can be addressed from any bank.

**4:** These bits are reserved; always maintain these bits clear.

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**TABLE 2-1: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets (2)
<b>Bank 1</b>											
80h <sup>(3)</sup>	INDF									0000 0000	0000 0000
81h	OPTION_REG	RBU	INTEG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
82h <sup>(3)</sup>	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	0000 0000
83h <sup>(3)</sup>	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	C	0001 1xxx	000q quuu
84h <sup>(3)</sup>	FSR	Indirect data memory address pointer								xxxx xxxx	uuuu uuuu
85h	TRISA	—	—	PORTA Data Direction Register						--11 1111	--11 1111
86h	TRISB	PORTB Data Direction Register								1111 1111	1111 1111
87h	TRISC	PORTC Data Direction Register								1111 1111	1111 1111
88h	—	Unimplemented								—	—
89h	—	Unimplemented								—	—
8Ah <sup>(1,3)</sup>	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					---0 0000	---0 0000
8Bh <sup>(3)</sup>	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
8Ch	PIE1	(4)	ADIE	(4)	(4)	SSPIE	CCP1IE	TMR2IE	TMR1IE	r0rr 0000	r0rr 0000
8Dh	PIE2	—	(4)	—	EEIE	BCLIE	—	—	(4)	-r-0 0--r	-r-0 0--r
8Eh	PCON	—	—	—	—	—	—	POR	BOR	---- --qq	---- --uu
8Fh	—	Unimplemented								—	—
90h	—	Unimplemented								—	—
91h	SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000	0000 0000
92h	PR2	Timer2 Period Register								1111 1111	1111 1111
93h	SSPADD	Synchronous Serial Port (I <sup>2</sup> C mode) Address Register								0000 0000	0000 0000
94h	SSPSTAT	SMP	CKE	D/A	P	S	R/W	UA	BF	0000 0000	0000 0000
95h	—	Unimplemented								—	—
96h	—	Unimplemented								—	—
97h	—	Unimplemented								—	—
98h	—	Unimplemented								—	—
99h	—	Unimplemented								—	—
9Ah	—	Unimplemented								—	—
9Bh	—	Unimplemented								—	—
9Ch	—	Unimplemented								—	—
9Dh	—	Unimplemented								—	—
9Eh	ADRESL	A/D Result Register Low Byte								xxxx xxxx	uuuu uuuu
9Fh	ADCON1	ADFM	—	—	—	PCFG3	PCFG2	PCFG1	PCFG0	0--- 0000	0--- 0000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented read as '0', r = reserved.

Shaded locations are unimplemented, read as '0'.

**Note 1:** The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8> whose contents are transferred to the upper byte of the program counter.

**2:** Other (non power-up) resets include external reset through MCLR and Watchdog Timer Reset.

**3:** These registers can be addressed from any bank.

**4:** These bits are reserved; always maintain these bits clear.

**TABLE 2-1: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets (2)
<b>Bank 2</b>											
100h <sup>(3)</sup>	INDF									0000 0000	0000 0000
101h	TMR0									xxxx xxxx	uuuu uuuu
102h <sup>(3)</sup>	PCL									0000 0000	0000 0000
103h <sup>(3)</sup>	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	C	0001 1xxx	000q quuu
104h <sup>(3)</sup>	FSR									xxxx xxxx	uuuu uuuu
105h	—									—	—
106h	PORTB									xxxx xxxx	uuuu uuuu
107h	—									—	—
108h	—									—	—
109h	—									—	—
10Ah <sup>(1,3)</sup>	PCLATH	—	—	—						---0 0000	---0 0000
10Bh <sup>(3)</sup>	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 000x	0000 000u
10Ch	EEDATA									xxxx xxxx	uuuu uuuu
10Dh	EEADR									xxxx xxxx	uuuu uuuu
10Eh	EEDATH	—	—							xxxx xxxx	uuuu uuuu
10Fh	EEADRH	—	—	—						xxxx xxxx	uuuu uuuu
<b>Bank 3</b>											
180h <sup>(3)</sup>	INDF									0000 0000	0000 0000
181h	OPTION_REG	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
182h <sup>(3)</sup>	PCL									0000 0000	0000 0000
183h <sup>(3)</sup>	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	C	0001 1xxx	000q quuu
184h <sup>(3)</sup>	FSR									xxxx xxxx	uuuu uuuu
185h	—									—	—
186h	TRISB									1111 1111	1111 1111
187h	—									—	—
188h	—									—	—
189h	—									—	—
18Ah <sup>(1,3)</sup>	PCLATH	—	—	—						---0 0000	---0 0000
18Bh <sup>(3)</sup>	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 000x	0000 000u
18Ch	EECON1	EEPGD	—	—	—	WRERR	WREN	WR	RD	x--- x000	x--- u000
18Dh	EECON2									-----	-----
18Eh	—									0000 0000	0000 0000
18Fh	—									0000 0000	0000 0000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented read as '0', r = reserved.

Shaded locations are unimplemented, read as '0'.

**Note 1:** The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8> whose contents are transferred to the upper byte of the program counter.

**2:** Other (non power-up) resets include external reset through MCLR and Watchdog Timer Reset.

**3:** These registers can be addressed from any bank.

**4:** These bits are reserved; always maintain these bits clear.

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## 2.2.2.1 STATUS REGISTER

The STATUS register contains the arithmetic status of the ALU, the RESET status and the bank select bits for data memory.

The STATUS register can be the destination for any instruction, as with any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the **TO** and **PD** bits are not writable, therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, **CLRF STATUS** will clear the upper-three bits and set the Z bit. This leaves the STATUS register as **000u uuu** (where **u** = unchanged).

It is recommended, therefore, that only **BCF**, **BSF**, **SWAPF** and **MOVWF** instructions are used to alter the STATUS register, because these instructions do not affect the Z, C or DC bits from the STATUS register. For other instructions not affecting any status bits, see the "Instruction Set Summary."

**Note:** The C and DC bits operate as a borrow and digit borrow bit, respectively, in subtraction. See the **SUBLW** and **SUBWF** instructions for examples.

## REGISTER 2-1: STATUS REGISTER (ADDRESS 03h, 83h, 103h, 183h)

R/W-0	R/W-0	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
IRP	RP1	RP0	<b>TO</b>	<b>PD</b>	Z	DC	C
bit7							bit0
							<p>R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' - n= Value at POR reset</p>
<p>bit 7: <b>IRP:</b> Register Bank Select bit (used for indirect addressing) 1 = Bank 2, 3 (100h - 1FFh) 0 = Bank 0, 1 (00h - FFh)</p>							
<p>bit 6-5: <b>RP&lt;1:0&gt;:</b> Register Bank Select bits (used for direct addressing) 11 = Bank 3 (180h - 1FFh) 10 = Bank 2 (100h - 17Fh) 01 = Bank 1 (80h - FFh) 00 = Bank 0 (00h - 7Fh) Each bank is 128 bytes</p>							
<p>bit 4: <b>TO:</b> Time-out bit 1 = After power-up, <b>CLRWDT</b> instruction, or <b>SLEEP</b> instruction 0 = A WDT time-out occurred</p>							
<p>bit 3: <b>PD:</b> Power-down bit 1 = After power-up or by the <b>CLRWDT</b> instruction 0 = By execution of the <b>SLEEP</b> instruction</p>							
<p>bit 2: <b>Z:</b> Zero bit 1 = The result of an arithmetic or logic operation is zero 0 = The result of an arithmetic or logic operation is not zero</p>							
<p>bit 1: <b>DC:</b> Digit carry/borrow bit (<b>ADDWF</b>, <b>ADDLW</b>, <b>SUBLW</b>, <b>SUBWF</b> instructions) (for borrow the polarity is reversed) 1 = A carry-out from the 4th low order bit of the result occurred 0 = No carry-out from the 4th low order bit of the result</p>							
<p>bit 0: <b>C:</b> Carry/borrow bit (<b>ADDWF</b>, <b>ADDLW</b>, <b>SUBLW</b>, <b>SUBWF</b> instructions) 1 = A carry-out from the most significant bit of the result occurred 0 = No carry-out from the most significant bit of the result occurred <b>Note:</b> For borrow the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (<b>RRF</b>, <b>RLF</b>) instructions, this bit is loaded with either the high or low order bit of the source register.</p>							

### 3.0 I/O PORTS

Some pins for these I/O ports are multiplexed with an alternate function for the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

Additional information on I/O ports may be found in the PICmicro™ Mid-Range Reference Manual, (DS33023).

#### 3.1 PORTA and the TRISA Register

PORTA is a 6-bit wide, bi-directional port. The corresponding data direction register is TRISA. Setting a TRISA bit (=1) will make the corresponding PORTA pin an input (i.e., put the corresponding output driver in a hi-impedance mode). Clearing a TRISA bit (=0) will make the corresponding PORTA pin an output (i.e., put the contents of the output latch on the selected pin).

Reading the PORTA register reads the status of the pins, whereas writing to it will write to the port latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, the value is modified and then written to the port data latch.

Pin RA4 is multiplexed with the Timer0 module clock input to become the RA4/T0CKI pin. The RA4/T0CKI pin is a Schmitt Trigger input and an open drain output. All other PORTA pins have TTL input levels and full CMOS output drivers.

Other PORTA pins are multiplexed with analog inputs and analog VREF input. The operation of each pin is selected by clearing/setting the control bits in the ADCON1 register (A/D Control Register1).

**Note:** On a Power-on Reset, these pins are configured as analog inputs and read as '0'.

The TRISA register controls the direction of the RA pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

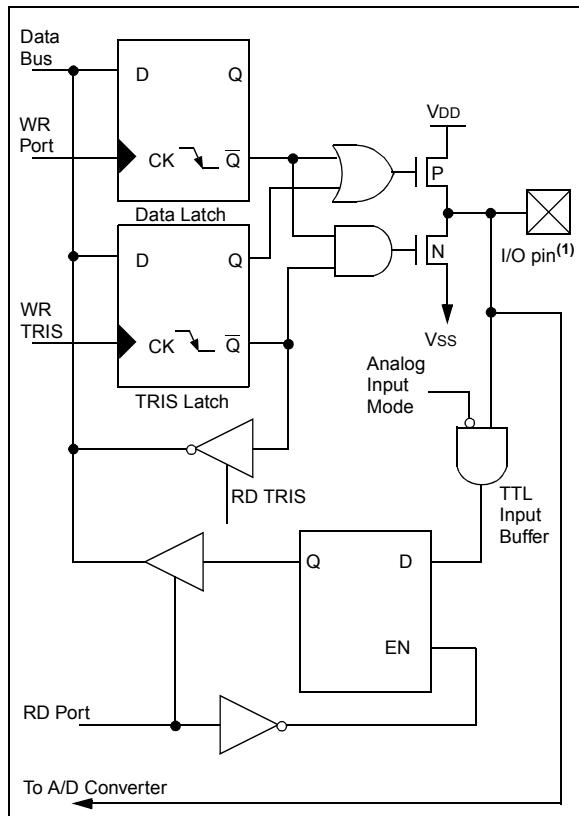
#### EXAMPLE 3-1: INITIALIZING PORTA

```

BCF      STATUS, RP0      ; 
BCF      STATUS, RP1      ; Bank0
CLRF    PORTA             ; Initialize PORTA by
                           ; clearing output
                           ; data latches
BSF      STATUS, RP0      ; Select Bank 1
MOVLW   0x06              ; Configure all pins
MOVWF   ADCON1            ; as digital inputs
MOVLW   0xCF              ; Value used to
                           ; initialize data
                           ; direction
MOVWF   TRISA             ; Set RA<3:0> as inputs
                           ; RA<5:4> as outputs
                           ; TRISA<7:6> are always
                           ; read as '0'.

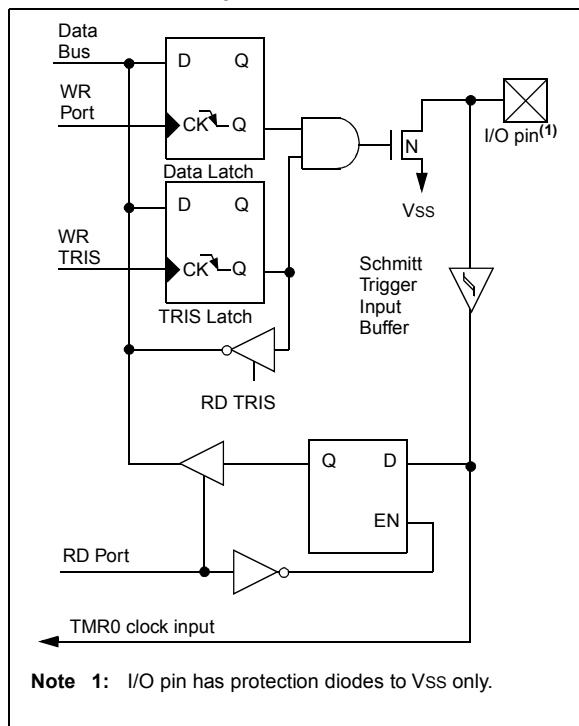
```

**FIGURE 3-1: BLOCK DIAGRAM OF RA<3:0> AND RA5 PINS**



**Note 1:** I/O pins have protection diodes to VDD and VSS.

**FIGURE 3-2: BLOCK DIAGRAM OF RA4/T0CKI PIN**



**Note 1:** I/O pin has protection diodes to VSS only.

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**TABLE 3-1: PORTA FUNCTIONS**

Name	Bit#	Buffer	Function
RA0/AN0	bit0	TTL	Input/output or analog input
RA1/AN1	bit1	TTL	Input/output or analog input
RA2/AN2	bit2	TTL	Input/output or analog input
RA3/AN3/VREF	bit3	TTL	Input/output or analog input or VREF
RA4/T0CKI	bit4	ST	Input/output or external clock input for Timer0 Output is open drain type
RA5/SS/AN4	bit5	TTL	Input/output or slave select input for synchronous serial port or analog input

Legend: TTL = TTL input, ST = Schmitt Trigger input.

**TABLE 3-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
05h	PORTA	—	—	RA5	RA4	RA3	RA2	RA1	RA0	--0x 0000	--0u 0000
85h	TRISA	—	—	PORTA Data Direction Register						--11 1111	--11 1111
9Fh	ADCON1	ADFM	—	—	—	PCFG3	PCFG2	PCFG1	PCFG0	--0- 0000	--0- 0000

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTA.

**Note:** When using the SSP module in SPI slave mode and SS enabled, the A/D converter must be set to one of the following modes where PCFG<3:0> = 0100, 0101, 011x, 1101, 1110, 1111.

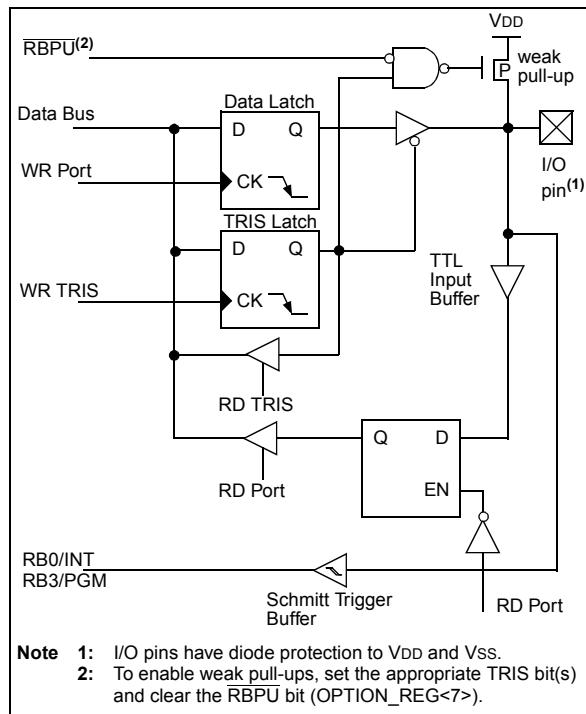
### 3.2 PORTB and the TRISB Register

PORTB is an 8-bit wide bi-directional port. The corresponding data direction register is TRISB. Setting a TRISB bit (=1) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a hi-impedance mode). Clearing a TRISB bit (=0) will make the corresponding PORTB pin an output (i.e., put the contents of the output latch on the selected pin).

Three pins of PORTB are multiplexed with the Low Voltage Programming function; RB3/PGM, RB6/PGC and RB7/PGD. The alternate functions of these pins are described in the Special Features Section.

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by clearing bit RBPU (OPTION\_REG<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

**FIGURE 3-3: BLOCK DIAGRAM OF RB<3:0> PINS**



Four of PORTB's pins, RB<7:4>, have an interrupt on change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB<7:4> pin configured as an output is excluded from the interrupt on change comparison). The input pins (of RB<7:4>) are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of RB<7:4> are OR'ed together to generate the RB Port Change Interrupt with flag bit RBIF (INTCON<0>).

This interrupt can wake the device from SLEEP. The user, in the interrupt service routine, can clear the interrupt in the following manner:

- Any read or write of PORTB. This will end the mismatch condition.
- Clear flag bit RBIF.

A mismatch condition will continue to set flag bit RBIF. Reading PORTB will end the mismatch condition and allow flag bit RBIF to be cleared.

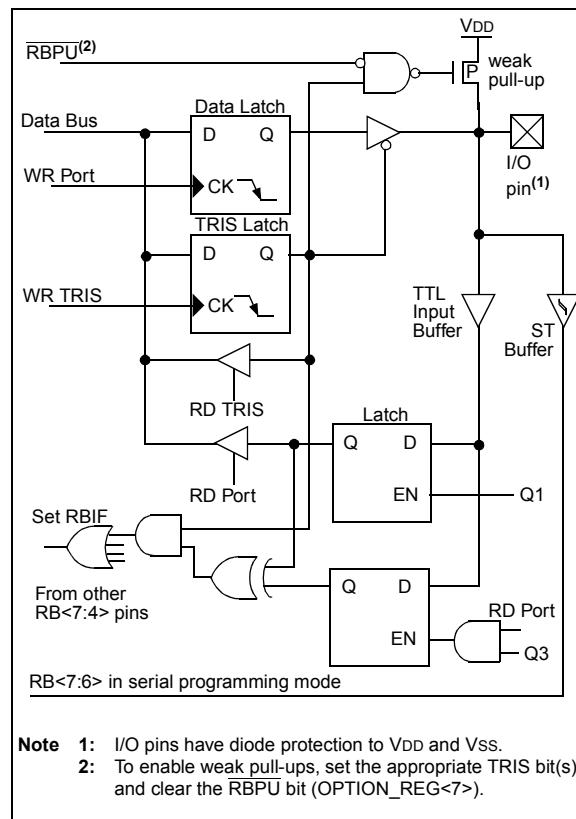
The interrupt on change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt on change feature. Polling of PORTB is not recommended while using the interrupt on change feature.

This interrupt on mismatch feature, together with software configurable pull-ups on these four pins, allow easy interface to a keypad and make it possible for wake-up on key-depression. Refer to the Embedded Control Handbook, "Implementing Wake-Up on Key Stroke" (AN552).

RB0/INT is an external interrupt input pin and is configured using the INTEDG bit (OPTION\_REG<6>).

RB0/INT is discussed in detail in Section 11.10.1.

**FIGURE 3-4: BLOCK DIAGRAM OF RB<7:4> PINS**



**Note:** When using Low Voltage ICSP Programming (LVP) and the pull-ups on PORTB are enabled, bit 3 in the TRISB register must be cleared to disable the pull-up on RB3 and ensure the proper operation of the device.

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**TABLE 3-3: PORTB FUNCTIONS**

Name	Bit#	Buffer	Function
RB0/INT	bit0	TTL/ST <sup>(1)</sup>	Input/output pin or external interrupt input. Internal software programmable weak pull-up.
RB1	bit1	TTL	Input/output pin. Internal software programmable weak pull-up.
RB2	bit2	TTL	Input/output pin. Internal software programmable weak pull-up.
RB3/PGM	bit3	TTL/ST <sup>(1)</sup>	Input/output pin or programming pin in LVP mode. Internal software programmable weak pull-up.
RB4	bit4	TTL	Input/output pin (with interrupt on change). Internal software programmable weak pull-up.
RB5	bit5	TTL	Input/output pin (with interrupt on change). Internal software programmable weak pull-up.
RB6/PGC	bit6	TTL/ST <sup>(2)</sup>	Input/output pin (with interrupt on change) or In-Circuit Debugger pin. Internal software programmable weak pull-up. Serial programming clock.
RB7/PGD	bit7	TTL/ST <sup>(2)</sup>	Input/output pin (with interrupt on change) or In-Circuit Debugger pin. Internal software programmable weak pull-up. Serial programming data.

Legend: TTL = TTL input, ST = Schmitt Trigger input.

**Note 1:** This buffer is a Schmitt Trigger input when configured as the external interrupt or LVP mode.

**2:** This buffer is a Schmitt Trigger input when used in serial programming mode.

**TABLE 3-4: SUMMARY OF REGISTERS ASSOCIATED WITH PORTB**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
06h, 106h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	uuuu uuuu
86h, 186h	TRISB	PORTB Data Direction Register								1111 1111	1111 1111
81h, 181h	OPTION_REG	RBPU	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged. Shaded cells are not used by PORTB.

### 3.3 PORTC and the TRISC Register

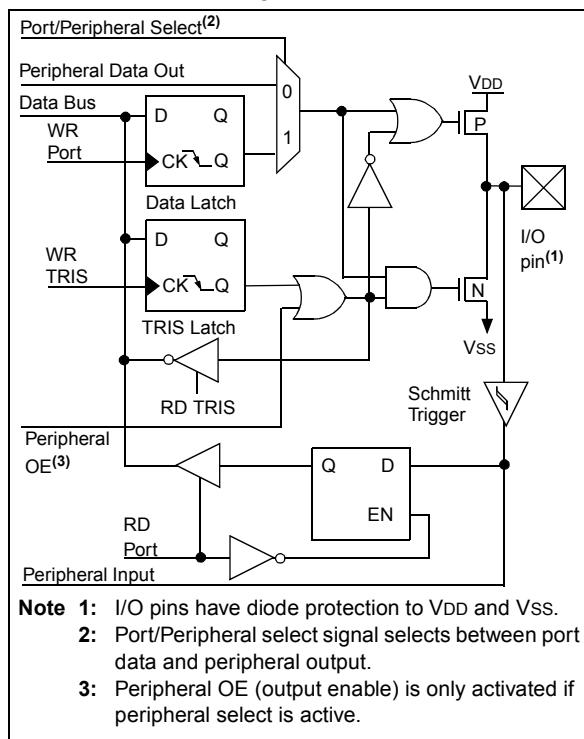
PORTC is an 8-bit wide, bi-directional port. The corresponding data direction register is TRISC. Setting a TRISC bit (=1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a hi-impedance mode). Clearing a TRISC bit (=0) will make the corresponding PORTC pin an output (i.e., put the contents of the output latch on the selected pin).

PORTC is multiplexed with several peripheral functions (Table 3-5). PORTC pins have Schmitt Trigger input buffers.

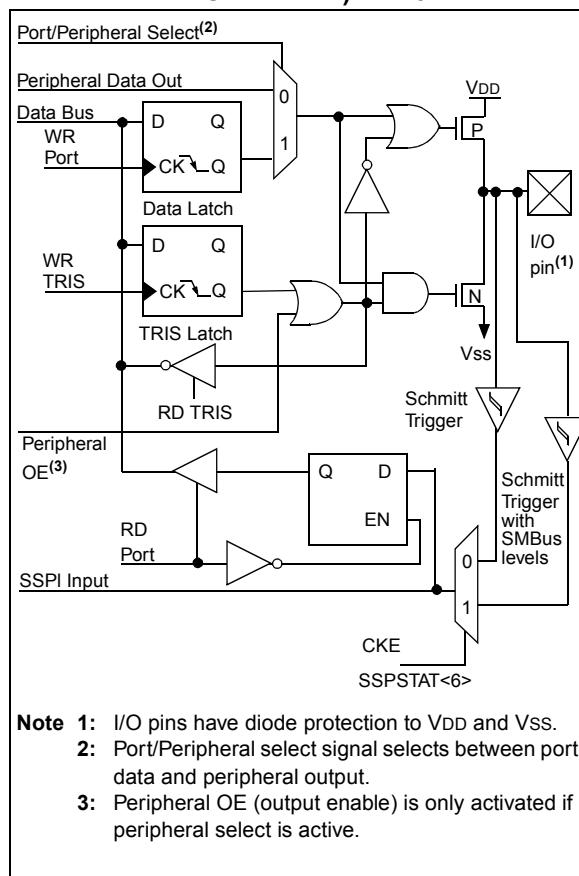
When the I<sup>2</sup>C module is enabled, the PORTC (3:4) pins can be configured with normal I<sup>2</sup>C levels or with SMBUS levels by using the CKE bit (SSPSTAT<6>).

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. Since the TRIS bit override is in effect while the peripheral is enabled, read-modify-write instructions (BSF, BCF, XORWF) with TRISC as destination should be avoided. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

**FIGURE 3-5: PORTC BLOCK DIAGRAM  
(PERIPHERAL OUTPUT OVERRIDE) RC<0:2>  
RC<5:7>**



**FIGURE 3-6: PORTC BLOCK DIAGRAM  
(PERIPHERAL OUTPUT OVERRIDE) RC<3:4>**



# PIC16F872

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**TABLE 3-5: PORTC FUNCTIONS**

Name	Bit#	Buffer Type	Function
RC0/T1OSO/T1CKI	bit0	ST	Input/output port pin or Timer1 oscillator output/Timer1 clock input.
RC1/T1OSI	bit1	ST	Input/output port pin or Timer1 oscillator input.
RC2/CCP1	bit2	ST	Input/output port pin or Capture1 input/Compare1 output/PWM1 output.
RC3/SCK/SCL	bit3	ST	RC3 can also be the synchronous serial clock for both SPI and I <sup>2</sup> C modes.
RC4/SDI/SDA	bit4	ST	RC4 can also be the SPI Data In (SPI mode) or data I/O (I <sup>2</sup> C mode).
RC5/SDO	bit5	ST	Input/output port pin or Synchronous Serial Port data output.
RC6	bit6	ST	Input/output port pin.
RC7	bit7	ST	Input/output port pin.

Legend: ST = Schmitt Trigger input.

**TABLE 3-6: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC**

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other resets
07h	PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	xxxx xxxx	uuuu uuuu
87h	TRISC	PORTC Data Direction Register								1111 1111	1111 1111

Legend: x = unknown, u = unchanged.

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## REGISTER 11-1: CONFIGURATION WORD

CP1	CP0	DEBUG	—	WRT	CPD	LVP	BODEN	CP1	CP0	PWRTE	WDTE	FOSC1	FOSC0	bit13	bit0	Register: CONFIG Address 2007h
-----	-----	-------	---	-----	-----	-----	-------	-----	-----	-------	------	-------	-------	-------	------	-----------------------------------

bit 13-12:

bit 5-4: **CP<1:0>**: Flash Program Memory Code Protection bits <sup>(2)</sup>

11 = Code protection off

10 = 0000h to 06FFh code protected

01 = 0000h to 03FFh code protected

00 = 0000h to 07FFh code protected

bit 11: **DEBUG**: In-Circuit Debugger Mode

1 = In-Circuit Debugger disabled, RB6 and RB7 are general purpose I/O pins.

0 = In-Circuit Debugger enabled, RB6 and RB7 are dedicated to the debugger.

bit 10: **Unimplemented: Read as '1'**

bit 9: **WRT**: Flash Program Memory Write Enable

1 = Unprotected program memory may be written to by EECON control

0 = Unprotected program memory may not be written to by EECON control

bit 8: **CPD**: Data EE Memory Code Protection

1 = Code protection off

0 = Data EEPROM memory code protected

bit 7: **LVP**: Low Voltage In-Circuit Serial Programming Enable bit

1 = RB3/PGM pin has PGM function, low voltage programming enabled

0 = RB3 is digital I/O, HV on MCLR must be used for programming

bit 6: **BODEN**: Brown-out Reset Enable bit <sup>(1)</sup>

1 = BOR enabled

0 = BOR disabled

bit 3: **PWRTE**: Power-up Timer Enable bit <sup>(1)</sup>

1 = PWRT disabled

0 = PWRT enabled

bit 2: **WDTE**: Watchdog Timer Enable bit

1 = WDT enabled

0 = WDT disabled

bit 1-0: **FOSC1:FOSC0**: Oscillator Selection bits

11 = RC oscillator

10 = HS oscillator

01 = XT oscillator

00 = LP oscillator

**Note 1:** Enabling Brown-out Reset automatically enables Power-up Timer (PWRT), regardless of the value of bit **PWRTE**. Ensure the Power-up Timer is enabled anytime Brown-out Reset is enabled.

**2:** All of the CP<1:0> pairs have to be given the same value to enable the code protection scheme listed.

# PIC16F872

**TABLE 11-2: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR**

Osc Type	Crystal Freq	Cap. Range C1	Cap. Range C2
LP	32 kHz	33 pF	33 pF
	200 kHz	15 pF	15 pF
XT	200 kHz	47-68 pF	47-68 pF
	1 MHz	15 pF	15 pF
	4 MHz	15 pF	15 pF
HS	4 MHz	15 pF	15 pF
	8 MHz	15-33 pF	15-33 pF
	20 MHz	15-33 pF	15-33 pF
<b>These values are for design guidance only.</b> See notes at bottom of page.			
<b>Crystals Used</b>			
32 kHz	Epson C-001R32.768K-A	$\pm 20$ PPM	
200 kHz	STD XTL 200.000KHz	$\pm 20$ PPM	
1 MHz	ECS ECS-10-13-1	$\pm 50$ PPM	
4 MHz	ECS ECS-40-20-1	$\pm 50$ PPM	
8 MHz	EPSON CA-301 8.000M-C	$\pm 30$ PPM	
20 MHz	EPSON CA-301 20.000M-C	$\pm 30$ PPM	

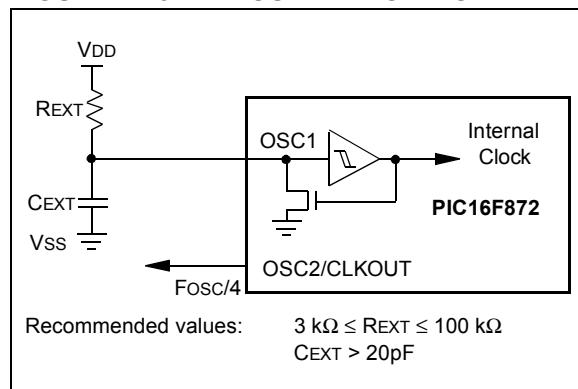
**Note 1:** Higher capacitance increases the stability of oscillator but also increases the start-up time.

- 2:** Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
- 3:** Rs may be required in HS mode, as well as XT mode, to avoid overdriving crystals with low drive level specification.
- 4:** When migrating from other PICmicro devices, oscillator performance should be verified.

### 11.2.3 RC OSCILLATOR

For timing insensitive applications, the “RC” device option offers additional cost savings. The RC oscillator frequency is a function of the supply voltage, the resistor ( $R_{EXT}$ ) and capacitor ( $C_{EXT}$ ) values, and the operating temperature. In addition to this, the oscillator frequency will vary from unit to unit due to normal process parameter variation. Furthermore, the difference in lead frame capacitance between package types will also affect the oscillation frequency, especially for low  $C_{EXT}$  values. The user also needs to take into account variation due to tolerance of external R and C components used. Figure 11-3 shows how the R/C combination is connected to the PIC16F872.

**FIGURE 11-3: RC OSCILLATOR MODE**



## 11.10 Interrupts

The PIC16F872 has 10 sources of interrupt. The interrupt control register (INTCON) records individual interrupt requests in flag bits. It also has individual and global interrupt enable bits.

**Note:** Individual interrupt flag bits are set, regardless of the status of their corresponding mask bit or the GIE bit.

A global interrupt enable bit, GIE (INTCON<7>) enables (if set) all un-masked interrupts or disables (if cleared) all interrupts. When bit GIE is enabled, and an interrupt's flag bit and mask bit are set, the interrupt will vector immediately. Individual interrupts can be disabled through their corresponding enable bits in various registers. Individual interrupt bits are set regardless of the status of the GIE bit. The GIE bit is cleared on reset.

The "return from interrupt" instruction, RETFIE, exits the interrupt routine, as well as sets the GIE bit, which re-enables interrupts.

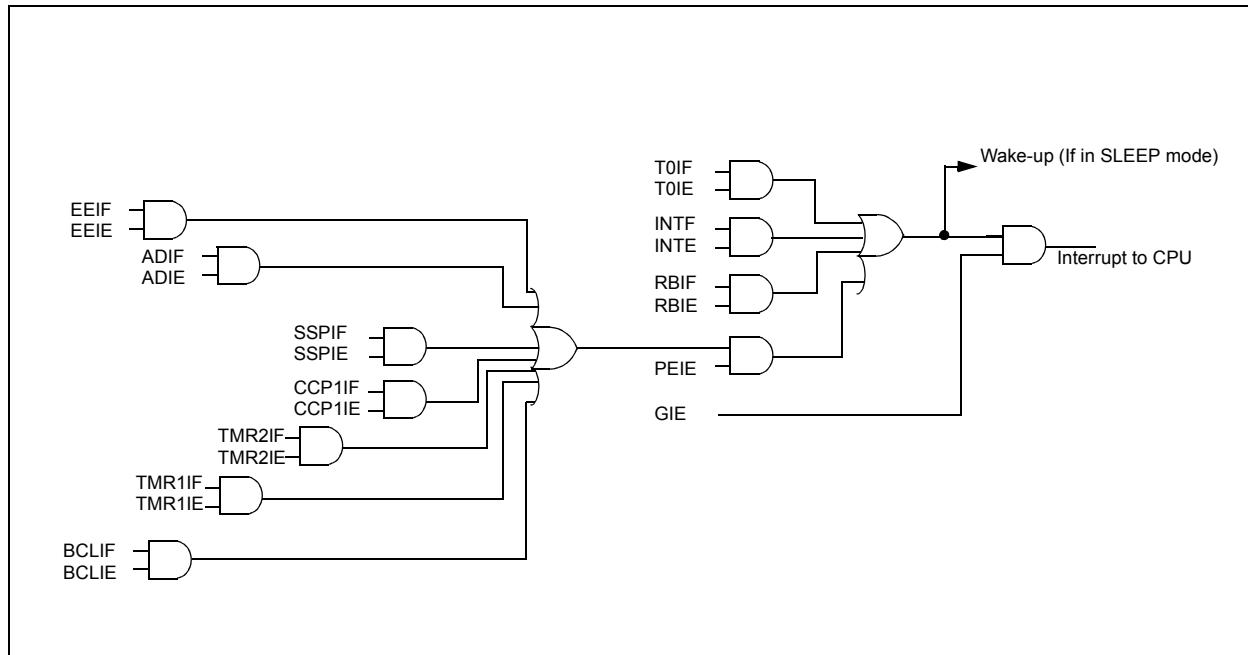
The RB0/INT pin interrupt, the RB port change interrupt and the TMR0 overflow interrupt flags are contained in the INTCON register.

The peripheral interrupt flags are contained in the Special Function Registers, PIR1 and PIR2. The corresponding interrupt enable bits are contained in Special Function Registers, PIE1 and PIE2, and the peripheral interrupt enable bit is contained in Special Function Register INTCON.

When an interrupt is responded to, the GIE bit is cleared to disable any further interrupt, the return address is pushed onto the stack and the PC is loaded with 0004h. Once in the interrupt service routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid recursive interrupts.

For external interrupt events, such as the INT pin or PORTB change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends when the interrupt event occurs. The latency is the same for one or two cycle instructions. Individual interrupt flag bits are set regardless of the status of their corresponding mask bit or the GIE bit

FIGURE 11-9: INTERRUPT LOGIC



## 12.0 INSTRUCTION SET SUMMARY

Each PIC16CXX instruction is a 14-bit word divided into an OPCODE which specifies the instruction type and one or more operands which further specify the operation of the instruction. The PIC16CXX instruction set summary in Table 12-2 lists **byte-oriented**, **bit-oriented**, and **literal and control** operations. Table 12-1 shows the opcode field descriptions.

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator which selects the number of the bit affected by the operation, while 'f' represents the number of the file in which the bit is located.

For **literal and control** operations, 'k' represents an eight or eleven bit constant or literal value.

**TABLE 12-1: OPCODE FIELD DESCRIPTIONS**

Field	Description
f	Register file address (0x00 to 0x7F)
w	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1) The assembler will generate code with x = 0. It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; d = 0: store result in W, d = 1: store result in file register f. Default is d = 1
PC	Program Counter
TO	Time-out bit
PD	Power-down bit

The instruction set is highly orthogonal and is grouped into three basic categories:

- **Byte-oriented** operations
- **Bit-oriented** operations
- **Literal and control** operations

All instructions are executed within one single instruction cycle, unless a conditional test is true or the program counter is changed as a result of an instruction. In this case, the execution takes two instruction cycles with the second cycle executed as a **NOP**. One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction

execution time is 1  $\mu$ s. If a conditional test is true or the program counter is changed as a result of an instruction, the instruction execution time is 2  $\mu$ s.

Table 12-2 lists the instructions recognized by the MPASM assembler.

Figure 12-1 shows the general formats that the instructions can have.

**Note:** To maintain upward compatibility with future PIC16CXX products, do not use the OPTION and TRIS instructions.

All examples use the following format to represent a hexadecimal number:

0xhh

where h signifies a hexadecimal digit.

**FIGURE 12-1: GENERAL FORMAT FOR INSTRUCTIONS**

Byte-oriented file register operations					
13	8	7	6		
OPCODE	d	f (FILE #)			
<i>d</i> = 0 for destination W <i>d</i> = 1 for destination f <i>f</i> = 7-bit file register address					
Bit-oriented file register operations					
13	10	9	7		
OPCODE	b (BIT #)	f (FILE #)			
<i>b</i> = 3-bit bit address <i>f</i> = 7-bit file register address					
Literal and control operations					
General					
13	8	7	0		
OPCODE	k (literal)				
<i>k</i> = 8-bit immediate value					
CALL and GOTO instructions only					
13	11	10	0		
OPCODE	k (literal)				
<i>k</i> = 11-bit immediate value					

A description of each instruction is available in the PICmicro™ Mid-Range Reference Manual, (DS33023).

# PIC16F872

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**TABLE 12-2: PIC16CXXX INSTRUCTION SET**

Mnemonic, Operands	Description	Cycles	14-Bit Opcode				Status Affected	Notes	
			MSb	LSb					
BYTE-ORIENTED FILE REGISTER OPERATIONS									
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C,DC,Z	
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	
CLRF	f	Clear f	1	00	0001	1fff	ffff	Z	
CLRW	-	Clear W	1	00	0001	0xxx	xxxx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	
DECFSZ	f, d	Decrement f	1	00	0011	dfff	ffff	Z	
DECF	f, d	Decrement f, Skip if 0	1(2)	00	1011	dfff	ffff	Z	
INCFSZ	f, d	Increment f	1	00	1010	dfff	ffff	Z	
INCFSZ	f, d	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff	Z	
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	
MOVWF	f	Move W to f	1	00	0000	1fff	ffff	Z	
NOP	-	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	C	
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C,DC,Z	
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff	Z	
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	
BIT-ORIENTED FILE REGISTER OPERATIONS									
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff	Z	
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff	Z	
BTFSZ	f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff	Z	
BTFSZ	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff	Z	
LITERAL AND CONTROL OPERATIONS									
ADDLW	k	Add literal and W	1	11	111x	kkkk	kkkk	C,DC,Z	
ANDLW	k	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	k	Call subroutine	2	10	0kkk	kkkk	kkkk		
CLRWD	-	Clear Watchdog Timer	1	00	0000	0110	0100	TO,PD	
GOTO	k	Go to address	2	10	1kkk	kkkk	kkkk	Z	
IORLW	k	Inclusive OR literal with W	1	11	1000	kkkk	kkkk		
MOVLW	k	Move literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	-	Return from interrupt	2	00	0000	0000	1001		
RETLW	k	Return with literal in W	2	11	01xx	kkkk	kkkk		
RETURN	-	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	-	Go into standby mode	1	00	0000	0110	0011	TO,PD	
SUBLW	k	Subtract W from literal	1	11	110x	kkkk	kkkk	C,DC,Z	
XORLW	k	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	

**Note 1:** When an I/O register is modified as a function of itself (e.g., MOVF PORTB, 1), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

**2:** If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 Module.

**3:** If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

**Note:** Additional information on the mid-range instruction set is available in the PICmicro™ Mid-Range MCU Family Reference Manual (DS33023).

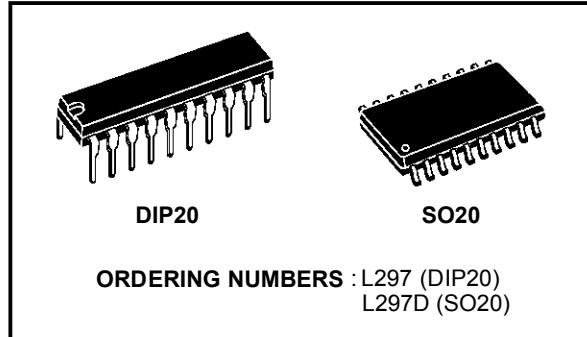


**SGS-THOMSON**  
MICROELECTRONICS

**L297**  
**L297D**

## STEPPER MOTOR CONTROLLERS

- NORMAL/WAVE DRIVE
- HALF/FULL STEP MODES
- CLOCKWISE/ANTICLOCKWISE DIRECTION
- SWITCHMODE LOAD CURRENT REGULATION
- PROGRAMMABLE LOAD CURRENT
- FEW EXTERNAL COMPONENTS
- RESET INPUT & HOME OUTPUT
- ENABLE INPUT



### DESCRIPTION

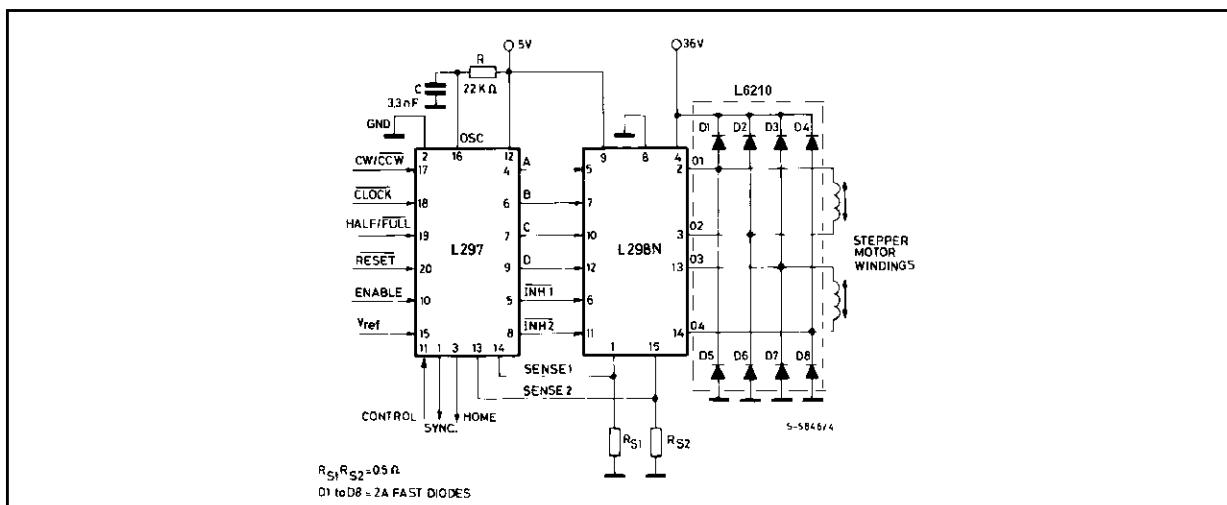
The L297/A/D Stepper Motor Controller IC generates four phase drive signals for two phase bipolar and four phase unipolar step motors in microcomputer-controlled applications. The motor can be driven in half step, normal and wave drive modes and on-chip PWM chopper circuits permit switch-mode control of the current in the windings. A

feature of this device is that it requires only clock, direction and mode input signals. Since the phase are generated internally the burden on the microprocessor, and the programmer, is greatly reduced. Mounted in DIP20 and SO20 packages, the L297 can be used with monolithic bridge drives such as the L298N or L293E, or with discrete transistors and darlingtons.

### ABSOLUTE MAXIMUM RATINGS

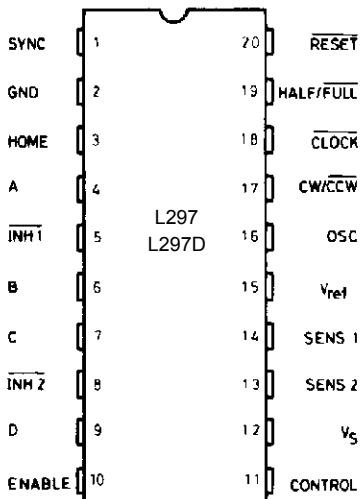
Symbol	Parameter	Value	Unit
$V_s$	Supply voltage	10	V
$V_i$	Input signals	7	V
$P_{tot}$	Total power dissipation ( $T_{amb} = 70^\circ\text{C}$ )	1	W
$T_{stg}, T_j$	Storage and junction temperature	-40 to + 150	$^\circ\text{C}$

### TWO PHASE BIPOLAR STEPPER MOTOR CONTROL CIRCUIT

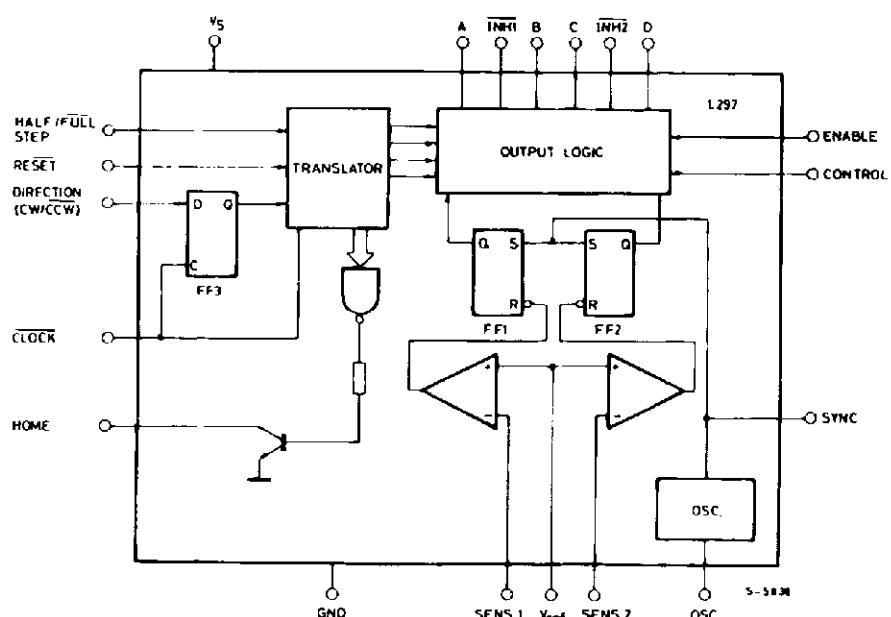


## L297-L297D

### PIN CONNECTION (Top view)



### BLOCK DIAGRAM (L297/L297D)



## PIN FUNCTIONS - L297/L297D

N°	NAME	FUNCTION
1	SYNC	Output of the on-chip chopper oscillator. The SYNC connections The SYNC connections of all L297s to be synchronized are connected together and the oscillator components are omitted on all but one. If an external clock source is used it is injected at this terminal.
2	GND	Ground connection.
3	HOME	Open collector output that indicates when the L297 is in its initial state (ABCD = 0101). The transistor is open when this signal is active.
4	A	Motor phase A drive signal for power stage.
5	INH1	Active low inhibit control for driver stage of A and B phases. When a bipolar bridge is used this signal can be used to ensure fast decay of load current when a winding is de-energized. Also used by chopper to regulate load current if CONTROL input is low.
6	B	Motor phase B drive signal for power stage.
7	C	Motor phase C drive signal for power stage.
8	INH2	Active low inhibit control for drive stages of C and D phases. Same functions as INH1.
9	D	Motor phase D drive signal for power stage.
10	ENABLE	Chip enable input. When low (inactive) INH1, INH2, A, B, C and D are brought low.
11	CONTROL	Control input that defines action of chopper. When low chopper acts on INH1 and INH2; when high chopper acts on phase lines ABCD.
12	V <sub>s</sub>	5V supply input.
13	SENS <sub>2</sub>	Input for load current sense voltage from power stages of phases C and D.
14	SENS <sub>1</sub>	Input for load current sense voltage from power stages of phases A and B.
15	V <sub>ref</sub>	Reference voltage for chopper circuit. A voltage applied to this pin determines the peak load current.
16	OSC	An RC network (R to V <sub>CC</sub> , C to ground) connected to this terminal determines the chopper rate. This terminal is connected to ground on all but one device in synchronized multi - L297 configurations f ≡ 1/0.69 RC
17	CW/CCW	Clockwise/counterclockwise direction control input. Physical direction of motor rotation also depends on connection of windings. Synchronized internally therefore direction can be changed at any time.
18	CLOCK	Step clock. An active low pulse on this input advances the motor one increment. The step occurs on the rising edge of this signal.

**PIN FUNCTIONS - L297/L297D (continued)**

N°	NAME	FUNCTION
19	HALF/FULL	Half/full step select input. When high selects half step operation, when low selects full step operation. One-phase-on full step mode is obtained by selecting FULL when the L297's translator is at an even-numbered state. Two-phase-on full step mode is set by selecting FULL when the translator is at an odd numbered position. (The home position is designate state 1).
20	RESET	Reset input. An active low pulse on this input restores the translator to the home position (state 1, ABCD = 0101).

**THERMAL DATA**

Symbol	Parameter	DIP20	SO20	Unit
R <sub>th-j-amb</sub>	Thermal resistance junction-ambient	max	80	100 °C/W

**CIRCUIT OPERATION**

The L297 is intended for use with a dual bridge driver, quad darlington array or discrete power devices in step motor driving applications. It receives step clock, direction and mode signals from the systems controller (usually a microcomputer chip) and generates control signals for the power stage.

The principal functions are a translator, which generates the motor phase sequences, and a dual PWM chopper circuit which regulates the current in the motor windings. The translator generates three different sequences, selected by the HALF/FULL input. These are normal (two phases energised), wave drive (one phase energised) and half-step (alternately one phase energised/two phases energised). Two inhibit signals are also generated by the L297 in half step and wave drive modes. These signals, which connect directly to the L298's enable inputs, are intended to speed current decay when a winding is de-energised. When the L297 is used to drive a unipolar motor the chopper acts on these lines.

An input called CONTROL determines whether the chopper will act on the phase lines ABCD or the inhibit lines INH1 and INH2. When the phase lines

are chopped the non-active phase line of each pair (AB or CD) is activated (rather than interrupting the line then active). In L297 + L298 configurations this technique reduces dissipation in the load current sense resistors.

A common on-chip oscillator drives the dual chopper. It supplies pulses at the chopper rate which set the two flip-flops FF1 and FF2. When the current in a winding reaches the programmed peak value the voltage across the sense resistor (connected to one of the sense inputs SENS<sub>1</sub> or SENS<sub>2</sub>) equals V<sub>ref</sub> and the corresponding comparator resets its flip flop, interrupting the drive current until the next oscillator pulse arrives. The peak current for both windings is programmed by a voltage divider on the V<sub>ref</sub> input.

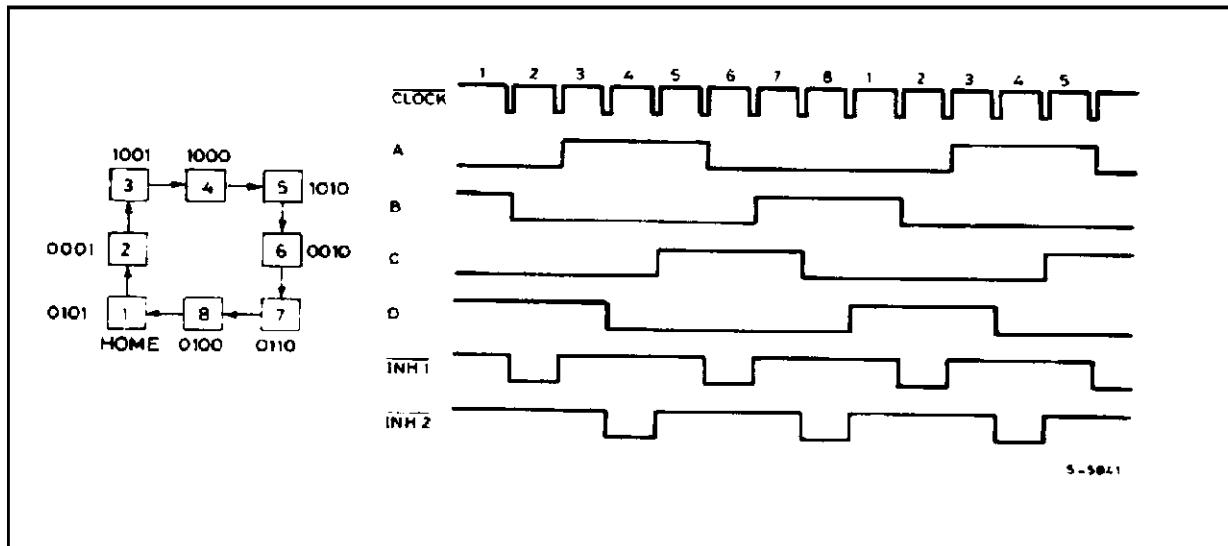
Ground noise problems in multiple configurations can be avoided by synchronising the chopper oscillators. This is done by connecting all the SYNC pins together, mounting the oscillator RC network on one device only and grounding the OSC pin on all other devices.

## MOTOR DRIVING PHASE SEQUENCES

The L297's translator generates phase sequences for normal drive, wave drive and half step modes. The state sequences and output waveforms for these three modes are shown below. In all cases the translator advances on the low to high transition of **CLOCK**.

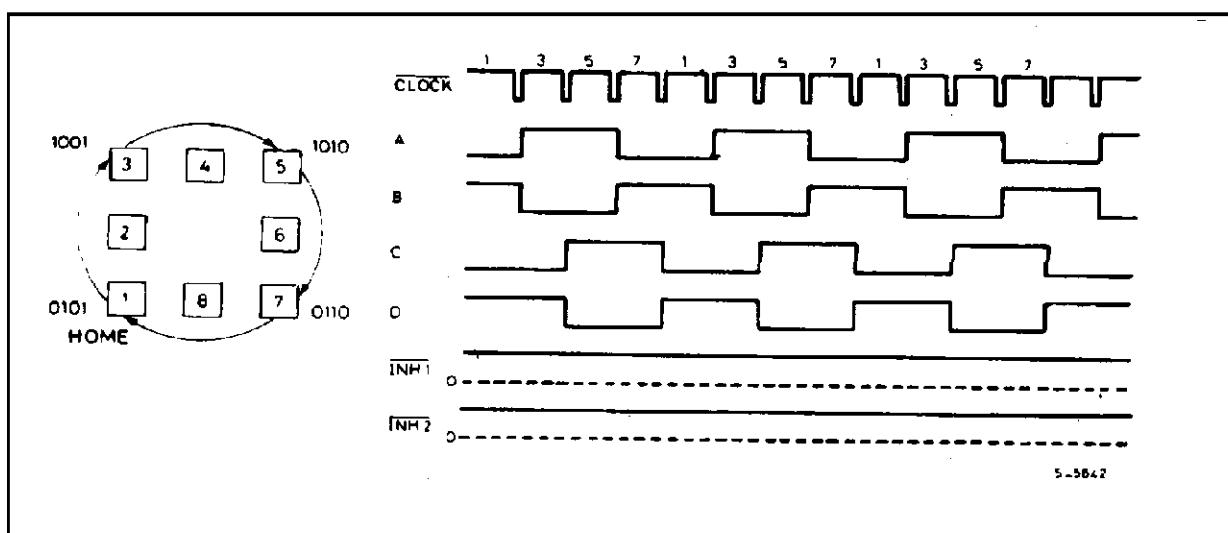
### HALF STEP MODE

Half step mode is selected by a high level on the **HALF/FULL** input.



### NORMAL DRIVE MODE

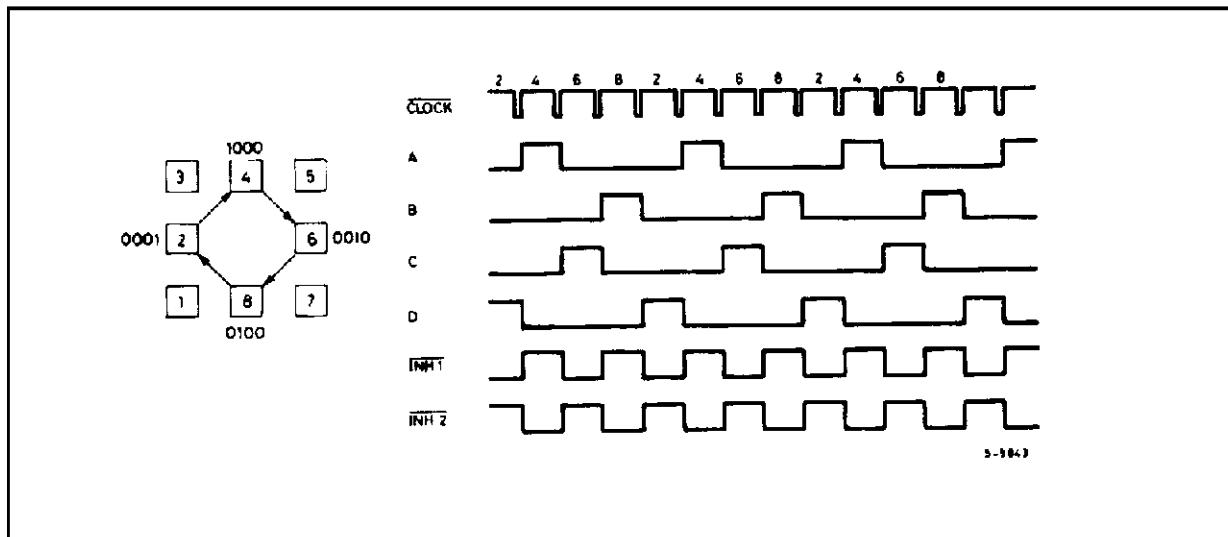
Normal drive mode (also called "two-phase-on" drive) is selected by a low level on the **HALF/FULL** input when the translator is at an odd numbered state (1, 3, 5 or 7). In this mode the **INH1** and **INH2** outputs remain high throughout.



**MOTOR DRIVING PHASE SEQUENCES** (continued)

**WAVE DRIVE MODE**

Wave drive mode (also called "one-phase-on" drive) is selected by a low level on the HALF/FULL input when the translator is at an even numbered state (2, 4, 6 or 8).



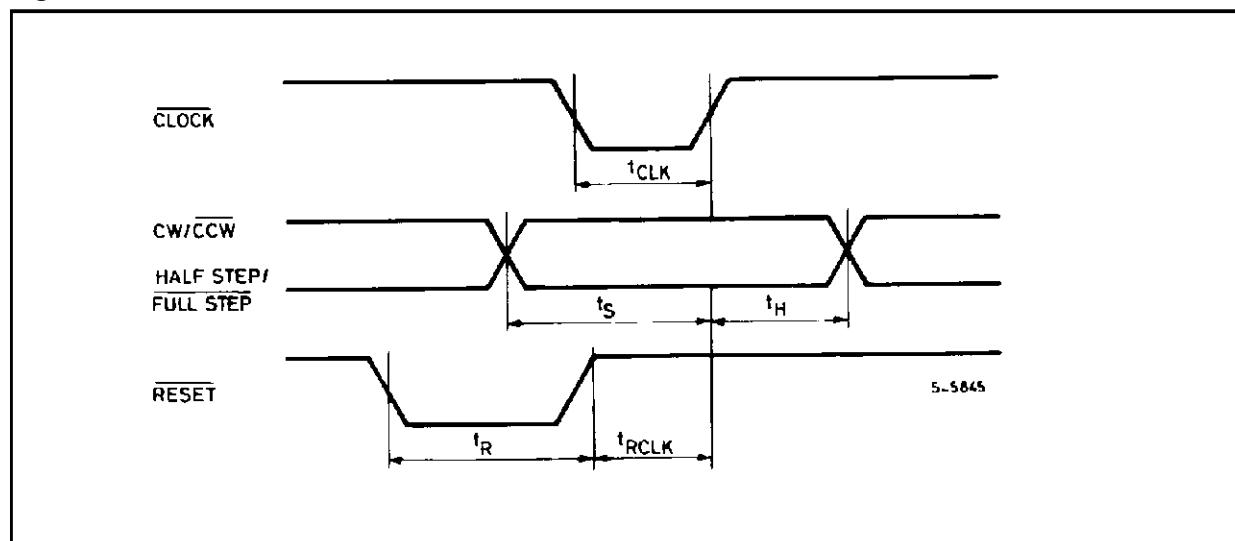
**ELECTRICAL CHARACTERISTICS** (Refer to the block diagram  $T_{amb} = 25^{\circ}\text{C}$ ,  $V_s = 5\text{V}$  unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ	Max.	Unit
$V_s$	Supply voltage (pin 12)		4.75		7	V
$I_s$	Quiescent supply current (pin 12)	Outputs floating		50	80	mA
$V_i$	Input voltage (pin 11, 17, 18, 19, 20)	Low			0.6	V
		High	2		$V_s$	V
$I_i$	Input current (pin 11, 17, 18, 19, 20)	$V_i = L$	100	$\mu\text{A}$		
		$V_i = H$			10	$\mu\text{A}$
$V_{en}$	Enable input voltage (pin 10)	Low			1.3	V
		High	2		$V_s$	V
$I_{en}$	Enable input current (pin 10)	$V_{en} = L$			100	$\mu\text{A}$
		$V_{en} = H$			10	$\mu\text{A}$
$V_o$	Phase output voltage (pins 4, 6, 7, 9)	$I_o = 10\text{mA}$	$V_{OL}$		0.4	V
		$I_o = 5\text{mA}$	$V_{OH}$	3.9		V
$V_{inh}$	Inhibit output voltage (pins 5, 8)	$I_o = 10\text{mA}$	$V_{inh\ L}$		0.4	V
		$I_o = 5\text{mA}$	$V_{inh\ H}$	3.9		V
$V_{SYNC}$	Sync Output Voltage	$I_o = 5\text{mA}$	$V_{SYNC\ H}$	3.3		V
		$I_o = 5\text{mA}$	$V_{SYNC\ V}$		0.8	

## ELECTRICAL CHARACTERISTICS (continued)

Symbol	Parameter	Test conditions	Min.	Typ	Max.	Unit
$I_{\text{leak}}$	Leakage current (pin 3)	$V_{\text{CE}} = 7 \text{ V}$			1	$\mu\text{A}$
$V_{\text{sat}}$	Saturation voltage (pin 3)	$I = 5 \text{ mA}$			0.4	V
$V_{\text{off}}$	Comparators offset voltage (pins 13, 14, 15)	$V_{\text{ref}} = 1 \text{ V}$			5	mV
$I_o$	Comparator bias current (pins 13, 14, 15)		-100		10	$\mu\text{A}$
$V_{\text{ref}}$	Input reference voltage (pin 15)		0		3	V
$t_{\text{CLK}}$	Clock time		0.5			$\mu\text{s}$
$t_s$	Set up time		1			$\mu\text{s}$
$t_h$	Hold time		4			$\mu\text{s}$
$t_R$	Reset time		1			$\mu\text{s}$
$t_{\text{RCLK}}$	Reset to clock delay		1			$\mu\text{s}$

Figure 1.



## ⇒ FOTO REALIZZAZIONE



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